## Single and dual PMOS high-side H-bridge



## Features

- Full path $R_{\text {DSON }}$ less than $540 \mathrm{~m} \Omega$
- Continuous load current > 3 A
- Operating battery supply voltage 5 V to 28 V
- Operating $\mathrm{V}_{\mathrm{DD}}$ supply voltage 4.5 V to 5.5 V
- All ECU internal pins can withstand up to 18 V
- Output switching frequency up to 11 kHz
- Monitoring of $\mathrm{V}_{\mathrm{DD}}$ supply voltage
- SPI programmable output current limitation from 5 A to 8.6 A (in 3 steps)
- Over temperature and short circuit protection
- Full diagnosis capability
- Fast switch-off open-drain input/output
- Current-monitoring with current feedback output signal CF
- SPI-interface for configuration and diagnosis
- Error history in second diagnosis register
- Two independent enable pins: "/ABE" and "DIS"
- Control of power stages by SPI or two input signals, PWM and DIR (configurable via SPI)
- Logic levels 5 V compatible
- Conformity to improved EMC requirements due to smart H -bridge switching


## Description

L9959S and L9959T are a single and dual integrated H -bridges for resistive and inductive loads featuring output current direction and supervising functions.

The PowerSSO24 houses one full H-Bridge, while the PowerSSO36 houses two H-Bridges that can work in parallel, through independent input driving commands.

Target application ranges from throttle control actuators to exhaust gas recirculation control valves in automotive domain to a more general use to drive DC and Stepper motors.

Table 1. Device summary

| Order code | Package | Packing |
| :---: | :---: | :---: |
| L9959S-TR | PowerSSO24 | Tape \& Reel |
| L9959T-TR | PowerSSO36 | Tape \& Reel |

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1 Block diagram

Figure 1. Block diagram


## 2 Pins description

Figure 2. PSSO24 pin connection (top view)


Figure 3. PSSO36 pin connection (top view)


### 2.1 Pin definitions and functions

Table 2. L9959S pinout

| Pin | Symbol | Function |
| :---: | :---: | :---: |
| 18 | VS | Power supply voltage for power stage outputs (external reverse protection required) |
| 2 | PGND | Power Ground |
| 16 | CAP3V3 | Pin for external capacitor: <br> This capacitor is used for the internal 3.3 V controller. A capacitance between 60 nF and 150 nF (typical 100 nF ) is required and has to be connected close to this pin with low inductance and resistance. |
| 17 | VDD | VDD Supply: 5 V Supply |
| 11 | GNDABE | Sense Ground for VDD monitoring |
| 15 | AGND | Device Ground |
| 3 | OUT1 | Bridge output 1 and 2: |
| 23 | OUT2 | The bridge outputs are built of a high-side p-channel and a low-side n-channel transistor. |
| 20 | DIR | Direction input: <br> The DIR pin controls the switch direction of OUT1 and OUT2. |
| 21 | PWM | PWM input: <br> The PWM input switches OUT1 and OUT2. |
| 19 | DIS | Disable input: <br> DIS switches OUT1 and OUT2 to tristate. |
| 10 | IABE | Bidirectional Ability/Enable Pin: <br> Open-Drain Output, which is pulled low in case of VDD over- and under-voltage. If the input is pulled to low, all output stages are switched off. |
| 9 | CF | Current Proportional Feedback output: <br> The CF pin provides in conjunction with an external resistor an output current, which is proportional to the H -Bridge current. |
| 5 | SCK | Serial clock input: <br> This input controls the internal shift register of the SPI. |
| 7 | SI | Slave in (Serial data input): <br> The input receives serial data from the microcontroller. |
| 8 | SO | Slave Out (Serial data output): <br> The diagnosis data is available via the SPI through this tristate-output. |
| 6 | SS | Slave Select input: <br> The serial data transfer between the device and the micro controller is enabled by pulling the input SS to low level. |
| EP | AGND | Exposed Pad: Connected to AGND. |

Table 3. L9959T (Two H-Bridge drivers in one package) pinout

| Pin | Symbol | Function |
| :---: | :---: | :---: |
| 13,14, | VS2 ${ }^{(1)}$ | Power supply voltage for power stage outputs (external reverse protection required): Important: For the capability of driving the full current at the outputs all pins of VS must be externally connected. |
| 31, 32 | VS1 ${ }^{(1)}$ |  |
| 1 | PGND1 ${ }^{(2)}$ | Ground: <br> Important: For the capability of driving the full current at the outputs, all ground pins must be externally connected. |
| 19 | PGND2 ${ }^{(2)}$ |  |
| 11, 29 | CAP3V3 | Pin for external capacitor: <br> This capacitor is used for the internal 3.3 V controller. A capacitance between 60 nF and 150 nF (typical 100 nF ) is required and has to be connected close to this pin with low inductance and resistance. |
| 12 | VDD1 ${ }^{(3)}$ | VDD Supply: 5V Supply. |
| 30 | VDD2 ${ }^{(3)}$ |  |
| 9 | GNDABE1 | Sense Ground for VDD monitoring |
| 27 | GNDABE2 |  |
| 10, 28 | AGND | Device Ground |
| 2 | OUT11 | Bridge output 11, 12, 21, and 22: <br> The bridge outputs are built of a high-side p-channel and a low-side n-channel transistor. The bridge outputs of chip 1 are OUT11 and OUT12, of chip 2 OUT21 and OUT22. |
| 36 | OUT12 |  |
| 20 | OUT21 |  |
| 18 | OUT22 |  |
| 34 | DIR1 | Direction input 1: <br> DIR1 pin controls the switch direction of OUT11 and OUT12. |
| 35 | PWM1 | PWM input 1: <br> PWM1 input switches OUT11 and OUT12. |
| 33 | DIS1 | Disable input 1: DIS1 switches OUT11 and OUT12 to tristate |
| 8 | /ABE1 | Bidirectional Ability/Enable Pin 1: <br> Open-Drain Output, which is pulled low in case of VDD over- and under-voltage. If the input is pulled to low, all output stages are switched off. /ABE1 belongs to chip 1. |
| 16 | DIR2 | Direction input 2: <br> DIR2 pin controls the switch direction of OUT21 and OUT22. |
| 17 | PWM2 | PWM input 2: <br> PWM1 input switches OUT21 and OUT22. |
| 15 | DIS2 | Disable input 2: <br> DIS2 switches OUT21 and OUT22 to tristate. |
| 26 | IABE2 | Bidirectional Ability/Enable Pin 2: <br> Open-Drain Output, which is pulled low in case of VDD over- and under-voltage. If the input is pulled to low, all output stages are switched off. /ABE2 belongs to chip 2. |
| 7 | CF1 | Current Proportional Feedback output: <br> The CF pin provides in conjunction with an external resistor an output current, which is proportional to the H-Bridge current. CF1 belongs to OUT11 and OUT12, CF2 to OUT21 and OUT22. |
| 25 | CF2 |  |

Table 3. L9959T (Two H-Bridge drivers in one package) pinout (continued)

| Pin | Symbol | Function |
| :---: | :---: | :--- |
| 3 | SCK1 | Serial clock input: <br> This input controls the internal shift register of the SPI. <br> SCK1 belongs to chip 1 and SCK2 to chip 2. |
| 21 | SCK2 | Slave in (Serial data input): <br> The input receives serial data from the microcontroller. <br> SI1 belongs to chip 1 and SI2 to chip 2. |
| 5 | SI1 | SI2 |
| 23 | SO1 | Slave Out (Serial data output): <br> The diagnosis data is available via the SPI through this tristate-output. <br> SO1 belongs to chip 1 and SO2 to chip 2. |
| 6 | SO2 | Slave Select input: <br> The serial data transfer between the device and the micro controller is enabled by pulling <br> the input SS to low level. <br> SS1 belongs to chip 1 and SS2 to chip 2. |
| 24 | SS1 | Sxposed PAD: connected to AGND <br> 22 |
| EP | AGND ${ }^{(4)}$ | Exper |

1. Pins 13 and 14 are referred to die2, whereas pins 31 and 32 are referred to die1.
2. Pins 1 is referred to die1, whereas 19 is referred to die2.
3. Pins 12 is referred to die2, whereas 30 is referred to die1.
4. Pins 10 is referred to die1, whereas 28 is referred to die2.

