



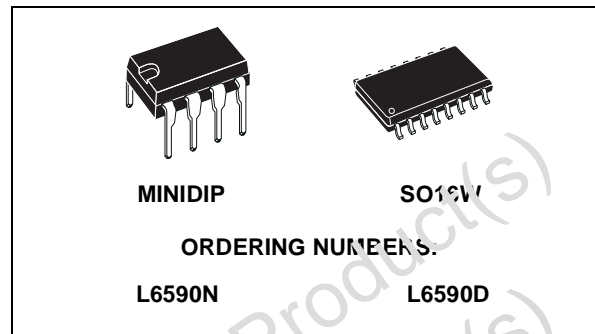
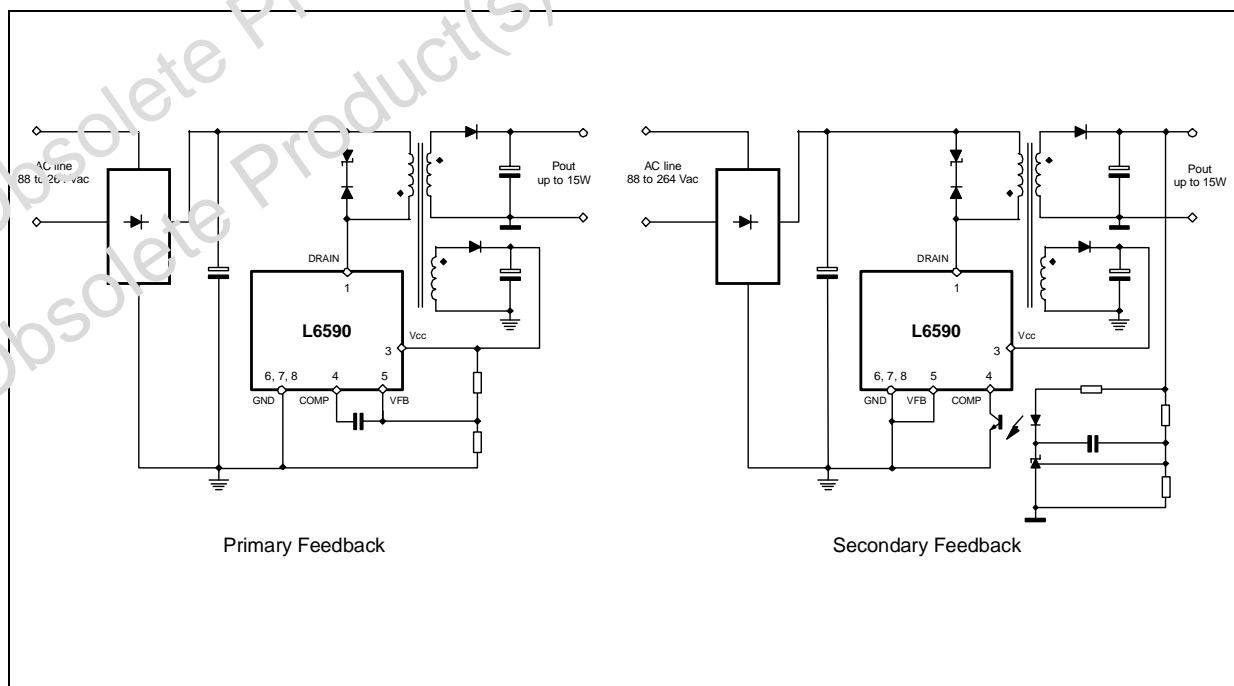
## FULLY INTEGRATED POWER SUPPLY

- WIDE-RANGE MAINS OPERATION
- "ON-CHIP" 700V V(BR)DSS POWER MOS
- 65 kHz INTERNAL OSCILLATOR
- 2.5V ± 2% INTERNAL REFERENCE
- STANDBY MODE FOR HIGH EFFICIENCY AT LIGHT LOAD
- OVERCURRENT AND LATCHED OVERVOLTAGE PROTECTION
- NON DISSIPATIVE BUILT-IN START-UP CIRCUIT
- THERMAL SHUTDOWN WITH HYSTERESIS
- BROWNOUT PROTECTION (SMD PACKAGE ONLY)

### MAIN APPLICATIONS

- WALL PLUG POWER SUPPLIES UP TO 15 W
- AC-DC ADAPTERS
- AUXILIARY POWER SUPPLIES FOR:
  - CRT AND LCD MONITOR (BLUE AND CEL)
  - DESKTOP PC/SERVER
  - FAX, TV, LASER PRINTER

### TYPICAL APPLICATION CIRCUIT



- HOME APPLIANCES/LIGHTING
- LINE CARD, DC-DC CONVERTERS

### DESCRIPTION

The L6590 is a monolithic switching regulator designed in BCD OFF-LINE technology, able to operate with wide range input voltage and to deliver up to 15W output power. The internal power switch is a lateral power MOSFET with a typical  $R_{DS(on)}$  of 13 $\Omega$  and a  $V_{(BR)DSS}$  of 700V minimum.

## L6590

### DESCRIPTION (continued)

The MOSFET is source-grounded, thus it is possible to build flyback, boost and forward converters.

The device can work with secondary feedback and a  $2.5V \pm 2\%$  internal reference, in addition to a high gain error amplifier, makes possible also the use in applications either with primary feedback or not isolated.

The internal fixed oscillator frequency and the integrated non dissipative start-up generator minimize the external component count and power consumption.

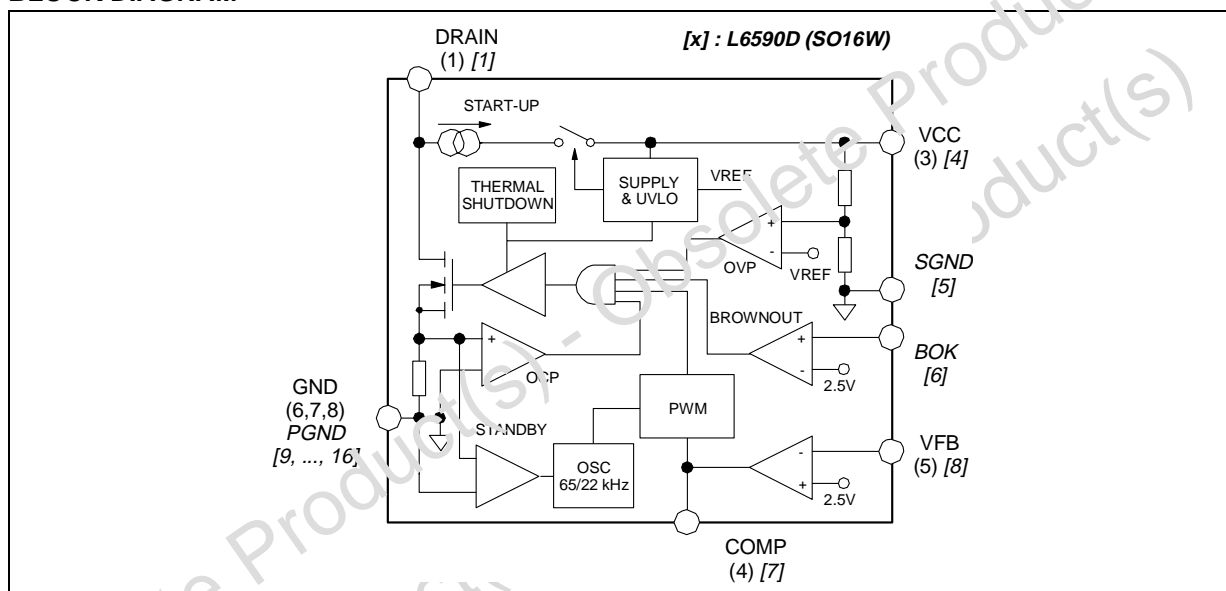
The device is equipped with a standby function that automatically reduces the oscillator frequency from 65 to 22 kHz under light load conditions to enhance

efficiency ( $P_{in} < 1W @ P_{out} = 0.5W$  with wide range mains).

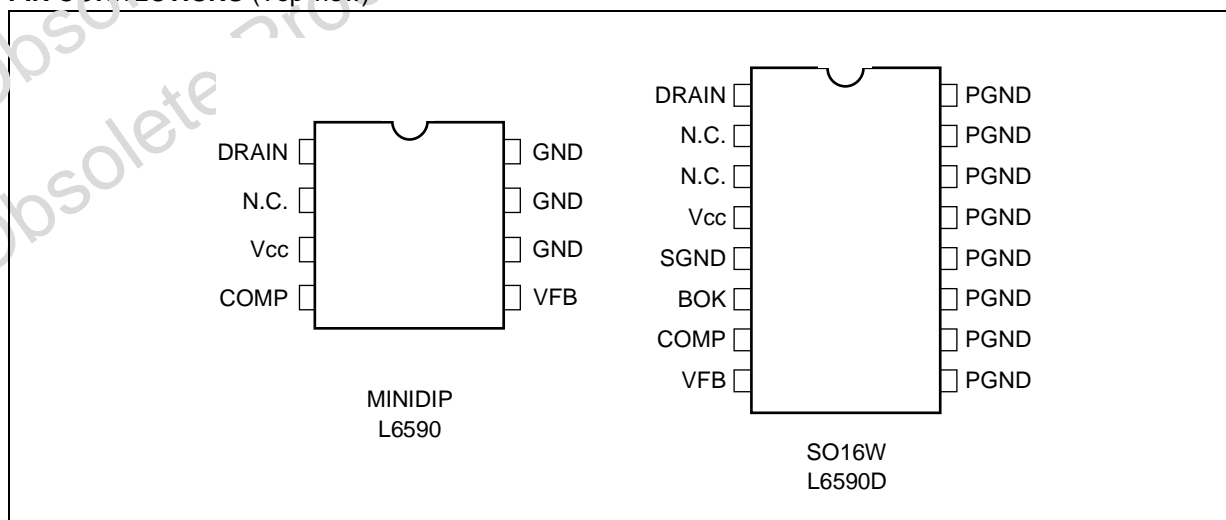
Internal protections like cycle-by-cycle current limiting, latched output overvoltage protection, mains undervoltage protection (SMD version only) and thermal shutdown generate a 'robust' design solution.

The IC uses a special leadframe with the ground pins (6, 7 and 8 in minidip, 9 to 16 in SO16W package) internally connected in order for heat to be easily removed from the silicon die. An heatsink can then be realized by simply making provision of few  $cm^2$  of copper on the PCB. Furthermore, the pin(s) close to the high-voltage one are not connected to ease compliance with safety distances on the PCB.

### BLOCK DIAGRAM



### PIN CONNECTIONS (Top view)



## PIN FUNCTIONS

| Pin#   |         | Name            | Description                                                                                                                                                                                                                                                                                                      |
|--------|---------|-----------------|------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| L6590  | L6590D  |                 |                                                                                                                                                                                                                                                                                                                  |
| 1      | 1       | DRAIN           | Drain connection of the internal power MOSFET. The internal high voltage start-up generator sinks current from this pin.                                                                                                                                                                                         |
| 2      | 2, 3    | N.C.            | Not internally connected. Provision for clearance on the PCB.                                                                                                                                                                                                                                                    |
| 3      | 4       | V <sub>CC</sub> | Supply pin of the IC. An electrolytic capacitor is connected between this pin and ground. The internal start-up generator charges the capacitor until the voltage reaches the start-up threshold. The PWM is stopped if the voltage at the pin exceeds a certain value.                                          |
| 4      | 7       | COMP            | Output of the Error Amplifier. Used for control loop compensation or to directly control PWM with an optocoupler.                                                                                                                                                                                                |
| 5      | 8       | VFB             | Inverting input of the Error Amplifier. The non-inverting one is internally connected to a 2.5V ± 2% reference. This pin can be grounded in some feedback schemes.                                                                                                                                               |
| 6 to 8 | -       | GND             | Connection of both the source of the internal MOSFET and the return of the bias current of the IC. Pins connected to the metal frame to facilitate heat dissipation.                                                                                                                                             |
| -      | 6       | BOK             | Brownout Protection. If the voltage applied to this pin is lower than 2.5V the PWM is disabled. This pin is typically used for sensing the input voltage of the converter through a resistor divider. If not used, the pin can be either left floating or connected to V <sub>CC</sub> through a 15 kΩ resistor. |
| -      | 5       | SGND            | Current return for the bias current of the IC.                                                                                                                                                                                                                                                                   |
| -      | 9 to 16 | PGND            | Connection of the source of the internal MOSFET. Pins connected to the metal frame to facilitate heat dissipation.                                                                                                                                                                                               |

## THERMAL DATA

| Symbol                | Parameter                                  | Minidip  | SO16W    | Unit |
|-----------------------|--------------------------------------------|----------|----------|------|
| R <sub>thj-amb</sub>  | Thermal Resistance Junction to ambient (*) | 35 to 60 | 40 to 65 | °C/W |
| R <sub>thj-pins</sub> | Thermal Resistance Junction to pins        | 15       | 20       | °C/W |

(\*) Value depending on PCB copper area and thickness.

