TS921

## Rail-to-rail high output current single operational amplifier

Datasheet -production data

## Features

- Rail-to-rail input and output

■ Low noise: $9 \mathrm{nV} / \mathrm{VHz}$
■ Low distortion
■ High output current: 80 mA (able to drive $32 \Omega$ loads)

■ High-speed: $4 \mathrm{MHz}, 1 \mathrm{~V} / \mu \mathrm{s}$
■ Operating from 2.7 V to 12 V
■ ESD internal protection: 1.5 kV
■ Latch-up immunity

- Macromodel included in this specification


## Applications

- Headphone amplifier
- Piezoelectric speaker driver

■ Sound cards, multimedia systems

- Line driver, actuator driver
- Servo amplifier
- Mobile phone and portable communication sets
- Instrumentation with low noise as key factor

Table 1. Device summary

| Order code | Temperature <br> range | Package | Packing | Marking |
| :--- | :---: | :---: | :---: | :---: |
| TS921IN | $-40^{\circ} \mathrm{C},+125^{\circ} \mathrm{C}$ | DIP8 | Tube | TS921IN |
| TS921ID/IDT |  | SO-8 | Tube or tape and reel | 921I |
| TS921IPT |  | TSSOP8 <br> (thin shrink outline package) | Tape and reel |  |

## 1 <br> Description

The TS921 device is a rail-to-rail single BiCMOS operational amplifier optimized and fully specified for 3 V and 5 V operation.
Its high output current allows low load impedances to be driven.
The TS921 device exhibits very low noise, low distortion and low offset. It has a high output current capability which makes this device an excellent choice for high quality, low voltage or battery operated audio systems.
The device is stable for capacitive loads up to 500 pF .

## 2 Absolute maximum ratings

Table 2. Key parameters and their absolute maximum ratings

| Symbol | Parameter | Condition | Value | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage ${ }^{(1)}$ |  | 14 | V |
| $V_{\text {id }}$ | Differential input voltage ${ }^{(2)}$ |  | $\pm 1$ | V |
| $\mathrm{V}_{\mathrm{i}}$ | Input voltage |  | $\mathrm{V}_{\mathrm{DD}}-0.3$ to $\mathrm{V}_{\mathrm{CC}}+0.3$ | V |
| $\mathrm{T}_{\text {stg }}$ | Storage temperature |  | -65 to +150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{j}}$ | Maximum junction temperature |  | 150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{R}_{\text {thja }}$ | Thermal resistance junction-to-ambient | $\begin{aligned} & \hline \text { SO-8 } \\ & \text { TSSOP8 } \\ & \text { DIP8 } \end{aligned}$ | $\begin{gathered} \hline 125 \\ 120 \\ 85 \end{gathered}$ | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\mathrm{R}_{\text {thic }}$ | Thermal resistance junction-to-case | $\begin{aligned} & \text { SO-8 } \\ & \text { TSSOP8 } \\ & \text { DIP8 } \end{aligned}$ | $\begin{aligned} & 40 \\ & 37 \\ & 41 \end{aligned}$ | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| ESD | Electrostatic discharge | HBM <br> Human body model ${ }^{(3)}$ | 1.5 | kV |
|  |  | MM <br> Machine model ${ }^{(4)}$ | 100 | V |
|  |  | CDM <br> Charged device model | 1.5 | kV |
|  | Output short-circuit duration |  | See ${ }^{(5)}$ |  |
|  | Latch-up immunity |  | 200 | mA |
|  | Soldering temperature | $\begin{aligned} & 10 \text { sec., } \\ & \text { standard package } \end{aligned}$ | 250 | ${ }^{\circ} \mathrm{C}$ |
|  |  | $\begin{aligned} & 10 \mathrm{sec} ., \\ & \text { lead-free package } \end{aligned}$ | 260 |  |