## 3 Electrical specifications

### 3.1 Absolute maximum ratings


#### Abstract

Warning: Stressing the device above the rating listed in the "Absolute maximum ratings" table may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the operating sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. Refer also to the STMicroelectronics SURE program and other relevant quality document.


Table 4. Absolute maximum ratings

| Symbol | Parameter / Test condition | Value [DC Voltage] | Unit |
| :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{Vs}}$ | DC supply voltage | -1.0 to +40 | V |
| $\mathrm{V}_{\mathrm{VDD}}$ | Stabilized supply voltage, logic supply | -0.3 to 18 | V |
| $\mathrm{V}_{3 \mathrm{~V} 3}$ | 3.3 V Controller output | -0.3 to 4.6 | V |
| $\mathrm{C}_{\mathrm{F}}$ | Current feedback output | -0.3 to 5 | V |
| $\begin{gathered} \mathrm{V}_{\mathrm{SI}}, \mathrm{~V}_{\mathrm{SCK}}, \mathrm{~V}_{\mathrm{SS}}, \mathrm{~V}_{\mathrm{SO}}, \\ \mathrm{~V}_{\mathrm{DIR}}, \mathrm{~V}_{\mathrm{PWM}}, \mathrm{~V}_{\mathrm{DIS}} \end{gathered}$ | Logic input / output voltage range | -0.3 to 18 | V |
| $\mathrm{V}_{\text {OUTn }}$ | Output voltage ( $\mathrm{n}=1,2$ or $11,12,21,22$ ); $\mathrm{V}_{\text {OUTn }}<\mathrm{V}_{\mathrm{S}}+1 \mathrm{~V}$ | -1.0 to 40 | V |
|  | Dynamic pulse / $\mathrm{t}<500 \mathrm{~ms}$; $\mathrm{V}_{\text {OUTn }}<\mathrm{V}_{\mathrm{S}}+2 \mathrm{~V}$ | -2.0 to 40 | V |
| $\mathrm{T}_{\mathrm{j}}$ | Operating junction temperature | -40 to 150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {stg }}$ | Storage temperature | -55 to 150 | ${ }^{\circ} \mathrm{C}$ |

### 3.2 ESD protection

Table 5. ESD protection

| Parameter | Value | Unit |
| :--- | :---: | :---: |
| All pins versus ground group (AGND, PGND1, PGND2, GND_ABE1, <br> GND_ABE2) | $\pm 2^{(1)}$ | kV |
| Power Output Pins: OUT1, OUT2 or OUT11, OUT12, OUT21, OUT22 versus <br> ground group (AGND, PGND1, PGND2, GND_ABE1, GND_ABE2) | $\pm 4^{(2)}$ | kV |

1. HBM according to MIL 883C, Method 3015.7 or EIA/JESD22-A114-A.
2. HBM with all unzipped pins grounded.

### 3.3 Thermal data

Table 6. Thermal data

| Symbol | Parameter | PSSO24 | PSSO36 | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\text {thj-case }}$ | Thermal resistance junction-to-case (max) | 2 | 3.5 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

### 3.4 Electrical characteristics

The voltages are referred to ground and currents are assumed positive, when the current flows into the pin. $4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{S}} \leq 18 \mathrm{~V}, 4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$; all outputs open; $\mathrm{T}_{\mathrm{j}}=-40^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$, unless otherwise specified.

Table 7. Supply

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{S}}$ | Operating voltage range | - | 4.5 | - | 28 | V |
| $\mathrm{I}_{\mathrm{vs}}$ | $\mathrm{V}_{\mathrm{S}}$ current consumption in active mode | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} ; \mathrm{V}_{\mathrm{S}}=5 \mathrm{~V} \text { and } \\ & \mathrm{V}_{\mathrm{S}}=18 \mathrm{~V} ; \\ & \text { Bridge disabled } \end{aligned}$ | - | - | 5 | mA |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} ; \mathrm{V}_{\mathrm{S}}=5 \mathrm{~V} \text { and } \\ & \mathrm{V}_{\mathrm{S}}=18 \mathrm{~V} ; \mathrm{f}_{\mathrm{OUT}}=2 \mathrm{kHz} ; \\ & \mathrm{l}_{\text {OUT }}=0 \mathrm{~A} \end{aligned}$ | - | - | 6 | mA |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} ; \mathrm{V}_{\mathrm{S}}=5 \mathrm{~V} \text { and } \\ & \mathrm{V}_{\mathrm{S}}=18 \mathrm{~V} ; \mathrm{f}_{\mathrm{OUT}}=10 \mathrm{kHz} ; \\ & \mathrm{I}_{\mathrm{OUT}}=0 \mathrm{~A} \end{aligned}$ | - | - | 14 | mA |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} ; \mathrm{V}_{\mathrm{S}}=28 \mathrm{~V} ; \\ & \mathrm{f}_{\mathrm{OUT}}=10 \mathrm{kHz} ; \mathrm{I}_{\mathrm{OUT}}=0 \mathrm{~A} \end{aligned}$ | - | - | 14 | mA |
| IvS(stby) | $\mathrm{V}_{\mathrm{S}}$ current consumption in passive mode | $V_{D D}=0 \mathrm{~V}$ | 0 | - | 2.5 | mA |
| $\mathrm{V}_{\text {VS_slew }}{ }^{(1)}$ | Slew rate on $\mathrm{V}_{\mathrm{S}}$ | - | - | - | 100 | V/us |
| $\mathrm{V}_{\text {VS_slew }}{ }^{(2)}$ | Slew rate on $\mathrm{V}_{\mathrm{S}}$ | - | - | - | 20 | V/ $/ \mathrm{s}$ |
| $V_{D D}$ | Operating voltage range | - | 4.5 | - | 5.5 | V |
| $\mathrm{I}_{\mathrm{VDD}}$ | $\mathrm{V}_{\mathrm{DD}}$ supply current | $\mathrm{V}_{\mathrm{S}}=18 \mathrm{~V} ; \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | - | - | 10 | mA |

1. No change of parameters for VDD-monitoring and in SPI logic
2. No change of parameters

Table 8. Power-on reset

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $V_{\text {DDRES }}$ | Reset active threshold | - | 2.8 | - | 3.4 | V |
| $\mathrm{~V}_{\text {DDPOR }}$ | Power-on reset threshold | - | 3.3 | - | 3.9 | V |
| $\mathrm{~V}_{\text {DDPORHYS }}$ | Power-on reset hysteresis | - | - | 600 | - | mV |
| $\mathrm{t}_{\text {POR }}$ | Power-on reset extension time | - | - | - | 1 | ms |