## ABSOLUTE MAXIMUM RATINGS

| Symbol             | Parameter                                                                                                                   | Value       | Unit |
|--------------------|-----------------------------------------------------------------------------------------------------------------------------|-------------|------|
| V <sub>DS</sub>    | Drain Source Voltage                                                                                                        | -0.3 to 700 | V    |
| I <sub>d</sub>     | Drain Current                                                                                                               | 0.7         | A    |
| V <sub>CC</sub>    | IC Supply Voltage                                                                                                           | 18          | V    |
| I <sub>clamp</sub> | V <sub>CC</sub> Zener Current                                                                                               | 20          | mA   |
|                    | Error Amplifier Output Sink Current                                                                                         | 3           | mA   |
|                    | Voltage on Feedback Input                                                                                                   | 5           | V    |
|                    | BOK pin Sink Current                                                                                                        | 1           | mA   |
| P <sub>tot</sub>   | Power Dissipation at T <sub>amb</sub> < 50°C (Minidip and SO16W)<br>3 cm <sup>2</sup> , 2 oz copper dissipating area on PCB | 1.5         | W    |
| T <sub>j</sub>     | Operating Junction Temperature                                                                                              | -40 to 150  | °C   |
| T <sub>stg</sub>   | Storage Temperature                                                                                                         | -40 to 150  | °C   |

**ELECTRICAL CHARACTERISTICS** ( $T_j = -25$  to  $125^\circ\text{C}$ ,  $V_{CC} = 10\text{V}$ ; unless otherwise specified)

| Symbol | Parameter | Test Condition | Min. | Typ. | Max. | Unit |
|--------|-----------|----------------|------|------|------|------|
|--------|-----------|----------------|------|------|------|------|

**POWER SECTION**

|               |                                                                       |                                                    |     |    |     |               |
|---------------|-----------------------------------------------------------------------|----------------------------------------------------|-----|----|-----|---------------|
| $V_{(BR)DSS}$ | Drain Source Voltage                                                  | $I_d < 200 \mu\text{A}$ ; $T_j = 25^\circ\text{C}$ | 700 |    |     | V             |
| $I_{dss}$     | Off state drain current                                               | $V_{ds} = 560\text{V}$ ; $T_j = 125^\circ\text{C}$ |     |    | 200 | $\mu\text{A}$ |
| $R_{DS(on)}$  | Drain-to-Source on resistance<br>$R_{DS(on)}$ vs. $T_j$ : see fig. 20 | $I_d = 120\text{mA}$ ; $T_j = 25^\circ\text{C}$    |     | 13 | 16  | $\Omega$      |
|               |                                                                       | $I_d = 120\text{mA}$ ; $T_j = 125^\circ\text{C}$   |     | 23 | 28  |               |

**ERROR AMP SECTION**

|              |                          |                                                     |      |      |      |               |
|--------------|--------------------------|-----------------------------------------------------|------|------|------|---------------|
| $V_{FB}$     | Input Voltage            | $T_j = 25^\circ\text{C}$                            | 2.45 | 2.5  | 2.55 | V             |
|              |                          | $T_j = 125^\circ\text{C}$                           | 2.4  | 2.5  | 2.6  |               |
| $I_b$        | E/A Input Bias Current   | $V_{FB} = 0$ to $2.5\text{V}$                       |      | 0.3  | 5    | $\mu\text{A}$ |
| $A_{vol}$    | DC Gain                  | open loop                                           | 60   | 70   |      | dB            |
| B            | Unity Gain Bandwidth     |                                                     | 0.7  | 1    |      | MHz           |
| SVR          | Supply voltage Rejection | $f = 120\text{Hz}$                                  |      | 70   |      | dB            |
| $I_{sink}$   | Output Sink Current      | $V_{COMP} = 1\text{V}$                              |      | 1    |      | mA            |
| $I_{source}$ | Output Source Current    | $V_{COMP} = 3.5\text{V}$ ; $V_{FB} = 2\text{V}$     | -0.5 | -1   | -2.5 | mA            |
| $V_{COMPH}$  | Vout High                | $I_{source} = -0.5\text{mA}$ ; $V_{FB} = 2\text{V}$ | 3.8  | 4.50 |      | V             |
| $V_{COMPL}$  | Vout Low                 | $I_{sink} = 1\text{mA}$ ; $V_{FB} = 3\text{V}$      |      |      | 1    | V             |

**OSCILLATOR SECTION**

|           |                      |                          |    |    |    |     |
|-----------|----------------------|--------------------------|----|----|----|-----|
| $F_{osc}$ | Oscillator Frequency | $T_j = 25^\circ\text{C}$ | 58 | 65 | 72 | kHz |
|           |                      |                          | 52 | 65 | 74 |     |
| $D_{min}$ | Min. Duty Cycle      | $V_{COMP} = 1\text{V}$   |    |    | 0  | %   |
| $D_{max}$ | Max. Duty Cycle      | $V_{COMP} = 4\text{V}$   | 67 | 70 | 73 | %   |

**DEVICE OPERATION SECTION**

|               |                                     |                                                                                                                  |      |      |      |    |
|---------------|-------------------------------------|------------------------------------------------------------------------------------------------------------------|------|------|------|----|
| $I_{op}$      | Operating Supply Current            | $f_{sw} = F_{osc}$                                                                                               |      | 4.5  | 7    | mA |
| $I_Q$         | Quiescent Current                   | MOS disabled                                                                                                     |      | 3.5  | 6    | mA |
| $I_{charge}$  | $V_{CC}$ charge Current             | $V_{CC} = 0\text{V}$ to $V_{CCON} - 0.5\text{V}$ ;<br>$V_{ds} = 100$ to $400\text{V}$ ; $T_j = 25^\circ\text{C}$ | -3   | -4.5 | -7   | mA |
|               |                                     | $V_{CC} = 0\text{V}$ to $V_{CCON} - 0.5\text{V}$ ;<br>$V_{ds} = 100$ to $400\text{V}$                            | -2.5 | -4.5 | -7.5 | mA |
| $V_{CCclamp}$ | $V_{CC}$ clamp Voltage              | $I_{clamp} = 10\text{mA}$ (*)                                                                                    | 16.5 | 17   | 17.5 | V  |
| $V_{CCON}$    | Start Threshold voltage             | (*)                                                                                                              | 14   | 14.5 | 15   | V  |
| $V_{CCOFF}$   | Min operating voltage after Turn on | (*)                                                                                                              | 6    | 6.5  | 7    | V  |
| $V_{dsmin}$   | Drain start voltage                 |                                                                                                                  |      |      | 40   | V  |

## ELECTRICAL CHARACTERISTICS (continued)

| Symbol | Parameter | Test Condition | Min. | Typ. | Max. | Unit |
|--------|-----------|----------------|------|------|------|------|
|--------|-----------|----------------|------|------|------|------|

## CIRCUIT PROTECTIONS

|             |                              |                                      |     |      |     |    |
|-------------|------------------------------|--------------------------------------|-----|------|-----|----|
| $I_{pklim}$ | Pulse-by-pulse Current Limit | $di/dt = 120 \text{ mA}/\mu\text{s}$ | 550 | 625  | 700 | mA |
| OVP         | Overvoltage Protection       | $I_{CC} = 10 \text{ mA}$ (*)         | 16  | 16.5 | 17  | V  |
| LEB         | Masking Time                 | After MOSFET turn-on (**)            |     | 120  |     | ns |

## STANDBY SECTION

|            |                                           |                                       |    |     |    |     |
|------------|-------------------------------------------|---------------------------------------|----|-----|----|-----|
| $F_{SB}$   | Oscillator Frequency                      |                                       | 19 | 22  | 25 | kHz |
| $I_{pkSB}$ | Peak switch current for Standby Operation | Transition from $F_{osc}$ to $F_{SB}$ |    | 80  |    | mA  |
| $I_{pkNO}$ | Peak switch current for Normal Operation  | Transition from $F_{SB}$ to $F_{osc}$ |    | 190 |    | mA  |

## BROWNOUT PROTECTION (L6590D only)

|           |                    |                                  |     |     |     |               |
|-----------|--------------------|----------------------------------|-----|-----|-----|---------------|
| $V_{th}$  | Threshold Voltage  | Voltage either rising or falling | 2.4 | 2.5 | 2.6 | V             |
| $I_{Hys}$ | Current Hysteresis | $V_{pin} = 3V$                   | -30 | -50 | -70 | $\mu\text{A}$ |
| $V_{CL}$  | Clamp Voltage      | $I_{pin} = 0.5 \text{ mA}$       | 5.6 | 6.4 | 7.2 | V             |

## THERMAL SHUTDOWN (\*\*\*)

|  |            |  |     |     |  |                    |
|--|------------|--|-----|-----|--|--------------------|
|  | Threshold  |  | 150 | 165 |  | $^{\circ}\text{C}$ |
|  | Hysteresis |  |     | 40  |  | $^{\circ}\text{C}$ |