1. All voltage values, except differential voltage are with respect to network ground terminal.
2. Differential voltages are the non-inverting input terminal with respect to the inverting input terminal. If $\mathrm{V}_{\text {id }}> \pm 1 \mathrm{~V}$, the maximum input current must not exceed $\pm 1 \mathrm{~mA}$. In this case ( $\mathrm{V}_{\text {id }}> \pm 1 \mathrm{~V}$ ) an input serie resistor must be added to limit input current.
3. Human body model, 100 pF discharged through a $1.5 \mathrm{k} \Omega$ resistor into pin of device.
4. Machine model ESD, a 200 pF cap is charged to the specified voltage, then discharged directly into the IC with no external series resistor (internal resistor $<5 \Omega$ ), into pin to pin of device.
5. There is no short-circuit protection inside the device: short-circuits from the output to $\mathrm{V}_{\mathrm{CC}}$ can cause excessive heating. The maximum output current is approximately 80 mA , independent of the magnitude of $\mathrm{V}_{\mathrm{Cc}}$. Destructive dissipation can result from simultaneous short-circuits on all amplifiers.

Table 3. Operating conditions

| Symbol | Parameter | Value | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | 2.7 to 12 | V |
| $\mathrm{~V}_{\mathrm{icm}}$ | Common mode input voltage range | $\mathrm{V}_{\mathrm{DD}}-0.2$ to $\mathrm{V}_{\mathrm{CC}}+0.2$ | V |
| $\mathrm{~T}_{\text {oper }}$ | Operating free air temperature range | -40 to +125 | ${ }^{\circ} \mathrm{C}$ |

## 3 Electrical characteristics

Table 4. Electrical characteristics for $\mathrm{V}_{\mathrm{Cc}}=3 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{icm}}=\mathrm{V}_{\mathrm{Cc}} / 2, \mathrm{R}_{\mathrm{L}}$ connected to $\mathrm{V}_{\mathrm{CC}} / 2, \mathrm{~T}_{\mathrm{amb}}=25{ }^{\circ} \mathrm{C}$ (unless otherwise specified)

| Symbol | Parameter | Conditions | Min. | Typ. | Max. | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {io }}$ | Input offset voltage | at $T_{\text {min }} \leq T_{\text {amb }} \leq T_{\max }$ |  |  |  |  |$)$

Table 5. Electrical characteristics for $\mathrm{V}_{\mathrm{Cc}}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{icm}}=\mathrm{V}_{\mathrm{Cc}} / 2$, $\mathrm{R}_{\mathrm{L}}$ connected to $\mathrm{V}_{\mathrm{Cc}} / 2$, $\mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}$ (unless otherwise specified)