Table 9. $\mathrm{V}_{\mathrm{DD}}$ monitoring

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{D D}$ | $\mathrm{V}_{\mathrm{DD}}$ monitoring voltage range | - | $V_{\text {DDPOR }}$ | - | 18 | V |
| $\mathrm{V}_{\text {DD_THL }}$ | Under voltage threshold | $\mathrm{V}_{\mathrm{S}}=0 \mathrm{~V}$ | 4.2 | - | 4.5 | V |
| $\mathrm{V}_{\text {DD_THH }}$ | Over voltage threshold | $\mathrm{V}_{\mathrm{S}}=0 \mathrm{~V}$ | 5.25 | - | 5.5 | V |
| $\mathrm{t}_{\text {FIL_OFF }}$ | Switch-off filtering time | Guaranteed by scan. | 60 | - | 135 | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {FIL_ON }}$ | Switch-on filtering time |  | 60 | - | 135 | $\mu \mathrm{s}$ |
| $\mathrm{V}_{\text {TEST_THL }}$ | Under voltage test threshold | - | 5.25 | - | 5.5 | V |
| $\mathrm{V}_{\text {TEST_THH }}$ | Over voltage test threshold | - | 4.2 | - | 4.4 | V |
| V ${ }_{\text {DD_MR }}$ | Full $\mathrm{V}_{\text {DD }}$ supply range | - | -0.3 | - | 18 | V |
| V ${ }_{\text {DD_SLEW }}$ | $V_{\text {DD }}$ slew | - |  | - | 500 | $\mathrm{mV} / \mathrm{\mu s}$ |
| $\Delta \mathrm{V}_{\text {DD_THX }}$ | Threshold ( $\mathrm{V}_{\mathrm{DD} \text { _THH }}, \mathrm{V}_{\mathrm{DD}, \mathrm{THL}}$ ) shift during vs. īnverse current | - | -0.1 | - | 0.1 | V |
| $\mathrm{V}_{\text {ABE_INL }}$ | IABE input low-level | - | -0.3 | - | 1.65 | V |
| $\mathrm{V}_{\text {ABE_INH }}$ | IABE input high-level | - | 3.15 | - | 18 | V |
| $\begin{gathered} \mathrm{V}_{\text {ABE_INHY }} \\ \mathrm{s} \end{gathered}$ | IABE input hysteresis | - | 0.2 | - | 1.0 | V |
| $\mathrm{I}_{\text {ABE_IN }}$ | IABE input pull-down current | $0 \mathrm{~V}<\mathrm{V}_{\mathrm{ABE}}<1.5 \mathrm{~V}$ | 0 | - | 60 | $\mu \mathrm{A}$ |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{ABE}}=2.1 \mathrm{~V}, 5 \mathrm{~V}, 18 \mathrm{~V} ; \\ & \mathrm{V}_{\mathrm{S}}=18 \mathrm{~V} ; \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \\ & 18 \mathrm{~V} \end{aligned}$ | 20 | 40 | 60 | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\text {ABE_OUTL }}$ | IABE output low voltage | $\begin{aligned} & 2.5 \mathrm{~V}<\mathrm{V}_{\mathrm{DD}}<\mathrm{V}_{\mathrm{DD}-\mathrm{THL}} ; \\ & \mathrm{I}_{\mathrm{ABE} \text { _OUTL }}<2.5 \mathrm{~mA} \end{aligned}$ | 0 | - | 1.0 | V |
| $\mathrm{V}_{\text {ABE_OUTL }}$ | IABE output low voltage | $\begin{aligned} & 2 \times \mathrm{V}_{\mathrm{DD} \_\mathrm{THL}}<\mathrm{V}_{\mathrm{DD}}<18 \mathrm{~V} ; \\ & \mathrm{I}_{\mathrm{ABE} \text { _OULL }}<7.5 \mathrm{~mA} \end{aligned}$ | 0 | - | 1.2 | V |
| $\mathrm{V}_{\text {ABE_OUTL }}$ | IABE output passive low voltage | - | 0 | - | 1.2 | V |
| $\Delta^{\text {I }}$ ABE | $I_{\text {ABE }}$ Change during vs. inverse current | - | -100 | - | 100 | $\mu \mathrm{A}$ |

Table 10. Undervoltage shutdown

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| V UV_OFF | VS UV threshold | VS decreasing | 3.1 | 3.8 | 4.5 | V |
| $\mathrm{~V}_{\text {UV_ON }}$ | VS UV threshold | VS increasing | 3.3 | 4.0 | 4.7 | V |
| $\mathrm{~V}_{\text {UV_HYS }}$ | VS UV hysteresis | V $_{\text {UV_ON }}-V_{\text {UV_OFF }}$ | 0.1 | - | 1 | V |
| $\mathrm{t}_{\text {FUV }}$ | VS UV detection time | - | - | - | 1.5 | $\mu \mathrm{~s}$ |

### 3.5 Outputs OUT1 and OUT2

Table 11. On-resistance (4.5 V $<\mathrm{V}_{\mathrm{S}}<28 \mathrm{~V}$ )

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $r_{\text {ONVS }}$ OUT1,2 | On-resistance to supply | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} ; \mathrm{V}_{\mathrm{S}}=10 \mathrm{~V}, \\ & \mathrm{I}_{\mathrm{OUT} 1,2}=3 \mathrm{~A} \end{aligned}$ | - | - | 315 | $\mathrm{m} \Omega$ |
| rongnd OUT1,2 | On-resistance to PGND | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} ; \mathrm{V}_{\mathrm{S}}=10 \mathrm{~V}, \\ & \mathrm{I}_{\text {OUT1,2 }}=3 \mathrm{~A} \end{aligned}$ | - | - | 225 | $\mathrm{m} \Omega$ |
| $l_{\text {LEAK }}$ | Switched-off output current of OUT1,2 | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} ; \mathrm{V}_{\mathrm{S}}=13 \mathrm{~V} ; \\ & \mathrm{V}_{\mathrm{OUT}}=0 \mathrm{~V} \end{aligned}$ | -200 | - | - | $\mu \mathrm{A}$ |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} ; \mathrm{V}_{\mathrm{S}}=13 \mathrm{~V} ; \\ & \mathrm{V}_{\mathrm{OUT}}=\mathrm{V}_{\mathrm{S}} \end{aligned}$ | - | - | 200 | $\mu \mathrm{A}$ |

Table 12. Power output switching times ( $8 \mathrm{~V}<\mathrm{V}_{\mathrm{S}}<18 \mathrm{~V}$ )

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{d}} \mathrm{ON}$ | Output delay time driver on | - | - | - | 6 | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{d} \text { OFF }}$ | Output delay time driver off | - | - | - | 16 | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{d} \text { dis }}{ }^{(1)}$ | Disable delay time | Guaranteed through scan. | - | - | 12.5 | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {d pwon }}$ | Power-on delay time |  | - | - | 1 | ms |
| $\mathrm{t}_{\mathrm{d} \text { en }}$ | Enable delay time |  | - | - | 50 | $\mu \mathrm{s}$ |
| $\mathrm{dl}_{\text {Out }} / \mathrm{dt}$ | Current slew rate | - | - | 1.6 |  | A/ $\mu \mathrm{s}$ |
| $\mathrm{dV}_{\mathrm{OUT}_{(2)}} / \mathrm{dt}$ | Output rise/fall slew-rate high-side low selected with bit $\mathrm{SR}=0$ fast selected with bit $\mathrm{SR}=1$ | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} ; \mathrm{V}_{\mathrm{S}}=14 \mathrm{~V} \\ & \mathrm{R}_{\mathrm{LOAD1,2}}=2.6 \Omega\left(8 \mathrm{~V}_{\mathrm{S}}\right), \\ & 6 \Omega\left(18 \mathrm{~V}_{\mathrm{S}}\right) \end{aligned}$ | $\begin{gathered} 0.975 \\ 2.8 \end{gathered}$ | - | $\begin{gathered} 2.7 \\ 8 \end{gathered}$ | $\mathrm{V} / \mu \mathrm{s}$ |
| $\mathrm{dV}_{\text {rout }}^{\text {(2) }}$ Ldt | Output rise slew-rate low-side valid only after the toggling of DIR input |  | 0.975 | - | 2.7 | $\mathrm{V} / \mathrm{\mu s}$ |
| $\mathrm{dV}_{\text {fOUT(2) }} / \mathrm{dt}$ | Output fall slew-rate low-side |  | 2.5 | 4 | 8 | $\mathrm{V} / \mathrm{\mu s}$ |
| $\mathrm{f}_{\text {pwmmax }}$ | PWM input frequency | - | - | - | 11 | kHz |

1. Driven by IABE or DIS input.
2. The slew-rates (dVour/dt1) are defined by dV (voltage difference $20 \%-80 \%$ ) divided by the rise-/fall times ( $\mathrm{t}_{\mathrm{r}} / \mathrm{t}_{\mathrm{f}}$ see Figure 5: Output rise and fall times).