(\*) Parameters tracking one the other

(\*\*) Parameter guaranteed by design, not tested in production

(\*\*\*) Parameters guaranteed by design, functionality tested in production

Figure 1. Start-up &amp; UVLO Thresholds

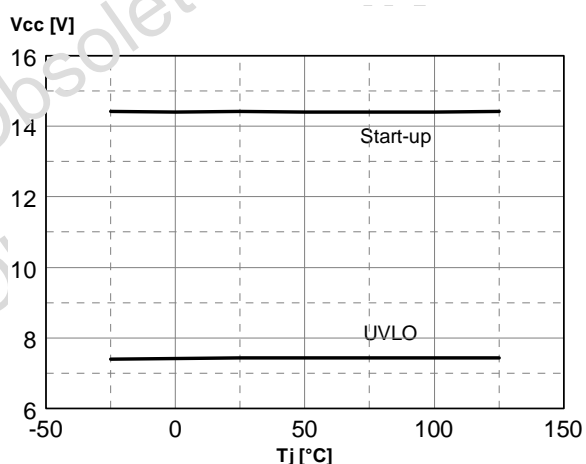


Figure 2. Start-up Current Generator

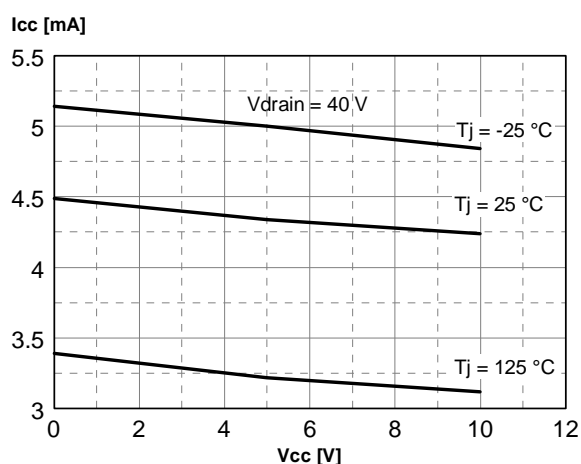


Figure 3. Start-up Current Generator

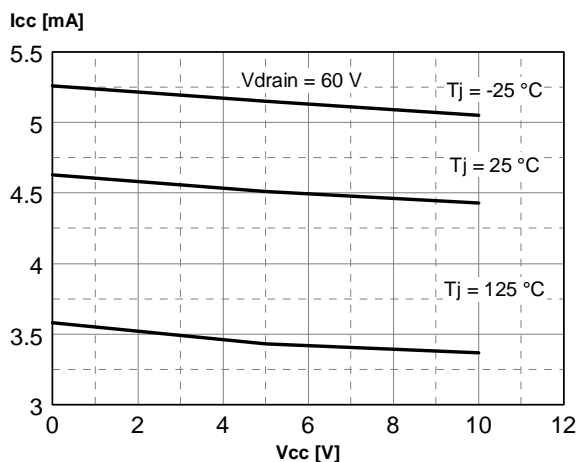


Figure 6. IC Operating Current

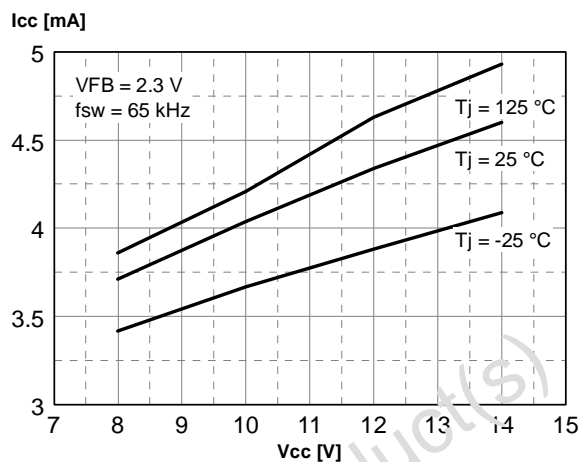


Figure 4. IC Consumption Before Start-up

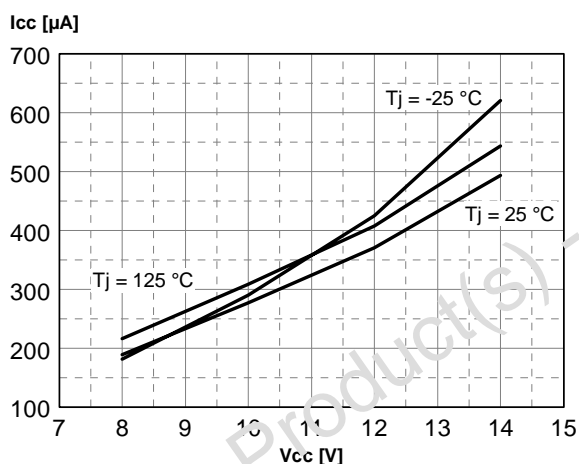


Figure 7. IC Operating Current

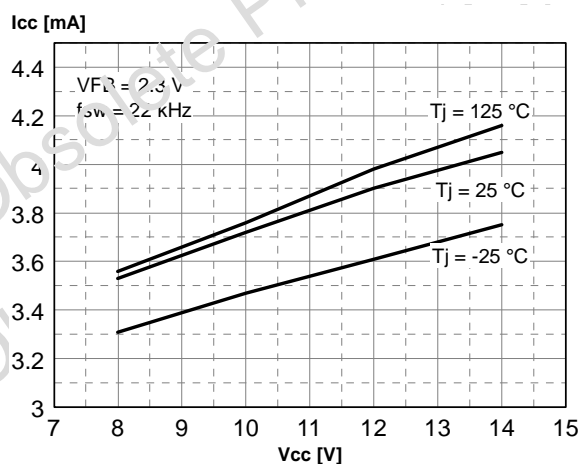


Figure 5. IC Quiescent Current

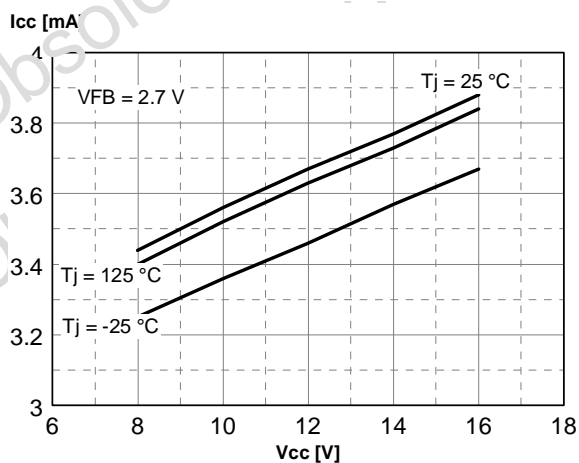


Figure 8. Switching Frequency vs. Temperature

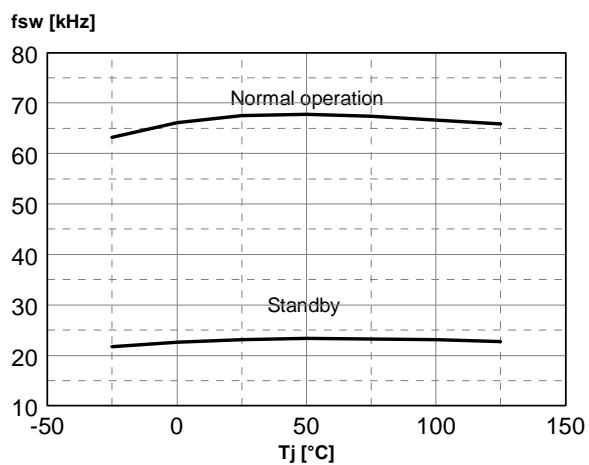


Figure 9. Vcc clamp vs. Temperature

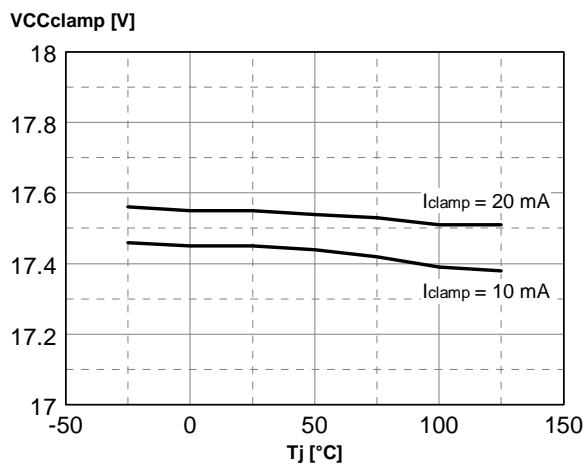


Figure 12. OCP threshold vs. Temperature