| Symbol | Parameter | Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {io }}$ | Input offset voltage | at $\mathrm{T}_{\text {min. }} \leq \mathrm{T}_{\text {amb }} \leq \mathrm{T}_{\text {max }}$ |  |  | $\begin{aligned} & 3 \\ & 5 \end{aligned}$ | mV |
| $\Delta \mathrm{V}_{\text {io }}$ | Input offset voltage drift |  |  | 2 |  | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| $\mathrm{I}_{\mathrm{i}}$ | Input offset current | $\mathrm{V}_{\text {out }}=1.5 \mathrm{~V}$ |  | 1 | 30 | nA |
| $\mathrm{l}_{\text {ib }}$ | Input bias current | $\mathrm{V}_{\text {out }}=1.5 \mathrm{~V}$ |  | 15 | 100 | nA |
| $\mathrm{V}_{\mathrm{OH}}$ | High level output voltage | $\begin{aligned} & \mathrm{R}_{\mathrm{L}}=600 \Omega \\ & \mathrm{R}_{\mathrm{L}}=32 \Omega \end{aligned}$ | 4.85 | 4.4 |  | V |
| $\mathrm{V}_{\text {OL }}$ | Low level output voltage | $\begin{aligned} & \mathrm{R}_{\mathrm{L}}=600 \Omega \\ & \mathrm{R}_{\mathrm{L}}=32 \Omega \end{aligned}$ |  | 300 | 120 | mV |
| $\mathrm{A}_{\mathrm{vd}}$ | Large signal voltage gain | $\begin{aligned} & \mathrm{V}_{\text {out }}=2 \mathrm{~V}_{\text {pk-pk }} \\ & \mathrm{R}_{\mathrm{L}}=600 \Omega \\ & \mathrm{R}_{\mathrm{L}}=32 \Omega \end{aligned}$ |  | $\begin{aligned} & 35 \\ & 16 \end{aligned}$ |  | $\mathrm{V} / \mathrm{mV}$ |
| GBP | Gain bandwidth product | $\mathrm{R}_{\mathrm{L}}=600 \Omega$ |  | 4 |  | MHz |
| $\mathrm{I}_{\mathrm{CC}}$ | Supply current | No load, $\mathrm{V}_{\text {out }}=\mathrm{V}_{\mathrm{CC}} / 2$ |  | 1 | 1.5 | mA |
| CMR | Common mode rejection ratio |  | 60 | 80 |  | dB |
| SVR | Supply voltage rejection ratio | $\mathrm{V}_{\mathrm{CC}}=4.5$ to 5.5 V | 60 | 80 |  | dB |
| $I_{0}$ | Output short-circuit current |  | 50 | 80 |  | mA |
| SR | Slew rate |  | 0.7 | 1.3 |  | V/ $\mu \mathrm{s}$ |
| Pm | Phase margin at unit gain | $\mathrm{R}_{\mathrm{L}}=600 \Omega, \mathrm{C}_{\mathrm{L}}=100 \mathrm{pF}$ |  | 68 |  | Degrees |
| GM | Gain margin | $\mathrm{R}_{\mathrm{L}}=600 \Omega, \mathrm{C}_{\mathrm{L}}=100 \mathrm{pF}$ |  | 12 |  | dB |
| $e_{n}$ | Equivalent input noise voltage | $\mathrm{f}=1 \mathrm{kHz}$ |  | 9 |  | $\frac{\mathrm{nV}}{\sqrt{\mathrm{Hz}}}$ |
| THD | Total harmonic distortion | $\begin{aligned} & V_{\text {out }}=2 V_{\text {pk-pk }}, f=1 \mathrm{kHz}, \\ & A_{v}=1, R_{L}=600 \Omega \end{aligned}$ |  | 0.005 |  | \% |

Figure 1. Output short-circuit vs. output voltage $\left(\mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD}}=0 \mathrm{~V}\right)$

Figure 2. Voltage gain and phase vs.
frequency $\left(R_{L}=10 \mathrm{k} \Omega, C_{L}=100 \mathrm{pF}\right)$


Figure 3. Output short-circuit vs. output voltage $\left(\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD}}=0 \mathrm{~V}\right)$


Figure 4. Equivalent input noise voltage vs. frequency $\left(\mathrm{V}_{\mathrm{CC}}= \pm 1.5 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=100 \Omega\right)$


Figure 5. Output supply current vs. supply voltage


Figure 6. $\quad \mathrm{THD}+$ noise vs. frequency $\left(\mathrm{R}_{\mathrm{L}}=\mathbf{2 k} \Omega\right.$ $\mathrm{V}_{\mathrm{o}}=10 \mathrm{Vpp}, \mathrm{V}_{\mathrm{CC}}= \pm 6 \mathrm{~V}, \mathrm{~A}_{\mathrm{V}}=1$ )


Figure 7. THD + noise vs. frequency $\left(R_{L}=32 \Omega \quad V_{0}=4 \mathrm{Vpp}\right.$, $\mathrm{V}_{\mathrm{CC}}= \pm 2.5 \mathrm{~V}, \mathrm{~A}_{\mathrm{v}}=1$ )


Figure 8. THD + noise vs. output voltage
$\left(R_{L}=600 \Omega, f=1 \mathrm{kHz}\right.$,
$V_{C C}=0 / 3 V, A_{v}=-1$ )


Figure 9. THD + noise vs. frequency
( $\mathrm{R}_{\mathrm{L}}=32 \Omega, \mathrm{~V}_{\mathrm{o}}=2 \mathrm{Vpp}$,
$\mathrm{V}_{\mathrm{CC}}= \pm 1.5 \mathrm{~V}, \mathrm{~A}_{\mathrm{V}}=10$ )


Figure 10. THD + noise vs. output voltage
( $\mathrm{R}_{\mathrm{L}}=32 \Omega \mathrm{f}=\mathbf{1} \mathrm{kHz}$,
$\mathrm{V}_{\mathrm{CC}}= \pm 1.5 \mathrm{~V}, \mathrm{~A}_{\mathrm{v}}=-1$ )