Figure 4. Output delay times (e.g. low-side output)


Figure 5. Output rise and fall times


Figure 6. Output disable and enable time (IABE Input)


GAPGPS00643

Figure 7. Output disable and enable time (DIS Input)


Table 13. Current feedback (CF)

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CF}}{ }^{(1)}$ | CF voltage range | $\mathrm{V}_{\mathrm{S}}>6.5 \mathrm{~V}$, OUTx $=0 \mathrm{~A}$, $\mathrm{T}_{\mathrm{J}}=-40^{\circ} \mathrm{C}$; Current level 2,3,4 | 0.01 | 0.05 | 0.20 | V |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{S}}>6.5 \mathrm{~V}, \text { OUTx }= \\ & 250 \mathrm{~mA}, \mathrm{~T}_{\mathrm{J}}=130^{\circ} \mathrm{C} ; \\ & \text { Current level } 2,3,4 \end{aligned}$ | 0.04 | 0.275 | 0.5 | V |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{S}}>6.5 \mathrm{~V} \text {, OUTx }=0.4 \text { * } \\ & \mathrm{I}_{\text {clx }}, \mathrm{T}_{\mathrm{J}}=130^{\circ} \mathrm{C} \text {; Current } \\ & \text { level } 2,3,4 \end{aligned}$ | 1.71 | 1.80 | 1.89 | V |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{S}}>6.5 \mathrm{~V}, \text { OUTx }=\mathrm{I}_{\text {max }}, \\ & \mathrm{T}_{\mathrm{J}}=-40^{\circ} \mathrm{C} \text { to } 150^{\circ} \mathrm{C} ; \\ & \text { Current level } 2,3,4 \end{aligned}$ | 3.82 | 4.5 | 5.18 | V |
| $\mathrm{R}_{\mathrm{CF}}{ }^{(2)}$ | CF resistor range | - | - | 5.1 | - | $\mathrm{k} \Omega$ |
| IOFFSET | CF offset current | - | - | 10 | - | $\mu \mathrm{A}$ |

1. Measured at a 5.1 k resistor between CF and GND ( $\mathrm{R}_{\mathrm{CF}}$ ). Levels see Table 33 Current Level (CONFIG_REG).
2. Defined by design, not tested.

Note: This signal has an individual error $\pm 5 \%$ in each of the three currents levels, at trimming temperature of $130^{\circ} \mathrm{C}$. Additional an individual error $\pm 10 \%$ in each of the three current levels over temperature and aging. So the maximum error is of $\pm 15 \%$ in each of the three current levels. The offset and the gain errors may be different in each current level. The adjustment is done at $130^{\circ} \mathrm{C}$ and compensates the error between 0.3 * Imax to 0.6 * Imax.

Table 14. Current limiting

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\left\|I_{C L 2}\right\|^{(1)}$ | Current limit ${ }_{2}$ | $\mathrm{R}_{\mathrm{CF}}=5.1 \mathrm{k}$ | 4.25 | 5 | 5.75 | A |
| $\left\|\mathrm{I}_{\mathrm{CL3}}\right\|^{(1)}$ | Current limit ${ }_{3}$ |  | 5.6 | 6.6 | 7.6 | A |
| $\left\|I_{C L 4}\right\|^{(1)}$ | Current limit $_{4}$ |  | 7.3 | 8.6 | 9.9 | A |
| $\left\|\\|_{\text {HYS } 2-4}\right\|^{(1)}$ | Current limit hysteresis ${ }_{1}$ | - | $\begin{gathered} -5 \% \\ I_{2} L_{2-4} \end{gathered}$ | - | $\begin{gathered} -10 \% \\ I_{2} L_{2-4} \end{gathered}$ | A |
| $t_{b}$ | Blanking time | Guaranteed through scan. | 8 | 11 | 15 | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {trans }}$ | Time between two transient |  | 90 | - | 130 | $\mu \mathrm{s}$ |

1. Programmable current levels see Table 33 Current Level (CONFIG_REG). Measured using a 5.1 k resistor between CF and GND ( $\mathrm{R}_{\mathrm{CF}}$ ).

Table 15. Over-current detection ( $8 \mathrm{~V}<\mathrm{V}_{\mathrm{S}}<18 \mathrm{~V}$ )

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \| $\mathrm{OCC2}_{\text {LS }}{ }^{(1)}$ | Low side over current threshold 2 | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 4.9 | - | 8.2 | A |
| $\mid \mathrm{lOC3}_{\text {Ls }}{ }^{(1)}$ | Low side over current threshold ${ }_{3}$ | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 6.7 | - | 11.1 | A |
| $\mid \mathrm{lOC4}_{\text {_Ls }}{ }^{(1)}$ | Low side over current threshold 4 | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 8.4 | - | 14 | A |
| $\mathrm{l}_{\mathrm{OC} 2 \text { _ }} \mathrm{SS}^{(1)}$ | High side over current threshold 2 | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 5.5 | - | 9.2 | A |
| $\mid \mathrm{OCC3}_{\text {_HS }}{ }^{(1)}$ | High side over current threshold 3 | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 6.9 | - | 11.5 | A |
| $\mid \mathrm{lOC}_{\text {_ }} \mathrm{HS}^{(1)}$ | High side over current threshold 4 | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 8.6 | - | 14.4 | A |
| $\mathrm{I}_{\text {TRACK-1 }}{ }^{(1)}$ | \| ${ }_{\text {OC1_LS }}$ - \| |lCL1_LS | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 0.4 | - | 5.5 | A |
| $\mathrm{I}_{\text {TRACK-2 }}{ }^{(1)}$ | \| ${ }_{\text {OC2_LS }}-\\|_{\text {CL2_LS }}$ | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 0.4 | - | 5.5 | A |
| $\mathrm{I}_{\text {TRACK-3 }}{ }^{(1)}$ | \| ${ }_{\text {OC3_LS }}$ - \| |lCL3_LS | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 0.4 | - | 5.5 | A |
| $\mathrm{I}_{\text {TRACK-4 }}{ }^{(1)}$ | \|loc4_LSl- |lCL4_LS | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 0.4 | - | 5.5 | A |
| $\mathrm{t}_{\mathrm{DF}}$ | Delay time for fault detection | - | 1 | 2 |  | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {DF_off }}$ | Switch-off delay time | - |  |  | 6 | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {DF_del }}$ | Delayed switch-off time | - | 20 |  | 200 | $\mu \mathrm{s}$ |

1. Programmable current levels see Table 33 Current Level (CONFIG_REG).

## Table 16. Openload detection

If the value of the connected load is below $2.5 \mathrm{k} \Omega$, no Open Load is detected; if the value of the connected load is more than $50 \mathrm{k} \Omega$, Open Load is detected.
If the load is in the range between $(2.5 \ldots 50) \mathrm{k} \Omega$, the Open Load diagnosis is not reliable.

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\mathrm{OL}}$ | Openload detection threshold | - | 5 | - | 50 | k $\Omega$ |
| $t_{\text {DIAGOL }}$ | Openload diagnosis enable delay | Guaranteed through scan. | 100 | - | 150 | ms |
| $\mathrm{t}_{\text {DIAGOL1 }}$ | Openload diagnosis filter time ${ }_{1}$ |  | 2.4 | - | 3.6 | ms |
| $\mathrm{t}_{\text {DIAGOL2 }}$ | Openload diagnosis filter time 2 |  | 200 | - | 300 | $\mu \mathrm{s}$ |

Table 17. Retest delay

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $t_{\text {delay retest }}$ | Retest delay for failures: SCB, <br> SCG, SCL | Guaranteed through <br> scan. | 290 | 350 | 410 | $\mu \mathrm{~s}$ |

### 3.6 Temperature dependent current reduction

Table 18. Temperature dependent current reduction

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{II}_{\text {L_TSD }}$ | Current limit at $\mathrm{T}_{\text {SD }}$ | - | 1.4 | 2.5 | 3.6 | A |
| $\mathrm{~T}_{\text {ILR }}$ | Start of temperature <br> dependent current reduction | - | 150 | 165 | - | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{SD}}$ | Thermal shut-down | - | 175 | - | - | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {SD }}-\mathrm{T}_{\text {ILR }}$ | Range of temperature <br> dependent current reduction | - | 20 | 25 | 30 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {fTSD }}$ | Thermo-shut-down detection <br> filter time | Guaranteed through <br> scan. | 6 | - | 18 | $\mu \mathrm{~s}$ |

Note: see also Figure 16: Temperature dependent current reduction.