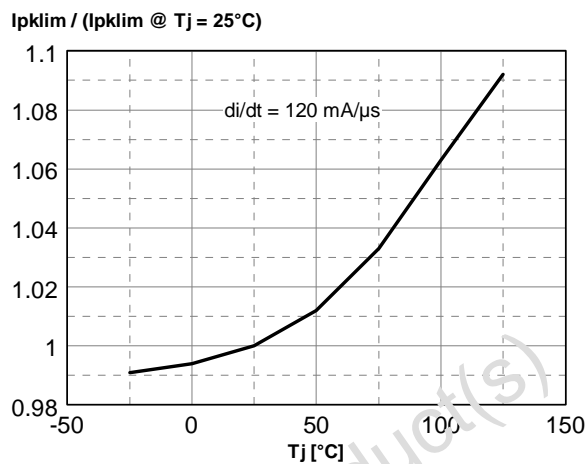


Figure 10. OVP Threshold vs. Temperature

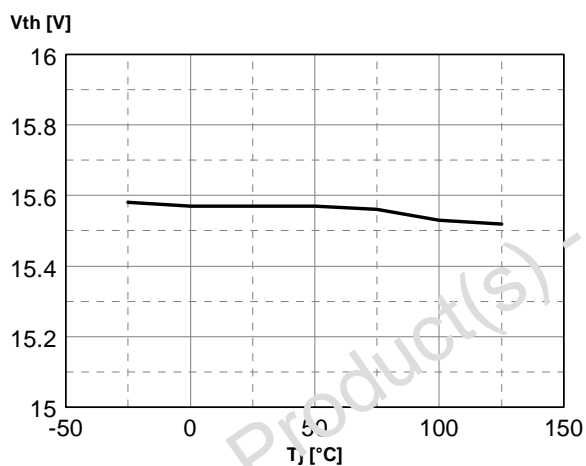


Figure 13. Internal E/A Reference Voltage

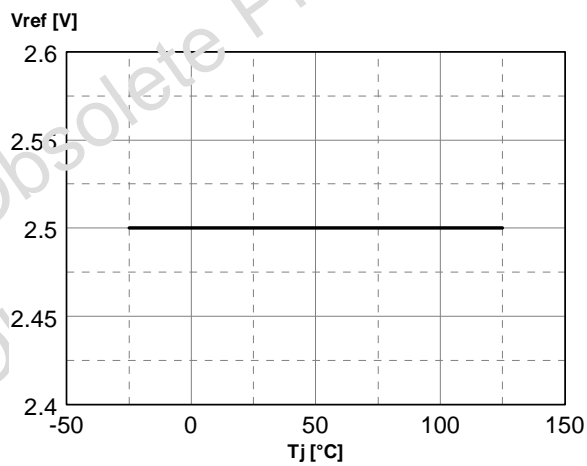


Figure 11. OCP Threshold vs. Current Slope

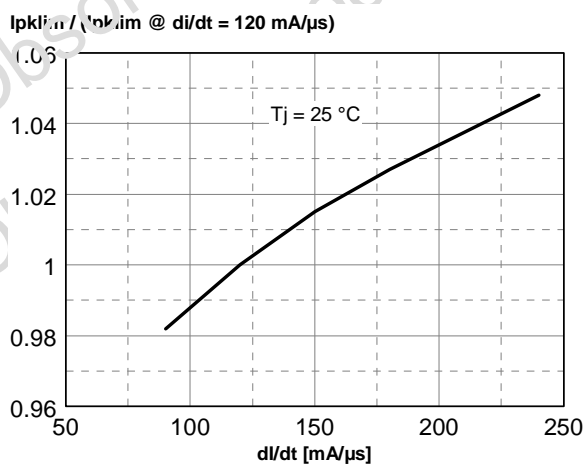


Figure 14. Error Amplifier Slew Rate

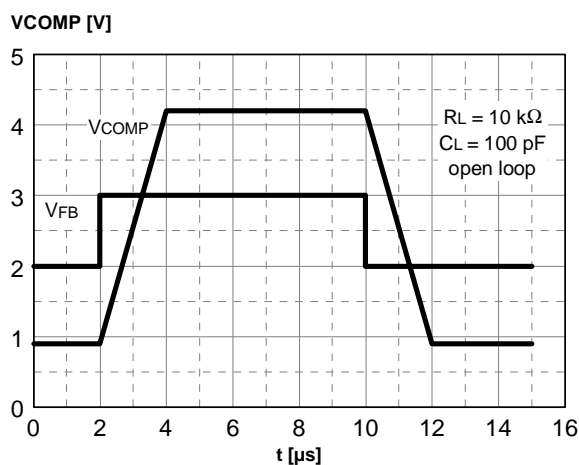


Figure 15. COMP pin Characteristic

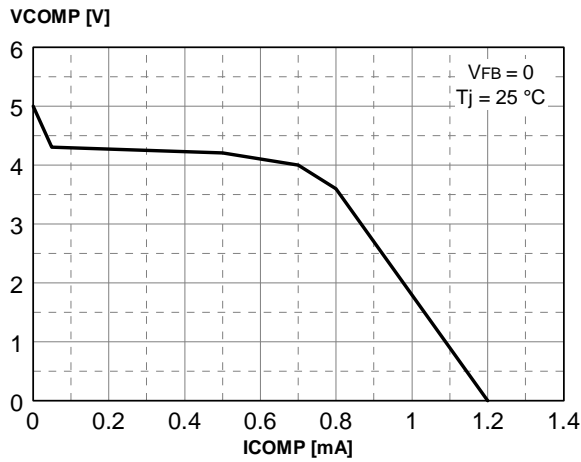


Figure 16. COMP pin Dynamic Resistance vs. Temperature

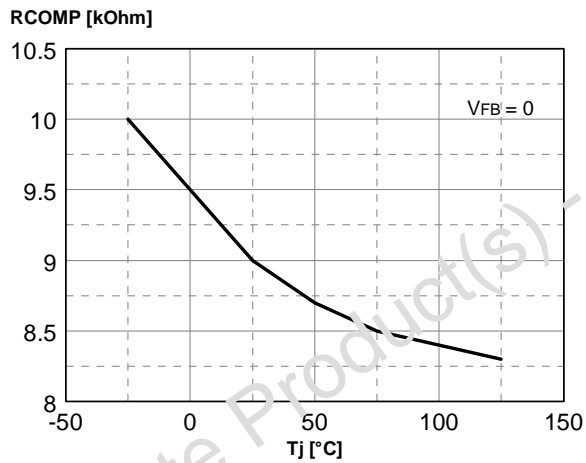


Figure 17. Error Amplifier Gain and Phase

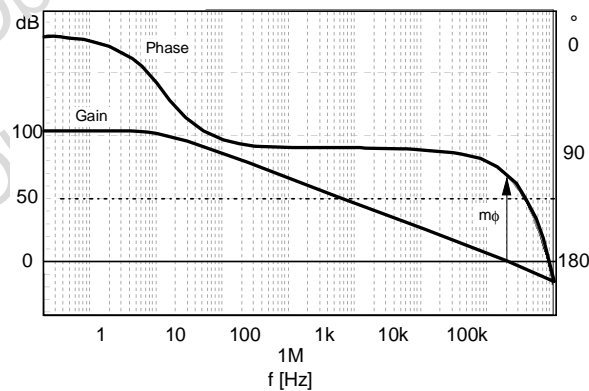


Figure 18. Breakdown Voltage vs. Temperature

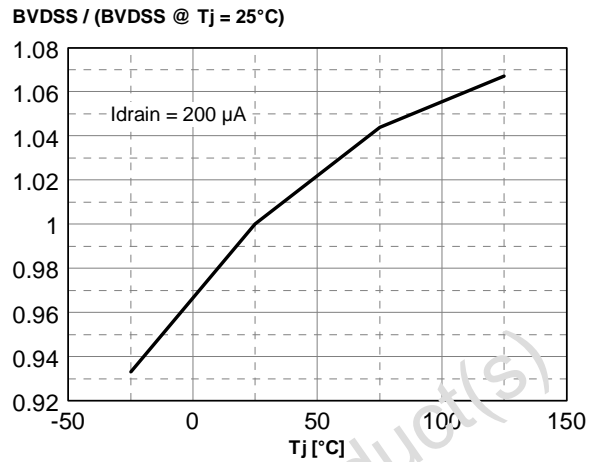


Figure 19. Drain Leakage vs. Drain Voltage

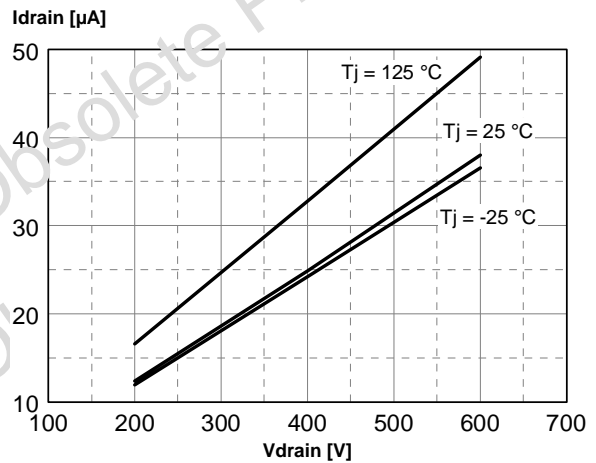


Figure 20. Rds(ON) vs. Temperature

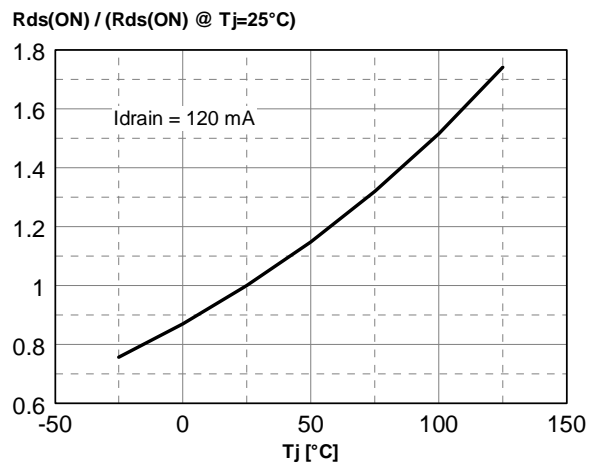