Figure 12. Open loop gain and phase vs. frequency $\left(C_{L}=500 \mathrm{pF}\right)$
$\left(R_{L}=2 \mathrm{k} \Omega \mathrm{f}=\mathbf{1} \mathrm{kHz}\right.$, $\mathrm{V}_{\mathrm{CC}}= \pm 1.5 \mathrm{~V}, \mathrm{~A}_{\mathrm{V}}=-1$ )


## 4 Macromodel

### 4.1 Important note concerning this macromodel

Please consider following remarks before using this macromodel:

- All models are a trade-off between accuracy and complexity (i.e. simulation time).
- Macromodels are not a substitute to breadboarding; rather, they confirm the validity of a design approach and help to select surrounding component values.
- A macromodel emulates the NOMINAL performance of a TYPICAL device within SPECIFIED OPERATING CONDITIONS (i.e. temperature, supply voltage, etc.). Thus the macromodel is often not as exhaustive as the datasheet, its goal is to illustrate the main parameters of the product.
- Data issued from macromodels used outside of its specified conditions
( $\mathrm{V}_{\mathrm{CC}}$, temperature, etc.) or even worse: outside of the device operating conditions $\left(\mathrm{V}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{icm}}\right.$, etc.) are not reliable in any way.
In Section 4.3, the electrical characteristics resulting from the use of these macromodels are presented.


### 4.2 Electrical characteristics from macromodelization

Table 6. Electrical characteristics resulting from macromodel simulation at $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$, $\mathrm{V}_{\mathrm{DD}}=0 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}, \mathrm{C}_{\mathrm{L}}$ connected to $\mathrm{V}_{\mathrm{CC} / 2}, \mathrm{~T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}$ (unless otherwise specified)

| Symbol | Conditions | Value | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{io}}$ |  | 0 | mV |
| $\mathrm{A}_{\mathrm{vd}}$ | $\mathrm{R}_{\mathrm{L}}=10 \mathrm{k} \Omega$ | 200 | $\mathrm{~V} / \mathrm{mV}$ |
| $\mathrm{I}_{\mathrm{CC}}$ | No load, per operator | 1.2 | mA |
| $\mathrm{~V}_{\mathrm{icm}}$ |  | -0.2 to 3.2 | V |
| $\mathrm{~V}_{\mathrm{OH}}$ | $\mathrm{R}_{\mathrm{L}}=10 \mathrm{k} \Omega$ | 2.95 | V |
| $\mathrm{~V}_{\mathrm{OL}}$ | $\mathrm{R}_{\mathrm{L}}=10 \mathrm{k} \Omega$ | 25 | mV |
| $\mathrm{I}_{\text {sink }}$ | $\mathrm{V}_{\mathrm{O}}=3 \mathrm{~V}$ | 80 | mA |
| $\mathrm{I}_{\text {source }}$ | $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ | 80 | mA |
| GBP | $\mathrm{R}_{\mathrm{L}}=600 \mathrm{k} \Omega$ | 4 | MHz |
| SR | $\mathrm{R}_{\mathrm{L}}=10 \mathrm{k} \Omega, \mathrm{C}_{\mathrm{L}}=100 \mathrm{pF}$ | 1.3 | $\mathrm{~V} / \mu \mathrm{s}$ |
| $\phi \mathrm{m}$ | $\mathrm{R}_{\mathrm{L}}=600 \mathrm{k} \Omega$ | 68 | Degrees |

### 4.3 Macromodel code

** Standard Linear Ics Macromodels, 1996.
** CONNECTIONS:

* 1 INVERTING INPUT
* 2 NON-INVERTING INPUT
* 3 OUTPUT
* 4 POSITIVE POWER SUPPLY
* 5 NEGATIVE POWER SUPPLY
. SUBCKT TS921 13245 (analog)
********************************************************* . MODEL MDTH D IS $=1 \mathrm{E}-8 \mathrm{KF}=2.664234 \mathrm{E}-16 \mathrm{CJO}=10 \mathrm{~F}$
* INPUT STAGE