### 3.7 Free-wheeling diodes

Table 19. Free-wheel diodes

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{U}_{\mathrm{D}}$ | Free-wheeling diode forward <br> voltage | IOUT $=3 \mathrm{~A}$ | - | - | 2 | V |
| $\mathrm{~T}_{\mathrm{it}}{ }^{(1)}$ | Free-wheeling diode reverse <br> recovery time | - | - | - | 100 | ns |

[^0]
### 3.8 SPI / logic electrical characteristics

The voltages are referred to ground and currents are assumed positive, when the current flows into the pin. $4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{S}} \leq 18 \mathrm{~V}, 4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CC}} \leq 5.5 \mathrm{~V}$; all outputs open; $\mathrm{T}_{\mathrm{j}}=-40^{\circ} \mathrm{C}$ to $150{ }^{\circ} \mathrm{C}$, unless otherwise specified.

Table 20. Inputs: SI, SS, SCK, DIR, DIS and PWM; Output: SO

| Symbol | Parameter | Test condition | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Inputs: SI, SS, SCK DIR, PWM |  |  |  |  |  |  |
| $\mathrm{V}_{\text {IL }}$ | Input voltage low-level | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | -0.3 | - | 0.75 | V |
| $\mathrm{V}_{\mathrm{IH}}$ | Input voltage high-level | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 1.75 | - | VDD+0.3 | V |
| $\mathrm{V}_{\mathrm{IHYS}}$ | Input hysteresis | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 0.2 | - | 1.0 | V |
| $\mathrm{R}_{\text {PUin }}$ | Input pull-up resistor | $\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$ | 50 | - | 250 | $\mathrm{k} \Omega$ |
| $\mathrm{I}_{\text {INx }}$ | PWM, DIR input current | $\mathrm{V}_{\text {INx }}>3.0 \mathrm{~V}$ | -5 | - | 5 | $\mu \mathrm{A}$ |
| $\mathrm{C}_{\text {Slin }}{ }^{(1)}$ | SI input capacitance | - | - | - | 10 | pF |
| $\mathrm{C}_{\text {SCKin }}{ }^{(1)}$ | SCK input capacitance | - | - | - | 10 | pF |
| $\mathrm{C}_{\mathrm{SSin}}{ }^{(1)}$ | SS Input Capacitance | - | - | - | 15 | pF |
| $\mathrm{C}_{\mathrm{DIRPW}_{(1)}}$ | DIR, PWM input capacitance | - | - | - | 20 | pF |
| Input: DIS |  |  |  |  |  |  |
| $\mathrm{R}_{\text {DISPU }}$ | Pull-up resistor | $0 \mathrm{~V}<\mathrm{V}_{\text {DIS }}<2.1 \mathrm{~V}$ | 10 | - | 45 | k $\Omega$ |
| $\mathrm{I}_{\text {DISx }}$ | DIS input current | $\mathrm{V}_{\text {DIS }}>3 \mathrm{~V}$ | -5 | - | 5 | $\mu \mathrm{A}$ |
| $\mathrm{C}_{\text {DIS in }}{ }^{(1)}$ | DIS input capacitance | - | - | - | 20 | pF |
| tDIS | DIS pulse width | - | 0.5 | 1 | 1.5 | $\mu \mathrm{s}$ |

Input pin disturbance (SI, SS, SCK DIR, PWM,DIS)

| $\Delta \mathrm{Vx}$ _HL | Change of $\mathrm{V}_{\mathrm{IH}}$ and $\mathrm{V}_{\mathrm{IL}}$ during inverse current on $V_{S}$ | Not subjected to test in production. | -0.1 | - | 0.1 | V |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\Delta \mathrm{ISx}$ | Change of input current of SPI input pins during inverse current on VS |  | -100 | - | 100 | $\mu \mathrm{A}$ |

Output: SO

| $\mathrm{V}_{\mathrm{SOL}}$ | Output voltage low level | $\mathrm{I}_{\mathrm{OL}}=2 \mathrm{~mA}$, | 0 | - | 0.4 | V |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{~V}_{\mathrm{SOH}}$ | Output voltage high level | $\mathrm{I}_{\mathrm{OH}}=-2 \mathrm{~mA}$ | $\mathrm{VDD}-0.5$ | - | VDD | V |
| $\mathrm{SR}_{\text {SO }}{ }^{(1)}$ | Slew rate | $\mathrm{C}_{\mathrm{LOAD}}=200 \mathrm{pF}$ | 0.3 | - | 0.6 | $\mathrm{~V} / \mathrm{ns}$ |
| $\mathrm{I}_{\text {SOLK }}$ | Tristate leakage current | $\mathrm{V}_{\mathrm{SS}}=\mathrm{V}_{\mathrm{DD}}$ | -10 | - | 10 | $\mu \mathrm{~A}$ |
| $\mathrm{C}_{\text {SO out }}{ }^{(1)}$ | SO output capacitance | - | - | - | 10 | pF |

## Output pin disturbance (SO)

| $\Delta I_{\text {SOLK }}$ | Change of $I_{\text {SOLK }}$ during <br> inverse Current on VS | - | -100 | - | 100 | $\mu \mathrm{~A}$ |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |

[^1]Table 21. Dynamic characteristics

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {cyc }}$ | Cycle time | - | 490 | - | - | ns |
| $\mathrm{t}_{\text {lead }}$ | Enable lead time | - | 300 | - | - | ns |
| $\mathrm{t}_{\text {lag }}$ | Enable lag time | - | 150 | - | - | ns |
| $\mathrm{t}_{\mathrm{v}}$ | Data valid | $\begin{aligned} & \mathrm{SCK}=2 \mathrm{~V} ; \mathrm{SO}=0.2 \mathrm{~V} ; \\ & \mathrm{C}_{\mathrm{L}}=40 \mathrm{pF} \end{aligned}$ | 40 | - | - | ns |
|  |  | $\begin{aligned} & \mathrm{SCK}=2 \mathrm{~V} ; \mathrm{SO}=0.2 \mathrm{~V} ; \\ & \mathrm{C}_{\mathrm{L}}=200 \mathrm{pF} \end{aligned}$ | 150 | - | - | ns |
|  |  | $\begin{aligned} & \mathrm{SCK}=2 \mathrm{~V} ; \mathrm{SO}=0.2 \mathrm{~V} ; \\ & \mathrm{C}_{\mathrm{L}}=350 \mathrm{pF} \end{aligned}$ | 230 | - | - | ns |
| $\mathrm{t}_{\text {su }}$ | Data setup time | - | 40 | - | - | ns |
| $t_{\text {h }}$ | Data hold time | - | 40 | - | - | ns |
| $\mathrm{t}_{\text {dis }}$ | Disable time | - | 0 | - | 100 | ns |
| $\mathrm{t}_{\mathrm{dt}}$ | Transfer delay | - | 300 | - | - | ns |
| $\mathrm{t}_{\text {dld }}$ | Disable lead time | - | 250 | - | - | ns |
| $\mathrm{t}_{\text {dlg }}$ | Disable lag time | - | 250 | - | - | ns |
| $\mathrm{t}_{\text {acc }}$ | Access time | - | 8.35 | - | - | $\mu \mathrm{s}$ |

Figure 8. SPI timing information


## 4 Application information

### 4.1 Power stage switching behavior

The L9959 output stages can either be controlled by the pins PWM and DIR or by their corresponding SPI registers (SPWM and SDIR: see Table 32: Configuration register (CONFIG_REG)). The SPI bit MUX in the configuration register (CONFIG_REG) determines this. If the power stages are disabled by IABE or DIS, this bit is reset and the pins PWM and DIR control the outputs.
The active free-wheeling, in which the body diode is actively shorted by its associated Power-MOS, can be disabled by the bit FW in the configuration register. By default, active free-wheeling is enabled.