Figure 21. Rds(ON) vs. Idrain

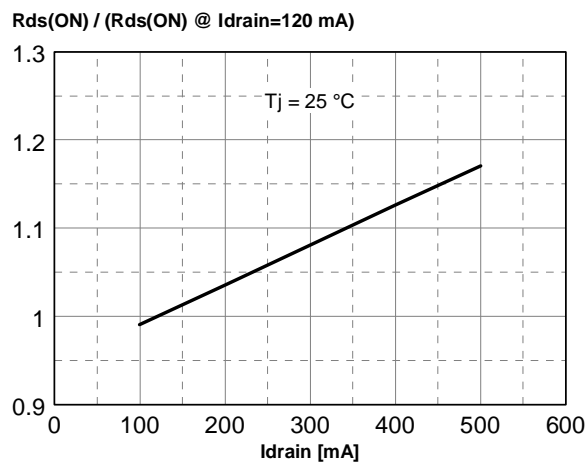


Figure 22. Coss vs. Drain Voltage

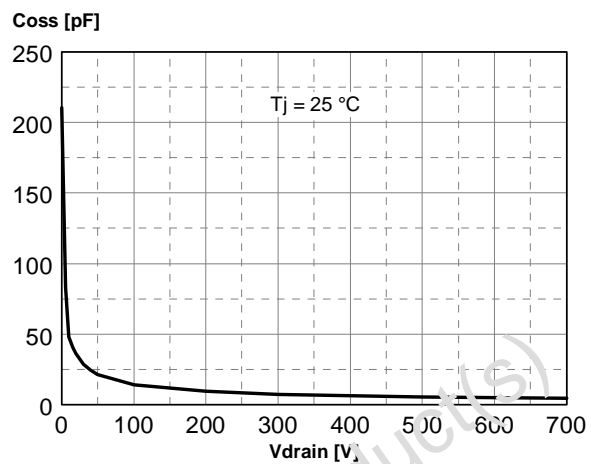


Figure 23. Standby Function Thresholds

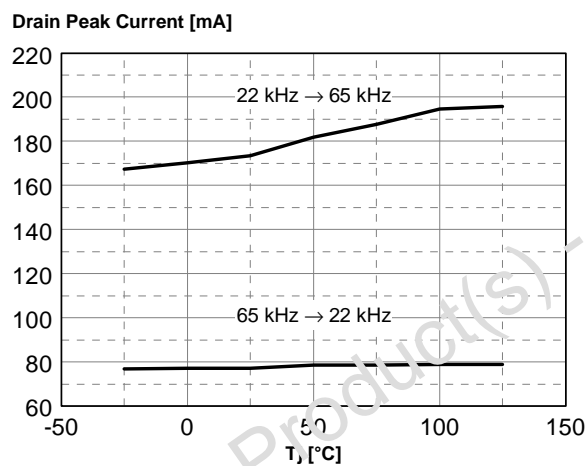


Figure 24. Test Board (1) with Primary Feedback: Electrical Schematic

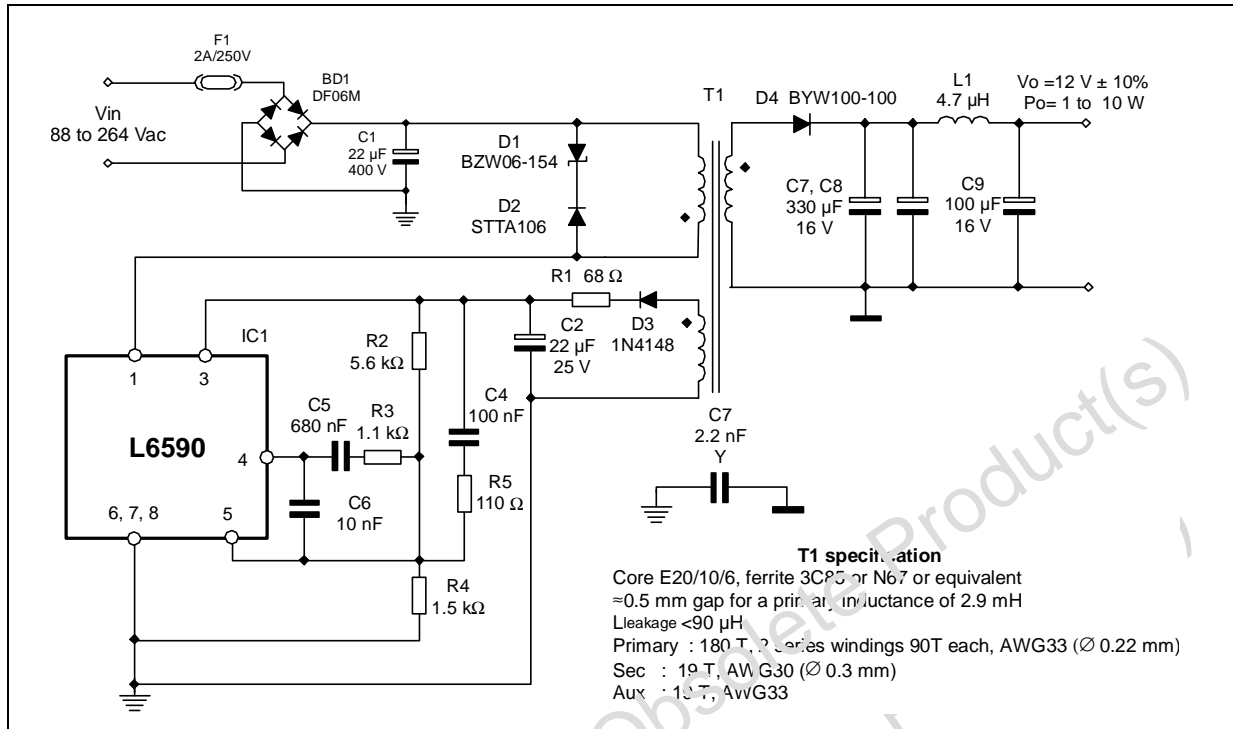


Figure 25. Test Board (1) Evaluation Data

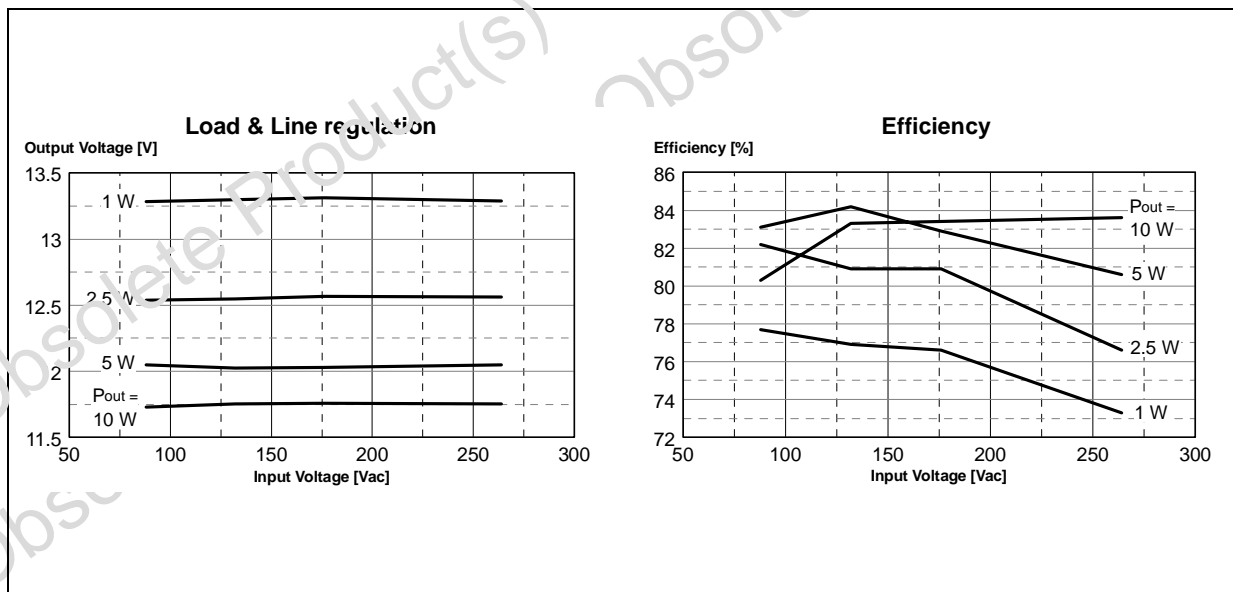


Figure 26. Test Board (1) Main Waveforms

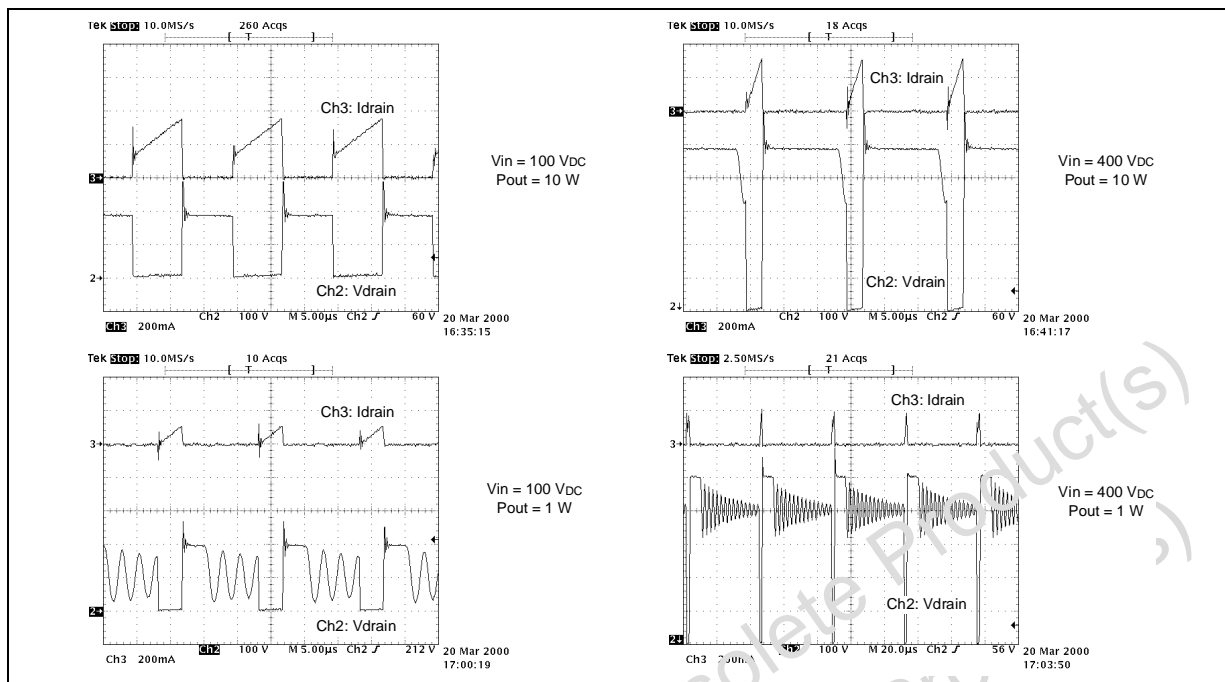


Figure 27. Test Board (2) with Secondary Feed'jack. Electrical Schematic

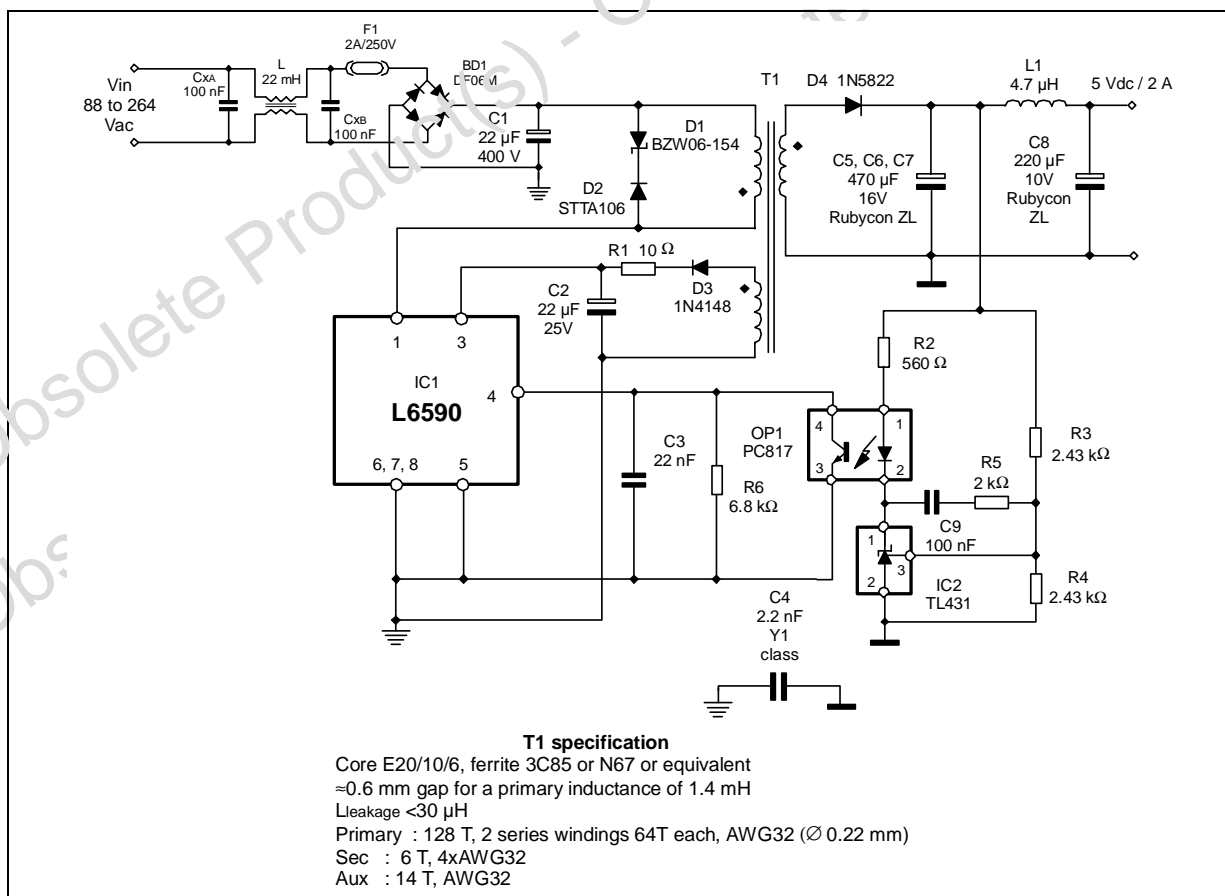


Figure 28. Test Board (2) evaluation data

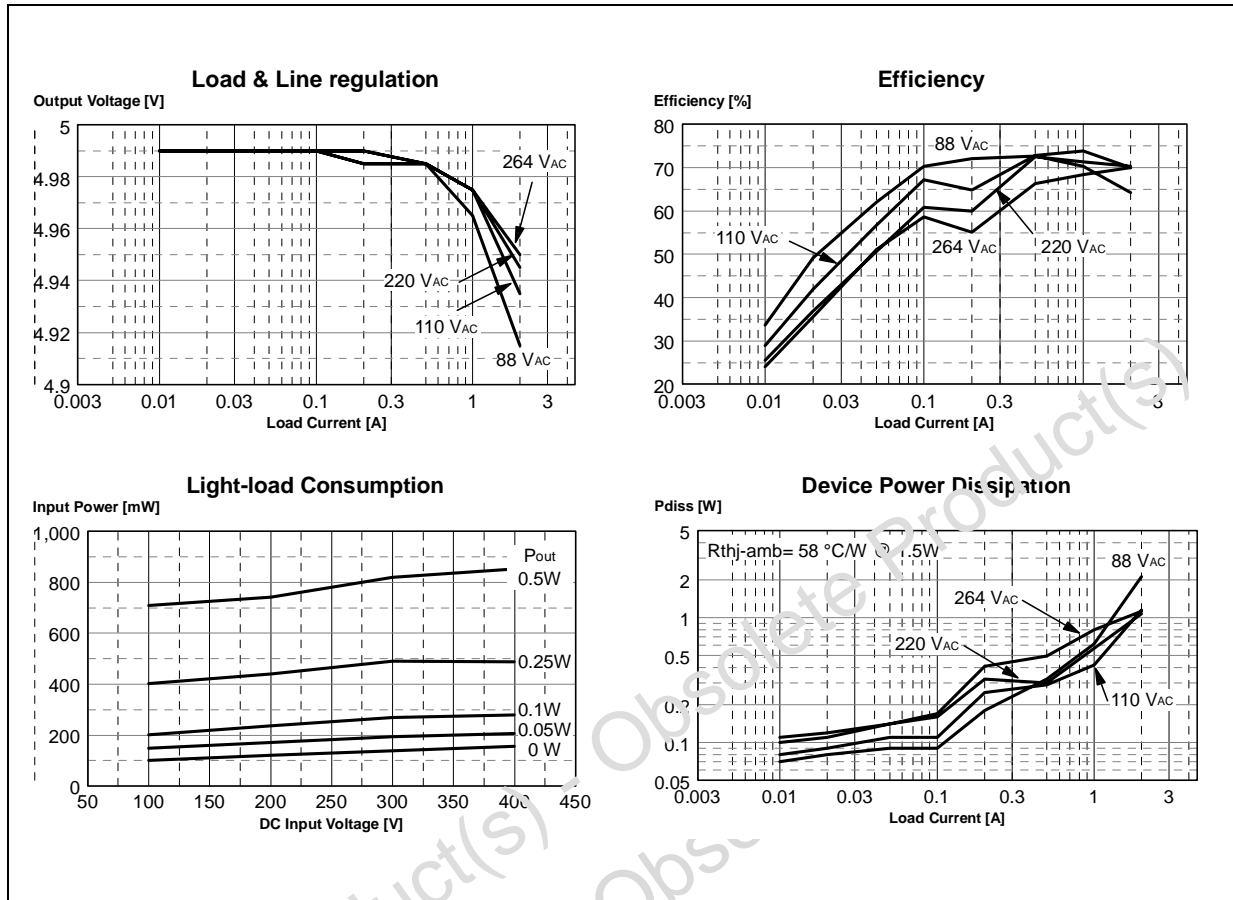


Figure 29. Test Board (2) EMI Characterization

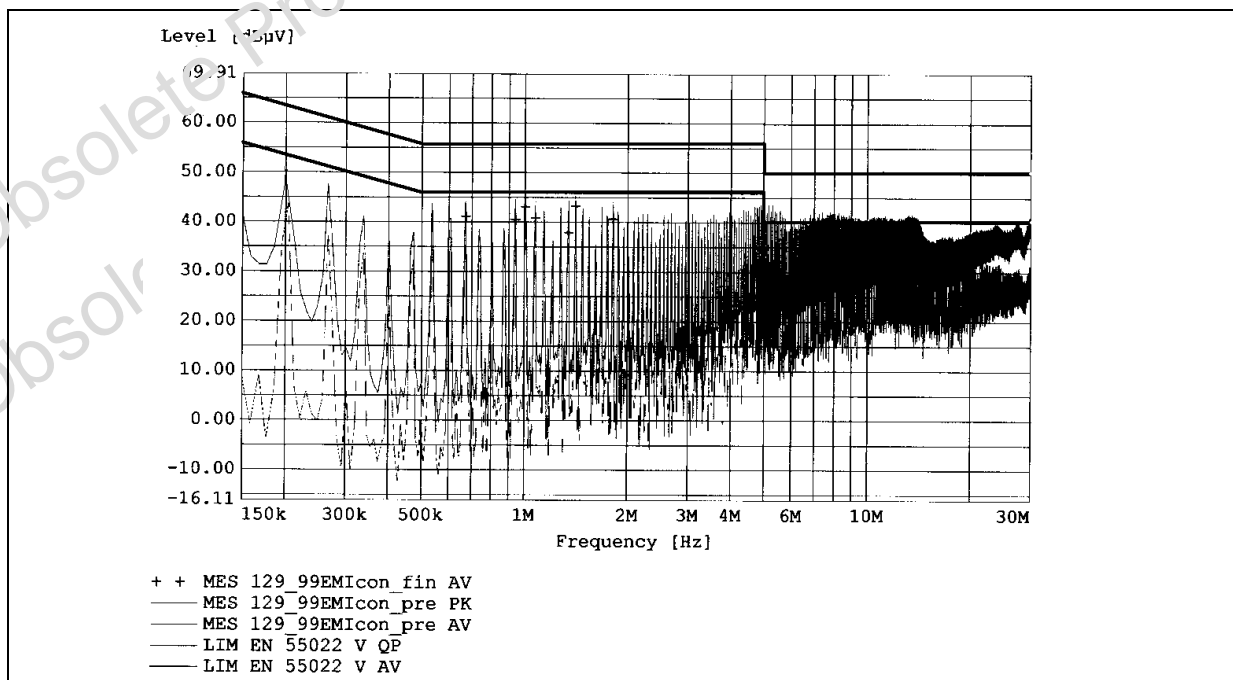


Figure 30. Test Board (2) Main Waveforms

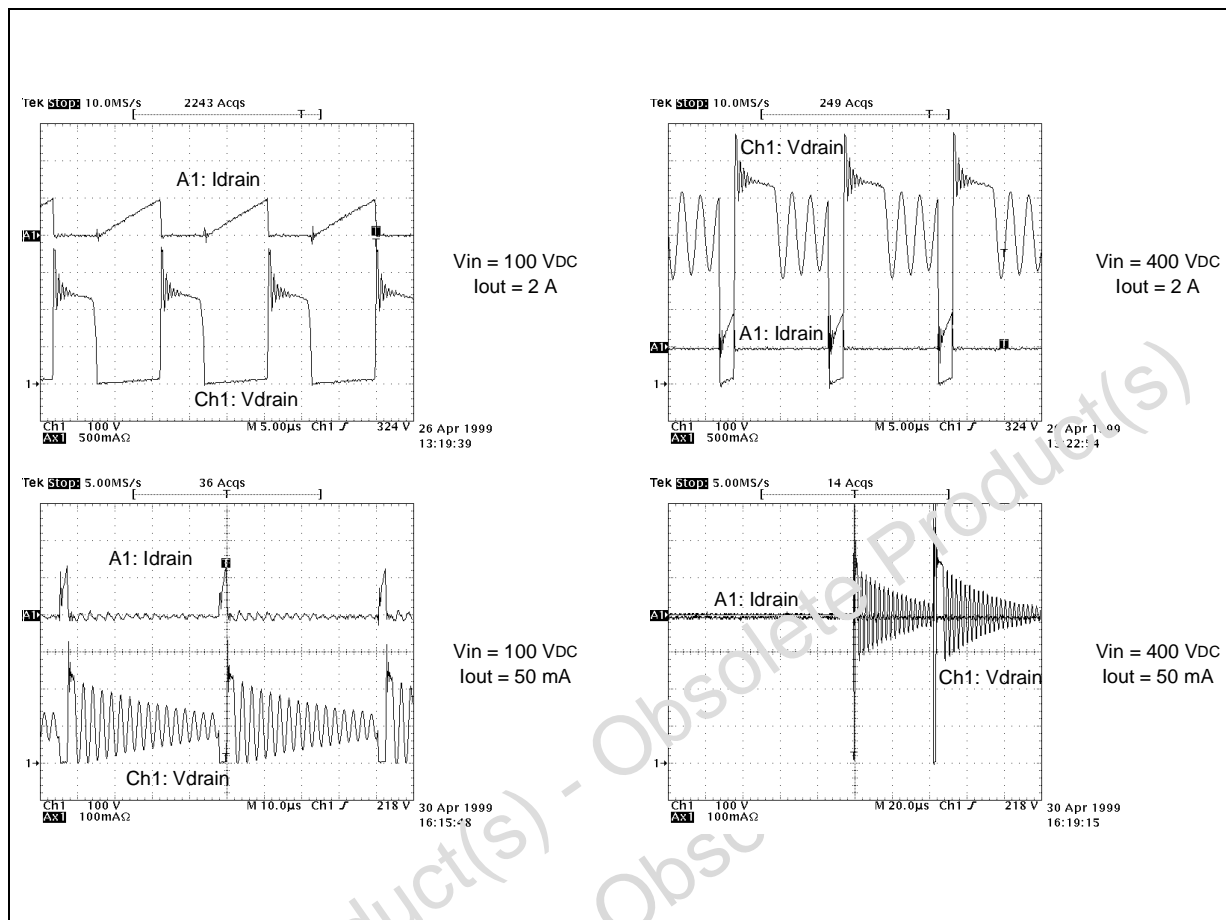
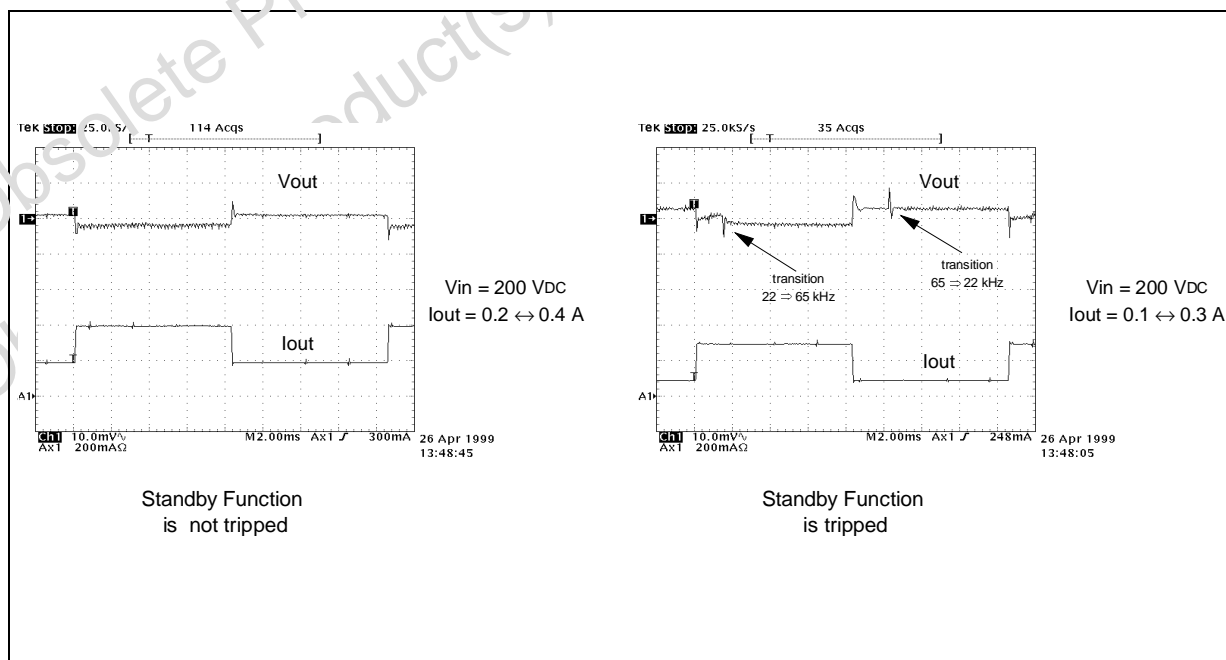