CIP 25 1.000000E-12
CIN $151.000000 \mathrm{E}-12$
EIP $10 \quad 5 \quad 2 \quad 51$
EIN $16 \begin{array}{llll}16 & 5 & 5 & 1\end{array}$
RIP $10118.125000 \mathrm{E}+00$
RIN $1516 \quad 8.125000 \mathrm{E}+00$
RIS 1115 2.238465E+02
DIP 1112 MDTH 400E-12
DIN 1514 MDTH 400E-12
VOFP 1213 DC 153.5u
VOFN 1314 DC 0
IPOL $1353.200000 \mathrm{E}-05$
CPS 1115 1e-9
DINN 1713 MDTH 400E-12
VIN $175-0.100000 \mathrm{e}+00$
DINR 1518 MDTH 400E-12
VIP $4180.400000 \mathrm{E}+00$
FCP 45 VOFP 1.865000E+02
FCN 54 VOFN 1.865000E+02
FIBP 25 VOFP 6.250000E-03
FIBN 51 VOFN 6.250000E-03

* GM1 STAGE ***************

FGM1P 1195 VOFP 1.1
FGM1N 1195 VOFN 1.1
RAP $11942.6 E+06$
RAN $11952.6 \mathrm{E}+06$

* GM2 STAGE ***************

G2P $19511951.92 \mathrm{E}-02$
G2N $19511941.92 \mathrm{E}-02$
R2P 194 1E+07
R2N 195 1E+07

VINT1 50005

GCONVP $500501119419.38!$ send ds VP, $I(V P)=(V 119-V 4) / 2 / U t$ VP 50100
GCONVN $500502119519.38!s e n d ~ d s ~ V N, ~ I(V N)=(V 119-V 5) / 2 / U t$ VN 50200
********* orientation isink isource *******
VINT2 50305
FCOPY 503504 VOUT 1
DCOPYP 504505 MDTH 400E-9
VCOPYP 50500
DCOPYN 506504 MDTH 400E-9
VCOPYN 05060

F2PP 195 poly(2) VCOPYP VP $00000.5!m u l t i p l y ~ I(v o u t) * I(V P)=I o u t *(V 119-$ V4) / 2 /Ut

F2PN 195 poly(2) VCOPYP VN 00000.5 !multiply 0 (vout)*I(VN)=Iout*(V119V5) / 2 /Ut

F2NP 195 poly(2) VCOPYN VP 00001.75 !multiply I(vout)*I(VP)=Iout*(V119V4) / 2 /Ut
F2NN 195 poly(2) VCOPYN VN 00001.75 ! multiply I(vout)*I(VN)=Iout*(V119V5 ) / 2 /Ut

* COMPENSATION

CC 19119 25p

* OUTPUT***********

DOPM 1922 MDTH 400E-12
DONM 2119 MDTH 400E-12
HOPM 2228 VOUT 6.250000E+02
VIPM $2845.000000 \mathrm{E}+01$
HONM 2127 VOUT 6.250000E+02
VINM 527 5.000000E+01
VOUT 3230
ROUT $2319 \quad 6$
COUT $351.300000 \mathrm{E}-10$
DOP 1925 MDTH 400E-12
VOP 4251.052
DON 2419 MDTH 400E-12
VON 2451.052
. ENDS

## 5 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK ${ }^{\circledR}$ packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

Figure 13. DIP8 package outline


Table 7. DIP8 package mechanical data

| Symbol | Dimensions |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | mm |  |  |  | Max. | Min. |
|  | Min. | Typ. | Typ. | Max. |  |  |
| A |  | 3.3 |  |  | 0.130 |  |
| a1 | 0.7 |  |  | 0.028 |  |  |
| B | 1.39 |  | 1.65 | 0.055 |  | 0.065 |
| B1 | 0.91 |  | 1.04 | 0.036 |  | 0.041 |
| b |  | 0.5 |  |  | 0.020 |  |
| b1 | 0.38 |  | 0.5 | 0.015 |  | 0.020 |
| D |  |  | 9.8 |  |  | 0.386 |
| E |  | 8.8 |  |  | 0.346 |  |
| e |  | 2.54 |  |  | 0.100 |  |
| e3 |  | 7.62 |  |  | 0.300 |  |
| e4 |  | 7.62 |  |  | 0.300 |  |
| F |  |  | 7.1 |  |  | 0.280 |
| I |  |  | 4.8 |  |  | 0.189 |
| L |  | 3.3 |  | 1.6 | 0.017 |  |
| Z | 0.44 |  |  |  | 0.130 |  |