The device minimizes electro-magnetic emission by switching the high-side and low-side drivers in a special sequence. Two cases are distinguished: The PWM-mode, during which the current direction does not change and the direction switch using the DIR, which changes the current direction (see Figure 9, Figure 11 and Figure 12).

### 4.1.1 PWM mode (same current direction)

The PWM input pin switches the high-/low-side output of the half-bridge, which is selected by the DIR pin. DIR = ' 0 ': OUT1 is switched, DIR = '1': OUT2 is switched.

PWM = '0': Switched low-side is on, PWM = '1': Switched high-side is on.
Figure 9. PWM mode current flow


Figure 10. PWM mode output voltage


During PWM mode the high-side (e.g. Figure 9 HS1) output is switched off with a slow slew rate until it is off and the low-side body-diode has taken over the entire current. Then the associated low-side transistor (e.g. Figure 9 LS1) is turned on with a fast slope to reduce the voltage across the device and to minimize the power.
The output is pulled to high voltage, by first turning off the low-side driver with a fast slew rate and, after it is off, the high-side driver is switched on by a slow one (e.g. Figure 9 LS1, HS1).

This assures, that the voltage and current change over the body diode is done smoothly, reducing the electromagnetic emission.

Figure 11. DIR-change (current is changing its direction)


Figure 12. DIR-change current flow phase 2


Figure 13. DIR-change output voltage


The first part of the sequence is identical to the PWM-mode (s.a.). After this has been finished and the associated low-side driver is on (e.g. Figure 11 LS1), in phase 1 the other low-side driver is turned on (e.g Figure 11 LS2). Then in phase 2 the low-side output of OUT2 is switched-off slowly and the current through the load is taken over by the bodydiode of the high-side (e.g. Figure 12 HS2). Depending on the inductance of the load, the current vanishes more or less quickly. After the low-side driver is turned off, the high-side is switched on with a slow slew-rate.

This assures, that direction switch occurs while the current over the load has vanished, which reduces the electromagnetic emission.

### 4.2 Protection and monitoring

All errors are confirmed after their occurrence by accessing the error condition after time $t d_{\text {elay_retest }}$ a second time. Only after the error is confirmed it is entered into the diagnosis register 1 (DIA_REG1), and the device is disabled and no further diagnosis is run.
The device can be enabled again by following actions: Power-on reset, disabling or enabling the device using the pins /ABE or DIS (e.g. disabling - enabling sequence). The diagnosis registers can be cleared by sending a reset command by SPI (STATCON_REG) to either diagnosis register 1 (DIA_REG1) or 2 (DIA_REG2). The bit1 (Reset) of the CONFIG_REG if forced to zero is resetting both the device registers configuration and diagnosis registers to default but is not able to restart the device. In order to restart IC is necessary to force a transition LOW/HIGH/LOW on DIS pin or a transition HIGH/LOW/HIGH on /ABE pin.

The errors in the diagnosis register 1 (DIA_REG1) are transferred to the diagnosis register 2 by setting the bit DIACLR1 in the status and configuration register (STATCON_REG) or by using the enabling -disabling sequence on /ABE or DIS. This will also clear the diagnosis register 1.

### 4.2.1 Current feedback

A feedback current signal is provided at pin CF (Current Feedback). This current is proportional to the current in the H -Bridge, but does not change its direction. It is measured in the low-side transistor, which is not switched by PWM. This is determined by the input DIR or the SDIR register respectively. Therefore, the direction of the current can be seen from this direction signal. Whenever the current direction changes, the current feedback signal is blanked. In Table 13: Current feedback (CF) the CF behavior over an external resistor of 5.1 k Ohm is specified. The current out of CF consists of a static offset current and a current proportional to the current in the select low-side transistor. The voltage at pin CF scales with the resistor at this pin.

Figure 14. Current feedback and current limiting


Figure 14 Current Feedback and Current Limiting shows the current feedback in case the OUT1 is controlled by PWM (DIR = 0). In this case, the current is measured through lowside 2. If the direction is inverted, the current is measured through low-side 1.

### 4.2.2 Current limitation

The H-Bridge output current can be limited to three different values (see Table 14: Current limiting). If the current reaches the current limiting threshold ICL, the output driver is switched off after the blanking time $t_{b}$, and switched on again after the current dropped below the lower current limit hysteresis threshold ( $\mathrm{I}_{\mathrm{CL}}-\mathrm{I}_{\mathrm{HYS}}$ ). The current limiting thresholds can be adjusted using the resistor at pin CF. The values in Table 14 refer to a 5.1 k Ohm external resistor. The current limiting threshold can be calculated by (4.5V/R $\mathrm{CF}^{-}$ $\left.\mathrm{I}_{\mathrm{OFFSET}}\right)$ * ( $\mathrm{I}_{\mathrm{CLx}} * 5.1 \mathrm{k} / 4.45 \mathrm{~V}$ ) from Table 14 and $\mathrm{I}_{\mathrm{OFFSET}}=10 \mu \mathrm{~A}$ (typical). $\mathrm{R}_{\mathrm{CF}}$ is the resistor used at pin CF. The overcurrent threshold is not changed by RCF (see Table 15). The current limitation is active as long as the output driver is switched on. The current limitation is also active during free-wheeling. The information that the device is in current limitation is stored in the diagnosis register 2 (DIA_REG2).

Figure 15. Current limitation


### 4.2.3 Temperature dependent current reduction

If the device reaches the temperature TILR, the current will be reduced (see Figure 16: Temperature dependent current reduction). If the temperature reaches the temperature shutdown threshold, the outputs are switched off. The current limitation information is written into the diagnosis register 2 (DIA_REG2).

Figure 16. Temperature dependent current reduction


### 4.2.4 Short to battery (SCB) and short to Ground (SCG)

While the power stages are on, the current through them is monitored. If the output current reaches the current limit IL, the output is switched off after the blanking time $t_{b}$. In case the current reaches the limit $\mathrm{I}_{\mathrm{Oc}}$ during this time, a short to battery (SCB) on low-sides or a short to ground (SCG) on high-sides is diagnosed, the affected output driver is switched off immediately, the not affected one after the time $t_{D F}$ del. In order to confirm this error, the outputs are turned on again. Only if the error is then again detected, it is entered into the diagnosis register (DIA_REG1) and the device is disabled. Otherwise, the event is disregarded.

Figure 17. Current limiting and short circuit


The three different over-current limits are related to the programmable current limitation, which can be programmed into the SPI register (Table 32: Configuration register (CONFIG_REG)). The over-current limits are independent of the resistor at pin CF.

### 4.2.5 Short circuit over load (SCL)

Short circuit over load (SCL) is diagnosed by a retest sequence after a short to battery (SCB) or a short to ground (SCG) has been detected and confirmed by a retry on the switched-on high-side and low-side driver. Then after the time $\mathrm{t}_{\text {SCLretest }}$, the opposite driver is switched off (i.e. the high-side in case of a short to battery on the low-side and vice versa). If the failure then disappears, a short over load (SCL) is detected.

The error is only entered into the diagnosis register and the device is disabled, if it is confirmed.

### 4.2.6 Open load (OL)

Open load can either be detected in active mode or while the output drivers are in tri-state condition, disabled by DIS or /ABE. Open load in active mode is enabled using the OLDA in the configuration register (CONFIG_REG).

## Open load in active mode

The open load condition can only be diagnosed if an inductive load is used. In normal operation, the output free wheels via the built-in diodes below ground, if the high-side output driver is switched-off. If the output does not go below ground, an open load is detected.