Figure 31. Test Board (2) Load Transient Response



## APPLICATION INFORMATION

In the following sections the functional blocks as well as the most important internal functions of the device will be described.

### Start-up Circuit

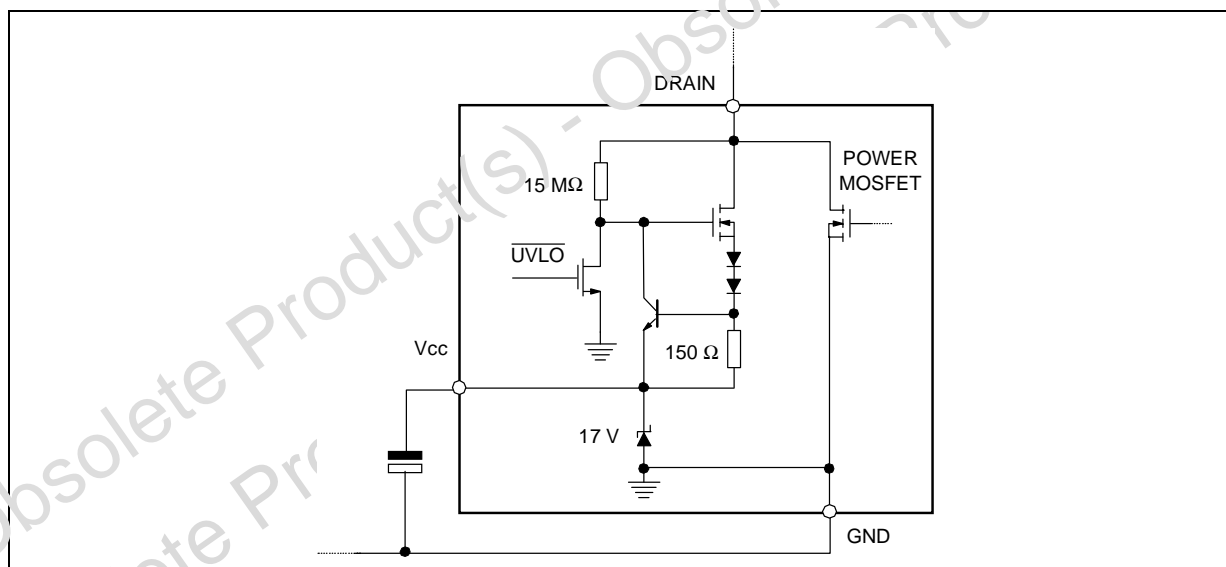
When power is first applied to the circuit and the voltage on the bulk capacitor is sufficiently high, an internal high-voltage current generator is sufficiently biased to start operating and drawing about 4.5 mA through the primary winding of the transformer and the drain pin. Most of this current charges the bypass capacitor connected between pin Vcc (3) and ground and makes its voltage rise linearly.

As the Vcc voltage reaches the start-up threshold (14.5V typ.) the chip, after resetting all its internal logic, starts operating, the internal power MOSFET is enabled to switch and the internal high-voltage generator is disconnected. The IC is powered by the energy stored in the Vcc capacitor until the self-supply circuit (typically an auxiliary winding of the transformer) develops a voltage high enough to sustain the operation.

As the IC is running, the supply voltage, typically generated by a self-supply winding, can range between 16 V (Overvoltage protection limit, see the relevant section) and 7 V, threshold of the Undervoltage Lockout. Below this value the device is switched off (and the internal start-up generator is activated). The two thresholds are in tracking.

The voltage on the Vcc pin is limited at safe values by a clamp circuit. Its 17V threshold tracks the Overvoltage protection threshold.

**Figure 32. Start-up circuit internal schematic**



### Power MOSFET and Gate Driver

The power switch is implemented with a lateral N-channel MOSFET having a  $V_{(BR)DSS}$  of 700V min. and a typical  $R_{DS(on)}$  of 13Ω. It has a SenseFET structure to allow a virtually lossless current sensing (used only for protection).

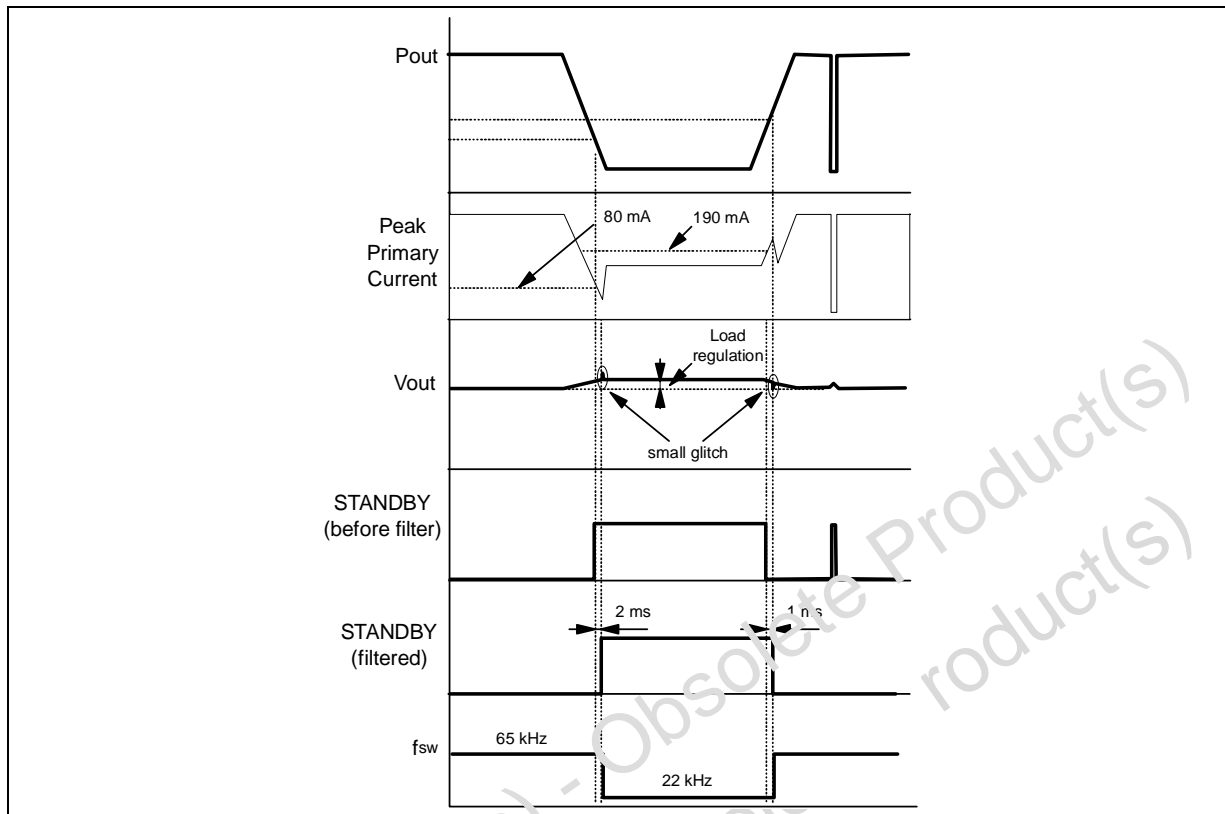
During operation in Discontinuous Conduction Mode at low mains the drain voltage is likely to go below ground. Any risk of injecting the substrate of the IC is prevented by an internal structure surrounding the switch.

The gate driver of the power MOSFET is designed to supply a controlled gate current during both turn-on and turn-off in order to minimize common mode EMI.

Under UVLO conditions an internal pull-down circuit holds the gate low in order to ensure that the power MOSFET cannot be turned on accidentally.



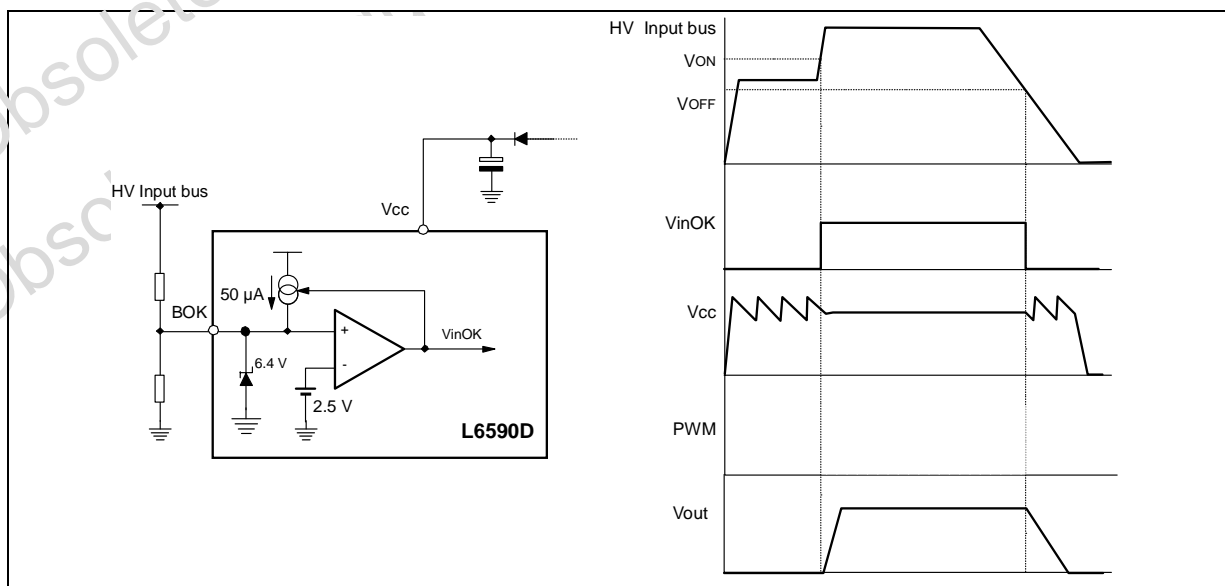
Figure 34. Standby Function timing diagram



**Brownout Protection (L6590D only)**

Brownout Protection is basically a not-latched device shutdown functionality. It will typically be used to detect a mains undervoltage (brownout). This condition may cause overheating of the primary power section due to an excess of RMS current

Figure 35. Brownout Protection Function internal schematic and timing diagram





Another problem is the spurious restarts that are likely to occur during converter power down if the input voltage decays slowly (e.g. with a large input bulk capacitor) and that cause the output voltage not to decay to zero monotonically.