Figure 14. SO-8 package outline


0016023/C

Table 8. SO-8 package mechanical data

| Symbol | Dimensions |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | mm |  |  | inch |  |  |
|  | Min. | Typ. | Max. | Min. | Typ. | Max. |
| A | 1.35 |  | 1.75 | 0.053 |  | 0.069 |
| A1 | 0.10 |  | 0.25 | 0.04 |  | 0.010 |
| A2 | 1.10 |  | 1.65 | 0.043 |  | 0.065 |
| B | 0.33 |  | 0.51 | 0.013 |  | 0.020 |
| C | 0.19 |  | 0.25 | 0.007 |  | 0.010 |
| D | 4.80 |  | 5.00 | 0.189 |  | 0.197 |
| E | 3.80 |  | 4.00 | 0.150 |  | 0.157 |
| e |  | 1.27 |  |  | 0.050 |  |
| H | 5.80 |  | 6.20 | 0.228 |  | 0.244 |
| h | 0.25 |  | 0.50 | 0.010 |  | 0.020 |
| L | 0.40 |  | 1.27 | 0.016 |  | 0.050 |
| k | $8^{\circ}$ (max.) |  |  |  |  |  |
| ddd |  |  | 0.1 |  |  | 0.04 |

Figure 15. TSSOP8 package outline


Table 9. TSSOP8 package mechanical data

| Symbol | Dimensions |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | mm |  |  | inch |  |  |
|  | Min. | Typ. | Max. | Min. | Typ. | Max. |
| A |  |  | 1.2 |  |  | 0.047 |
| A1 | 0.05 |  | 0.15 | 0.002 |  | 0.006 |
| A2 | 0.80 | 1.00 | 1.05 | 0.031 | 0.039 | 0.041 |
| b | 0.19 |  | 0.30 | 0.007 |  | 0.012 |
| c | 0.09 |  | 0.20 | 0.004 |  | 0.008 |
| D | 2.90 | 3.00 | 3.10 | 0.114 | 0.118 | 0.122 |
| E | 6.20 | 6.40 | 6.60 | 0.244 | 0.252 | 0.260 |
| E1 | 4.30 | 4.40 | 4.50 | 0.169 | 0.173 | 0.177 |
| e |  | 0.65 |  |  | 0.0256 |  |
| K | $0^{\circ}$ |  | $8^{\circ}$ | $0^{\circ}$ |  | $8^{\circ}$ |
| L | 0.45 | 0.60 | 0.75 | 0.018 | 0.024 | 0.030 |
| L1 |  | 1 |  |  | 0.039 |  |

## 6 Revision history

Table 10. Document revision history

| Date | Revision | Changes |
| :---: | :---: | :--- |
| Feb. 2001 | 1 | Initial release - Product in full production. |
| Dec. 2004 | 2 | Modifications on AMR table page 2 (explanation of $\mathrm{V}_{\text {id }}$ and $\mathrm{V}_{\mathrm{i}}$ limits, <br> ESD, MM and CDM values added, Rthja added) |
| Nov. 2005 | 3 | The following changes were made in this revision: <br> PPAP references inserted in the datasheet see Table 1. <br> Data in tables Electrical characteristics on page 4 reformatted for <br> easier use. <br> Thermal Resistance Junction to Case added in Table 2 on page 3. |
| 19-Sep-2012 | 4 | Updated Figure on page 1(replaced VCC <br> Updated (renamed) Table 1, removed TS921IYD/IYDT devices from <br> Table 1. <br> Moved Description to page 2. <br> Updated Figure 1 to Figure 4, Figure 6 to Figure 12 (added <br> conditions to titles). <br> Updated ECOPACK text and reformatted Section 5 (added Table 7to <br> Table 9, reversed order of figures and tables). <br> Minor corrections throughout document. |

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