## Open load in inactive mode

In inactive mode the open load condition is detected by applying a pull-down current ( $\mathrm{I}_{\mathrm{PD}}$ ) to both outputs. A pull-up current is generated at one output to compensate these two output currents. If the pull-up current is in the range of one pull-down current, an open load is diagnosed. If the load is connected, the pull up current is in the range of the sum of both pull-down currents. An open load is detected, if the load resistance is above the open load resistance threshold, no open load is detected, if it is below this threshold ( $\mathrm{R}_{\mathrm{OL}}$ ). After the outputs are disabled, it takes the time $t_{\text {DIAGOL }}$ until the open load diagnostic can be enabled. The open load settling time to reach the correct pull up current is $t_{\text {diagOL1 }}$, the open load filter time is $\mathrm{t}_{\text {diagOL2 }}$.

### 4.3 VS-undervoltage

VS is monitored for under-voltage. If VS goes below the VS-undervoltage threshold, the outputs are switched to tristate after the time $\mathrm{t}_{\text {FUV }}$.

### 4.4 Inverse current at $\mathbf{V}_{\mathbf{S}}$

An inverse current of maximum 5 A, which decreases during a period of max 250 ms out of the device at VS does not lead to any destruction. After the exposure to such an inverse current the device returns to the specified functionality.

### 4.5 IABE pin

/ABE (Ability/Enable) is a bidirectional pin, with an open-drain output. In normal operating condition, this pin is pulled up by an external resistor. If /ABE is set to low, the outputs enter tristate mode.

IABE can be used to switch off the outputs quickly by an external signal. It is especially possible to connect the /ABE pins of several devices together, so all of them can be disabled in case one detects an error, which is flagged by the /ABE-pin.

### 4.6 VDD-monitor

$\mathrm{V}_{\mathrm{DD}}$ is monitored for under- and over-voltage referenced to $\mathrm{GND}_{\mathrm{ABE}}$. If $\mathrm{V}_{\mathrm{DD}}$ goes below $V_{D D}$ THL or above $V_{D D T H}$, IABE is pulled to low and the outputs enter tristate mode after the time $t_{\text {FIL }}$ OFF. The VDD-monitoring state is stored into the status and control register (STACON_REG). If VDD increases above $\mathrm{V}_{\mathrm{DD}}$ THL, IABE is pulled to high after the filter time $t_{\text {FIL_ON }}$. The SPI remains functional as long as $V_{D D}$ is above the power-on reset threshold.

The behavior of the pin /ABE and the output stages after VDD goes below VDD_THH from VDD-overvoltage is determined by bit CONFIG 0 in the status and configuration register (STATCON):

CONFIGO = 1: $/$ ABE is latched and the outputs remain in tristate
CONFIGO $=0$ : $/$ ABE goes to inactive and the output stages are enabled after the filtering time $\mathrm{t}_{\text {FIL_ON }}$.

### 4.7 VDD-monitor test

VDD-Monitor blocks can be tested in the application via SPI. During this test, the output stages are still switched off in case of over- and under-voltage.

## Upper threshold

The over-voltage threshold can be reduced using the configuration registers 1 and 2 (CONFIG1 and CONFIG2) in the status and control register (STACON_REG) to $\mathrm{V}_{\text {TEST_THH }}$ (see TTable 34: Status and configuration register (STATCON_REG)). Since $\mathrm{V}_{\text {TEST THH }}$ is below the normal VDD voltage, the status bit STATUS0 shows a VDD overvoltage.

## Lower threshold

The under-voltage threshold can be increased to $\mathrm{V}_{\text {TEST_THL }}$ using CONFIG1 and CONFIG2 in the STATCON register. Since the VDD voltage is below $\mathrm{V}_{\text {TEST_THL }}$, the resulting VDDundervoltage resets STATUSO.

After leaving the VDD-monitor test mode, the bits in the STACON register return to their normal state.

### 4.8 Power-on reset

At power-on, while VDD increases, the internal registers are cleared and the outputs are set to tristate at the reset-active voltage $\mathrm{V}_{\text {DDRES }}$. Above the power-on reset threshold $\mathrm{V}_{\text {DDPOR }}$ the device starts to operate after the time $t_{P O R}$. If $V D D$ drops below $V_{D D P O R}$, the device enters its reset state, i.e. all internal registers are cleared and the outputs are set to tristate.

Table 22. Device states with respect to supply voltage

| $\mathrm{V}_{\mathbf{S}}[\mathrm{V}]$ | $\mathrm{V}_{\mathrm{DD}}[\mathrm{V}]$ | Functional state |
| :---: | :---: | :--- |
| $28-40$ | $0-18$ | No damage to the device, no functional behavior <br> guaranteed |
| $4.5-6.5$ | $4.5-5.5$ | Device functional, Current Feedback accuracy reduced |
| $6.5-28$ | $4.5-5.5$ | Device functional |
| $4.5-28$ | $4.0-4.55 .5-18$ | Device functional, but power-outputs tristate by <br> VDD-monitor, /ABE pulled to low, SPI functional |
| $0-4.54 .5-28$ | VDD $_{\text {POR }}-4.5$ | Device in reset mode, SPI functional, power-outputs <br> tristate, /ABE pulled to low |
| $0-4.54 .5-28$ | $2.5-$ VDD $_{\text {POR }}$ | Device in reset mode, SPI reset, power-outputs tristate, <br> IABE tristate |
| $0-4.5$ | $4.5-5.5$ | Device functional, outputs are tristate by Vs- <br> undervoltage |
| $0-4.5$ | $4.0-4.55 .5-18$ | Device functional, outputs are tristate by Vs- <br> undervoltage and VDD-monitor, /ABE pulled to low |

Note: $\quad$ All voltages are nominal. Please refer to Section 3: Electrical specifications for their specified values.

## 5 SPI functional description

### 5.1 General description

The SPI communication is based on a Serial Peripheral Interface structure using SS (SPI Select), SI (Serial Data In), SO (Serial Data Out) and SCK (Serial Clock) signal lines. The first data at pin SI is latched into the device with the first falling edge of the clock SCK after the clock has changed from low to high, which is the second edge after SPI-Select has been pulled to low. Therefore the $\mu \mathrm{C}$ protocol is according to $\mathrm{CPOL}=0$ and CPHA $=1$ (see Figure 8: SPI timing information).

### 5.1.1 SPI select (SS)

The SS input pin is used to select the serial interface of this device. When SS is high, the output pin (SO) is in high impedance state. A low signal starts the serial communication. A communication frame is the time between the falling edge of $S S$ and its rising edge.

### 5.1.2 Serial data In (SI)

The SI input pin is used to transfer data serially into the device. The data applied to the SI is sampled at the falling edge of the SCK signal.

### 5.1.3 Serial clock (SCK)

The Data Input (SI) is latched at the falling edge of Serial Clock SCK. Data on Serial Data Out (SO) is shifted out at the rising edge of the serial clock (SCK). The serial clock SCK must be active only during a frame (SS low).

### 5.1.4 Serial out (SO)

The content of the selected status or control register is transferred out of the device using the SO pin on the rising edge of SCK. Each subsequent rising edge of the SCK will shift the next bit out.

### 5.1.5 SPI communication flow

The SPI communication is started by sending an SPI instruction to the device beginning with the MSB. The first two bits of this instruction are used as a device identifier (see Table 23: SPI instruction byte). Whether the transfer is a read or a write access is determined by the SPI command (see Table 25: Command overview). The SPI data is transmitted from the device at the same time as the data is received, although on different SCK edges. While the 8 -bit instruction is sent, the device responds with the check byte. Since the first two bits of the instruction are used as a device identifier, the first two bits of the check byte are tristate. This avoids bus conflicts on the SO line. During a write access, the 8 -bit data byte is received after the instruction byte. The device responds with $00_{H}$. In a read cycle the device sends the 8-bit data, while the receive data bits are ignored (see Figure 18: Write access and Figure 19: Read access). If an invalid instruction is detected, the register of the device are not modified and the data byte $\mathrm{FF}_{\mathrm{H}}$ is transmitted instead of the data or $00_{\mathrm{H}}$ respectively. The bit TRANS_F in the check byte is set in case of an invalid instruction and transmitted during the next SPI-access. An instruction is invalid, if an unused instruction code is
detected, the previous transmission has not been completed or the number of clocks is not equal to 16.