Converter shutdown can be accomplished with the L6590D by means of an internal comparator that can be used to sense the voltage across the input bulk capacitor. This comparator is internally referenced to 2.5V and disables the PWM if the voltage applied at its non-inverting input, externally available, is below the reference. PWM operation is re-enabled as the voltage at the pin is more than 2.5V.

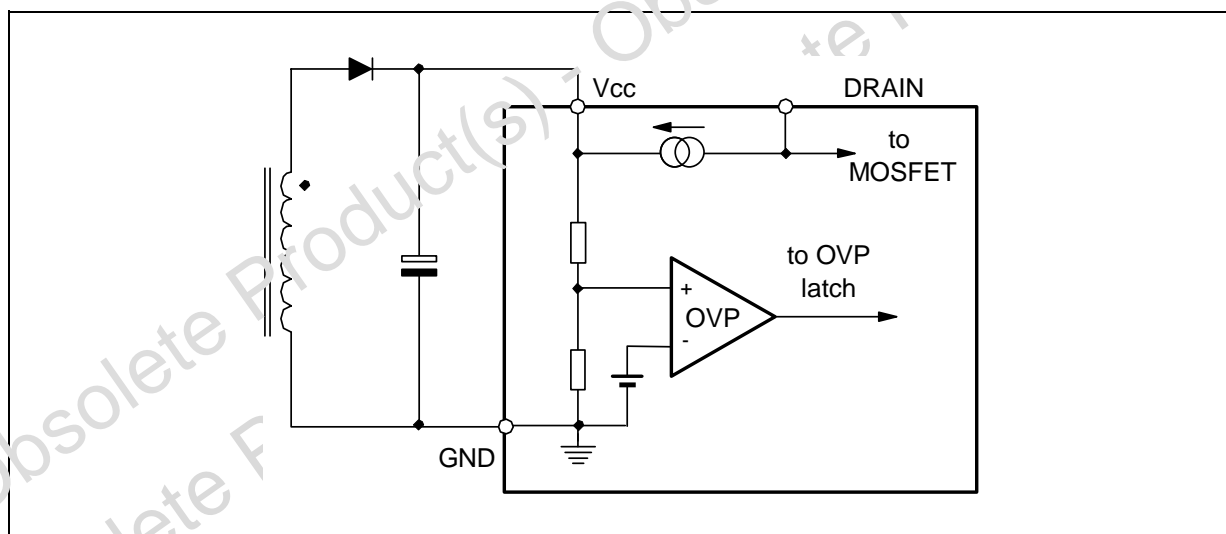
The brownout comparator is provided with current hysteresis instead of a more usual voltage hysteresis: an internal 50  $\mu$ A current generator is ON as long as the voltage applied at the non-inverting input exceeds 2.5V and is OFF if the voltage is below 2.5V. This approach provides an additional degree of freedom: it is possible to set the ON threshold and the OFF threshold separately by properly choosing the resistors of the external divider, which is not possible with voltage hysteresis.

### Overvoltage Protection

The IC incorporates an Overvoltage Protection (OVP) that can be particularly useful to protect the converter and the load against voltage feedback loop failures. This kind of failure causes the output voltage to rise with no control and easily leads to the destruction of the load and of the converter itself if not properly handled.

If such an event occurs, the voltage generated by the auxiliary winding that supplies the IC will fly up tracking the output voltage. An internal comparator continuously monitors the V<sub>CC</sub> voltage and stops the operation of the IC if the voltage exceeds 16.5 V. This condition is latched and maintained until the V<sub>CC</sub> voltage falls below the UVLO threshold. The converter will then operate intermittently.

Figure 36. OVP internal schematic



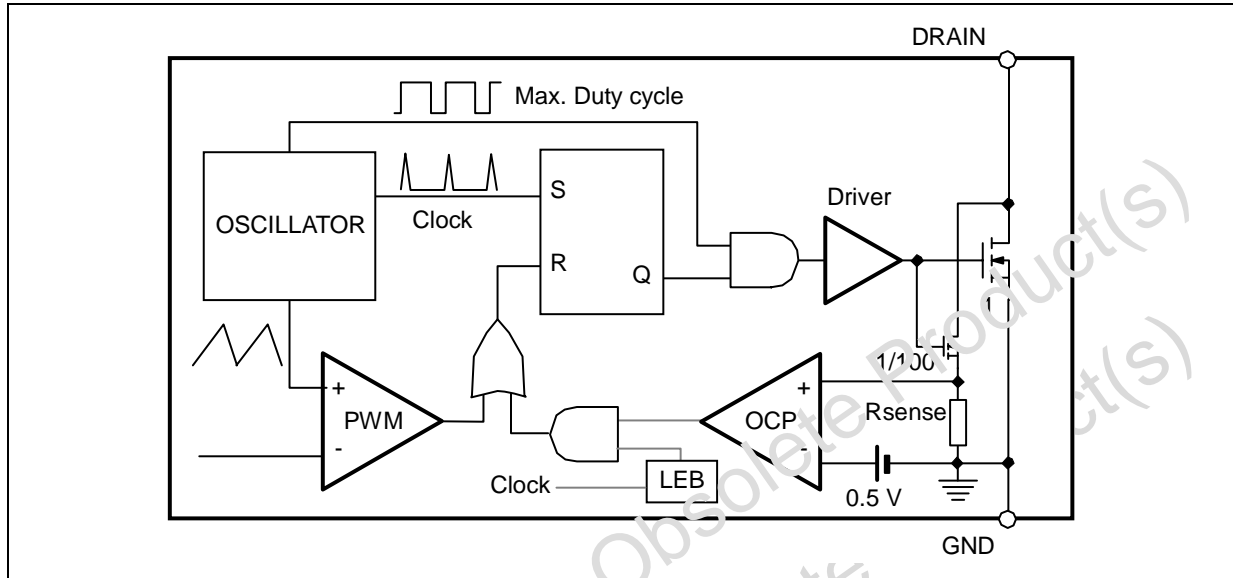
### Overcurrent Protection

The device uses pulse-by-pulse current limiting for Overcurrent Protection (OCP), in order to prevent overstress of the internal MOSFET: its current during the ON-time is monitored and, if it exceeds a determined value, the conduction is terminated immediately. The MOSFET will be turned on again in the subsequent switching cycle.

As previously mentioned, the internal powerMOSFET has a SenseFET structure: the source of a few cells are connected together and kept separate from the other source connections so as to realize a 1:100 current divider. The "sense" portion is connected to a ground referenced, sense resistor having a low thermal coefficient. The OCP comparator senses the voltage drop across the sense resistor and resets the PWM latch if the drop exceeds a threshold, thus turning off the MOSFET. In this way the overcurrent threshold is set at about 0.65 A (typical value).

At turn-on, there are large current spikes due to the discharge of parasitic capacitances and, in case of Continuous Conduction Mode operation, to secondary diode reverse recovery as well, which could falsely trigger the OCP comparator. To increase noise immunity the output of the OCP comparator is blanked for a short time (about 120 ns) just after the MOSFET is turned on, so that any disturbance within this time slot is rejected (Leading Edge Blanking).

**Figure 37. OCP internal schematic**



### Thermal Shutdown

Overheating of the device due to an excessive power throughput or insufficient heatsinking is avoided by the Thermal Shutdown function. A thermal sensor monitors the junction temperature close to the power MOSFET and, when the temperature exceeds 150 °C (min.), sets an alarm signal that stops the operation of the device. This is a not-latched function and the power MOSFET is re-enabled as the temperature falls about 40 °C.

APPLICATION IDEAS

Figure 38. 10W AC-DC adapter with no isolation

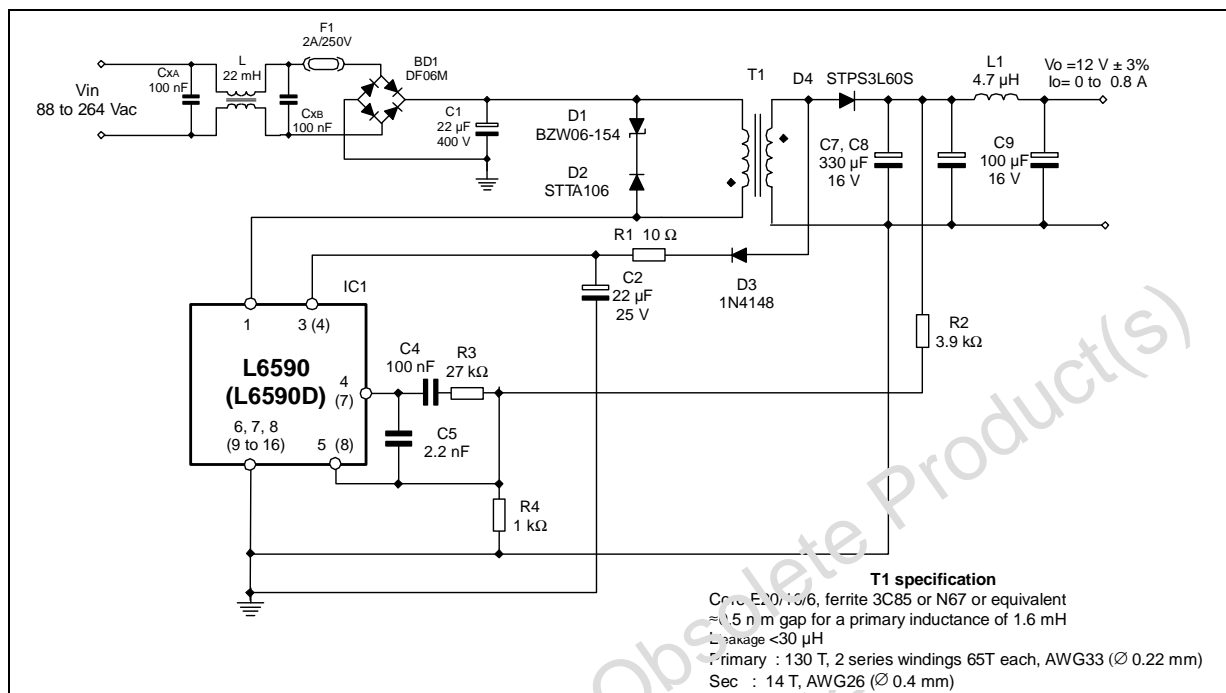


Figure 39. 15W Auxiliary SMPS for PC

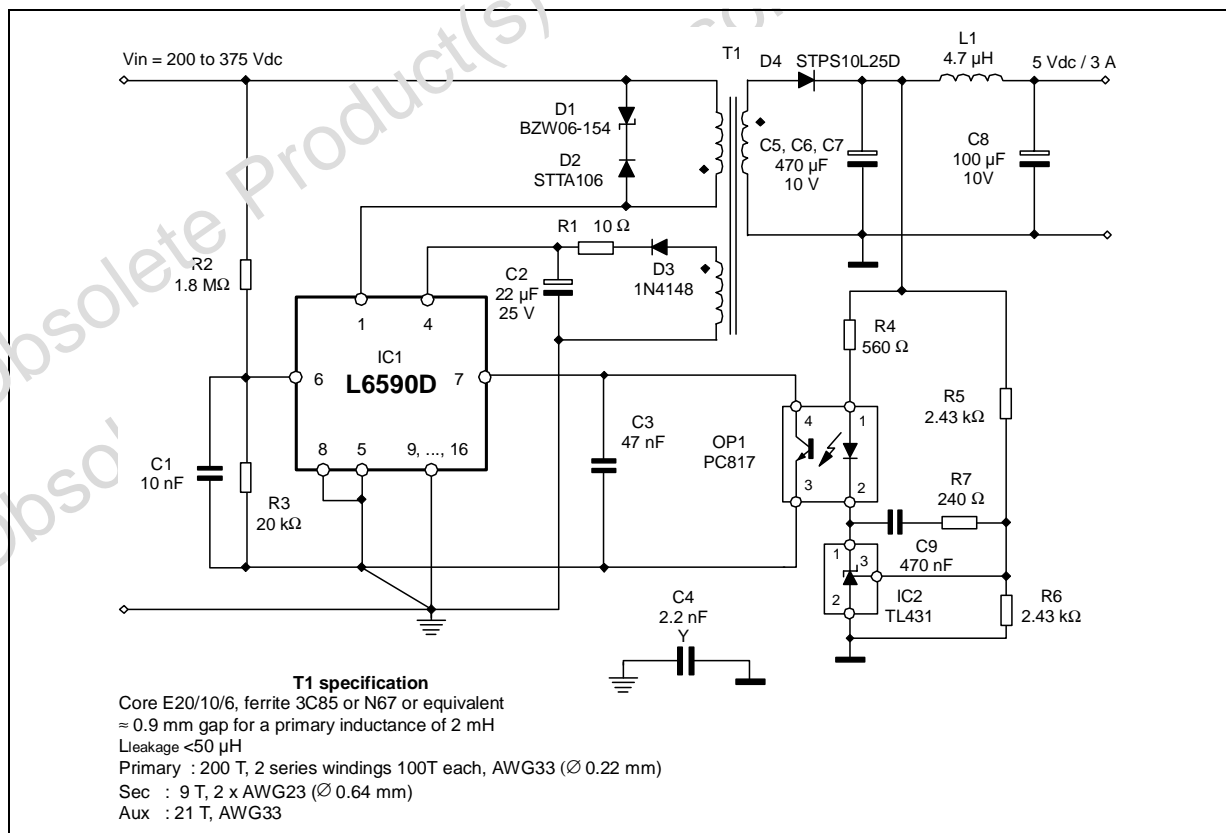
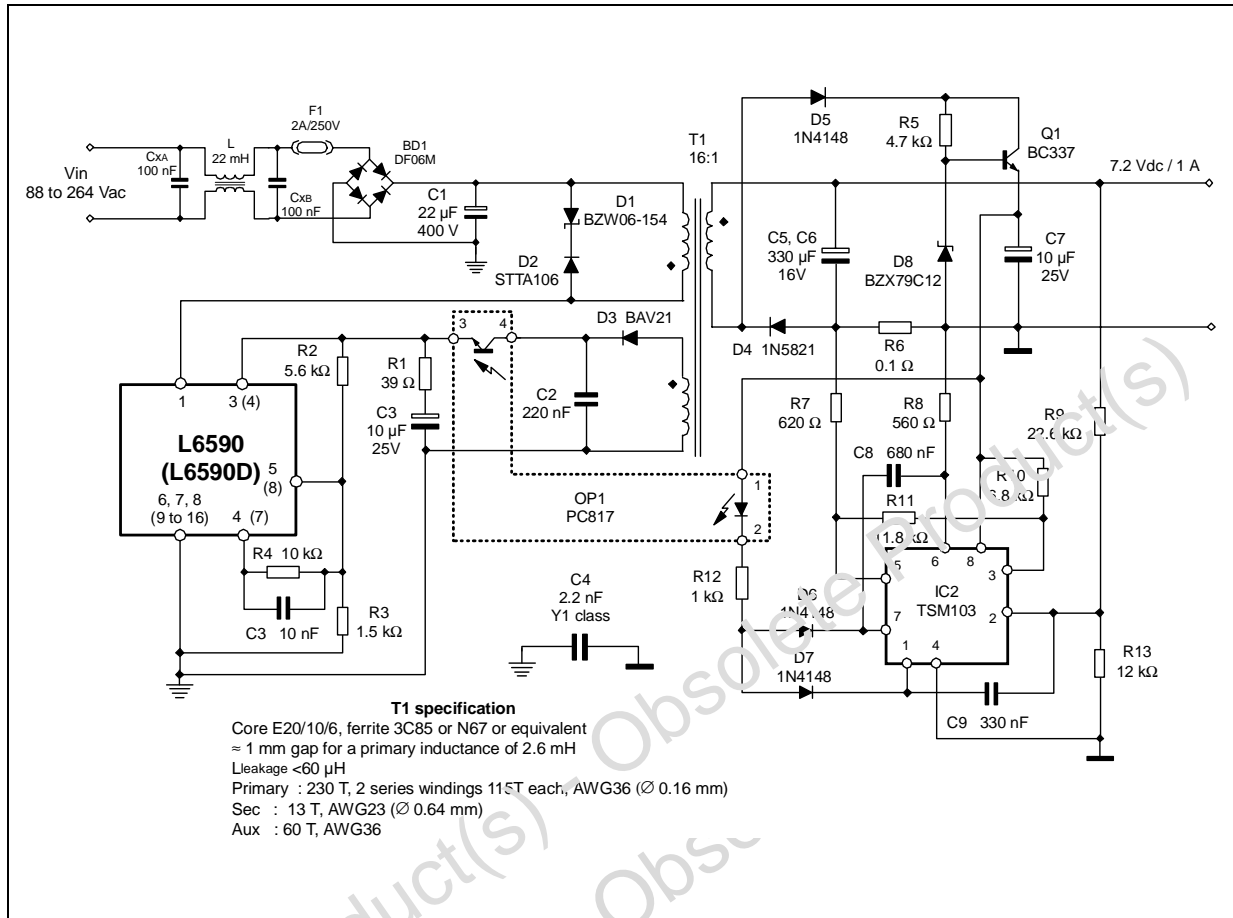


Figure 40. 7.2V/7W Battery Charger

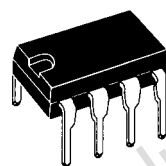


REFERENCES

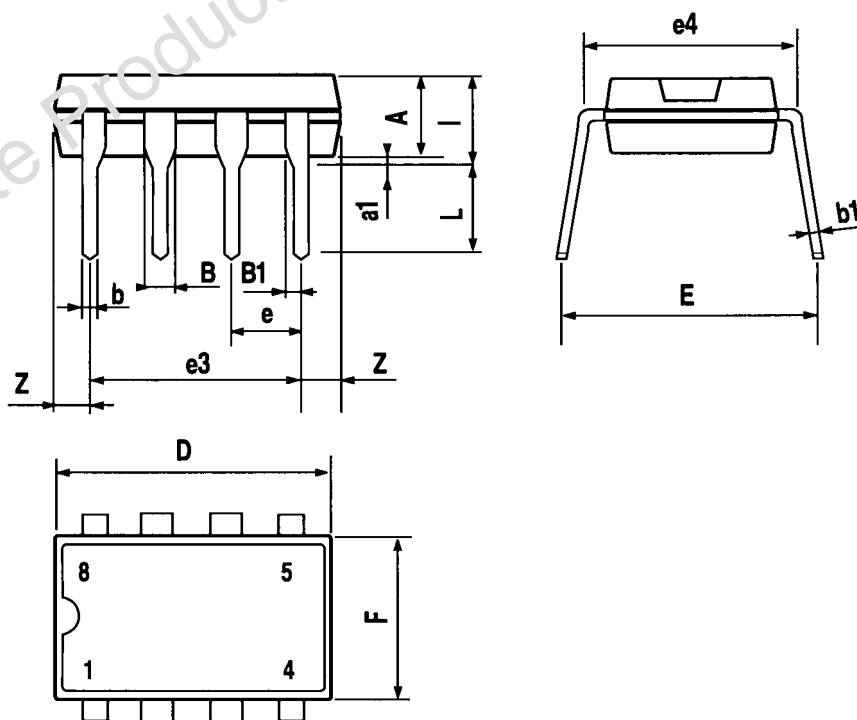
- [1] "Getting Familiar with the L6590 Family, High-voltage Fully Integrated Power Supply" (AN1261)
- [2] "Offline Flyback Converters Design Methodology with the L6590 Family" (AN1262)

| DIM. | mm    |      |       | inch  |       |       |
|------|-------|------|-------|-------|-------|-------|
|      | MIN.  | TYP. | MAX.  | MIN.  | TYP.  | MAX.  |
| A    |       | 3.32 |       |       | 0.131 |       |
| a1   | 0.51  |      |       | 0.020 |       |       |
| B    | 1.15  |      | 1.65  | 0.045 |       | 0.065 |
| b    | 0.356 |      | 0.55  | 0.014 |       | 0.022 |
| b1   | 0.204 |      | 0.304 | 0.008 |       | 0.012 |
| D    |       |      | 10.92 |       |       | 0.430 |
| E    | 7.95  |      | 9.75  | 0.313 |       | 0.384 |
| e    |       | 2.54 |       |       | 0.100 |       |
| e3   |       | 7.62 |       |       | 0.300 |       |
| e4   |       | 7.62 |       |       | 0.300 |       |
| F    |       |      | 6.6   |       |       | 0.260 |
| I    |       |      | 5.08  |       |       | 0.200 |
| L    | 3.18  |      | 3.81  | 0.125 |       | 0.150 |
| Z    |       |      | 1.52  |       |       | 0.060 |

## OUTLINE AND MECHANICAL DATA

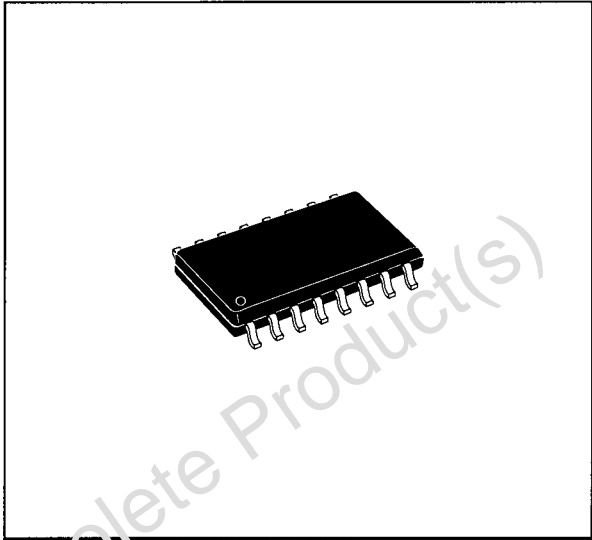


**Minidip**

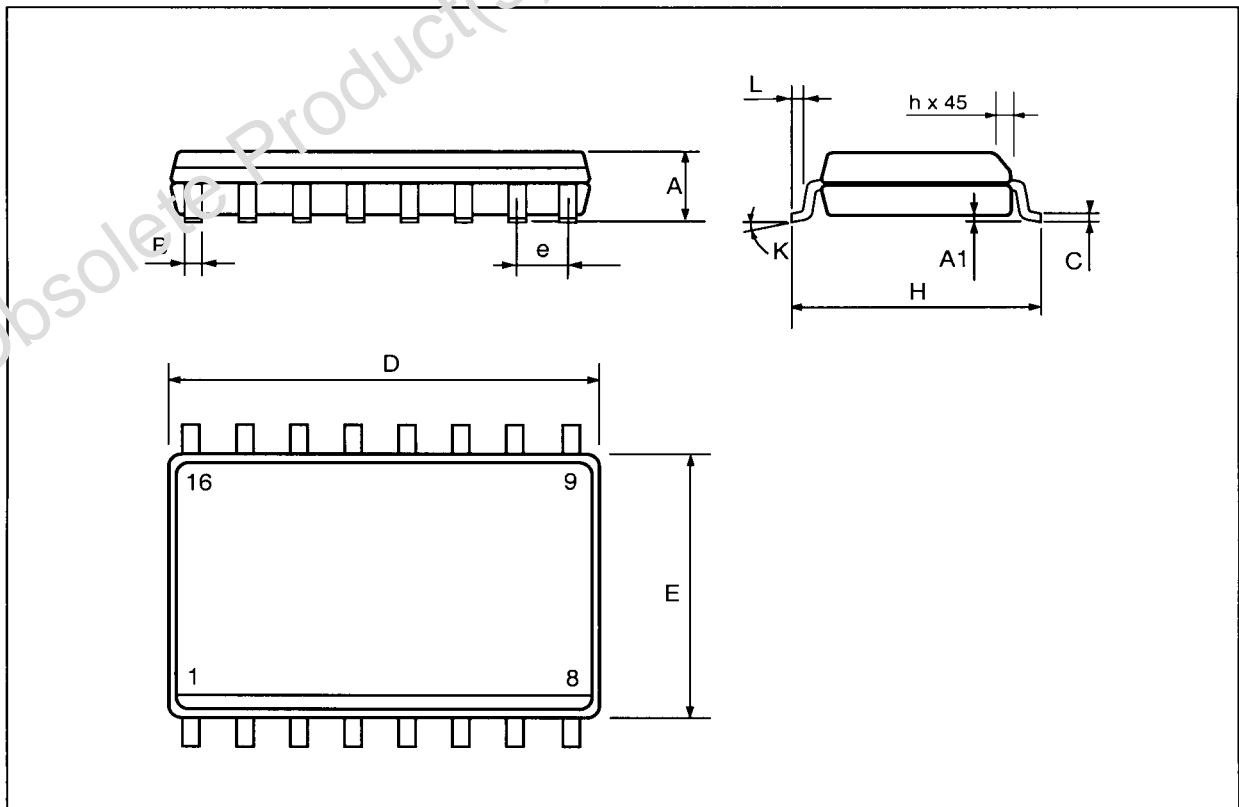


| DIM. | mm                 |      |       | inch  |       |       |
|------|--------------------|------|-------|-------|-------|-------|
|      | MIN.               | TYP. | MAX.  | MIN.  | TYP.  | MAX.  |
| A    | 2.35               |      | 2.65  | 0.093 |       | 0.104 |
| A1   | 0.1                |      | 0.3   | 0.004 |       | 0.012 |
| B    | 0.33               |      | 0.51  | 0.013 |       | 0.020 |
| C    | 0.23               |      | 0.32  | 0.009 |       | 0.013 |
| D    | 10.1               |      | 10.5  | 0.398 |       | 0.413 |
| E    | 7.4                |      | 7.6   | 0.291 |       | 0.299 |
| e    |                    | 1.27 |       |       | 0.050 |       |
| H    | 10                 |      | 10.65 | 0.394 |       | 0.419 |
| h    | 0.25               |      | 0.75  | 0.010 |       | 0.030 |
| L    | 0.4                |      | 1.27  | 0.016 |       | 0.050 |
| K    | 0° (min.)8° (max.) |      |       |       |       |       |

**OUTLINE AND MECHANICAL DATA**



**SO16 Wide**



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