Figure 18. Write access


Figure 19. Read access


### 5.2 SPI-instruction

Table 23. SPI instruction byte

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Name | CPAD1 | CPADO | INST $<5>$ | INST<4> | INST < 3 > | INST<2> | INST<1> | INST<0> |
| Bit | Name | Content |  |  |  |  |  |  |
| 7 | CPAD1 | Chip Address: 0 |  |  |  |  |  |  |
| 6 | CPAD0 | Chip Address: 0 |  |  |  |  |  |  |
| 5 | INST<5> | Read/Write: Read: 0 Write: 1 |  |  |  |  |  |  |
| 4:0 | INST<4:0> | SPI Instruction |  |  |  |  |  |  |

Table 24. Check byte

| Bit | $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Name | Tristate | Tristate | 1 | 0 | 1 | 0 | 1 | TRANS_F |
| Bit | Name | Content |  |  |  |  |  |  |
| $7: 6$ | Tristate | Tristate |  |  |  |  |  |  |
| $6: 1$ |  | Fix Content: 10101 |  |  |  |  |  |  |
| 0 | TRANS_F | Transfer-Failure |  |  |  |  |  |  |

### 5.3 Device register map

Table 25. Command overview

| Command | INST<5:0> | Content |
| :---: | :---: | :--- |
| RD_ID | $00 \_0100$ | Read Device ID |
| RD_REV | $00 \_0110$ | Read Device Revision |
| RD_DIA1 | $01 \_0000$ | Read Diagnostic Information Register 1 |
| RD_DIA2 | $01 \_1000$ | Read Diagnostic Information Register 2 |
| RD_CONFIG | $00 \_1000$ | Read Configuration |
| RD_STATCON | $00 \_1100$ | Read VDD Monitoring Status |
| RD_SPECIAL | $00 \_1110$ | Read information from SPECIAL |
| WR_DIA1 | $11 \_0000$ | Write to Diagnostic Information Register 1 |
| WR_DIA2 | $11 \_1000$ | Write to Diagnostic Information Register 2 |
| WR_CONFIG | $10 \_1000$ | Write Configuration |
| WR_STATCON | $10 \_1100$ | Write VDD Monitoring Status |
| WR_SPECIAL | $10 \_1110$ | Write information to SPECIAL |
| All Other | - | Invalid Command: TRANS_F: 1 |

### 5.4 SPI - control and status registers

Table 26. Device identifier (ID)

| Bit | $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :--- | :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Name | $\mid \mathrm{ID}<7>$ | $\mathrm{ID}<6>$ | $\mathrm{ID}<5>$ | $\mathrm{ID}<4>$ | $\mathrm{ID}<3>$ | $\mathrm{ID}<2>$ | $\mathrm{ID}<1>$ | $\mathrm{ID}<0>$ |
| Bit | Name | Content |  |  |  |  |  |  |
| $7: 0$ | $I D<7: 0>$ | Device ID: DFH |  |  |  |  |  |  |

Table 27. Revision register (REV)

| Bit | $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Name | $\mathrm{SWR}<3>$ | $\mathrm{SWR}<2>$ | $\mathrm{SWR}<1>$ | $\mathrm{SWR}<0>$ | $\mathrm{MSR}<3>$ | $\mathrm{MSR}<2>$ | $\mathrm{MSR}<1>$ | $\mathrm{MSR}<0>$ |
| Bit | Name | Content |  |  |  |  |  |  |
| $7: 4$ | SWR<3:0> | Software Revision: 0 H |  |  |  |  |  |  |
| $3: 0$ | $M S R<3: 0>$ | Mask Set Revision: 03 H |  |  |  |  |  |  |

Table 28. DIA_REG1

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Name | /ABE / DIS | OT | Res | Res | DIA21 | DIA20 | DIA11 | DIA10 |
| Bit | Name | Content |  |  |  |  |  |  |
| 7 | IABE / DIS | Disable: 0 , if $/ \mathrm{ABE}=0$ or DIS $=1$ |  |  |  |  |  |  |
| 6 | OT | Over temperature |  |  |  |  |  |  |
| 5:4 | Reserved | 0 |  |  |  |  |  |  |
| 3 | DIA21 | Diagnose Bits (Table 29: Diagnosis bits (DIA_REG1)) |  |  |  |  |  |  |
| 2 | DIA20 |  |  |  |  |  |  |  |
| 1 | DIA11 |  |  |  |  |  |  |  |
| 0 | DIA10 |  |  |  |  |  |  |  |
| Reset ${ }^{(5.4}$. <br> 1) | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| POR | X | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| SPIR | X | 1 | X | X | 1 | 1 | 1 | 1 |
| ENDISR | X | 1 | X | X | 1 | 1 | 1 | 1 |
| RDR | X | X | X | X | X | X | X | X |
| DIACLR1 | X | 1 | X | X | 1 | 1 | 1 | 1 |

Table 29. Diagnosis bits (DIA_REG1)

| DIA21 | DIA20 | DIA11 | DIA10 | Description | Remark |
| :---: | :---: | :---: | :---: | :--- | :---: |
| 0 | 0 | 0 | 1 | Short Circuit to Ground at OUT1 (SCG1) | Latched |
| 0 | 0 | 1 | 0 | Short Circuit to Ground at OUT2 (SCG2) | Latched |
| 0 | 1 | 0 | 1 | Short Circuit to Battery at OUT1 (SCB1) | Latched |
| 0 | 1 | 1 | 0 | Short Circuit to Battery at OUT2 (SCB2) | Latched |
| 0 | 1 | 1 | 1 | Short Circuit over Load (SCL) | Latched |
| 1 | 0 | 0 | 0 | Short Circuit to Battery at Disabled Output | Latched |
| 1 | 0 | 0 | 1 | Short Circuit to Ground at Disabled Output | Latched |
| 1 | 0 | 1 | 0 | Open Load at disabled or active Output (OL) | Latched |
| 1 | 1 | 0 | 1 | Under Voltage at VS | Not Latched |
| 1 | 1 | 1 | 1 | No Failure | - |

Reading this register does not reset the bits. Writing STACON_REG.DIACLR1 $=0$ transfers all latched errors to DIA_REG2 and resets DIA_REG1 afterwards, if there is no VSundervoltage.

Table 30. Diagnosis register 2 (DIA_REG2)

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Name | CurrRed | CurrLim | OT | Res | DIA21 | DIA20 | DIA11 | DIA10 |
| Bit | Name | Content |  |  |  |  |  |  |
| 7 | CurrRed | Current Reduction: 0 , if temperature dependent current reduction is active This information bit is reset after each read access |  |  |  |  |  |  |
| 6 | CurrLim | Current Limitation: 0 , if current limitation is active This information bit is reset after each read access |  |  |  |  |  |  |
| 5 | OT | Over temperature |  |  |  |  |  |  |
| 4 | Reserved | 0 |  |  |  |  |  |  |
| 3 | DIA21 | Diagnosis Bits (see Table 31: Diagnosis bits (DIA_REG2)) |  |  |  |  |  |  |
| 2 | DIA20 |  |  |  |  |  |  |  |
| 1 | DIA11 |  |  |  |  |  |  |  |
| 0 | DIA10 |  |  |  |  |  |  |  |
| $\text { Reset }{ }^{(5.4}$ 1) | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| POR | 1 | 1 | 1 | 0 | 1 | 1 | 1 | 1 |
| SPIR | 1 | 1 | 1 | X | 1 | 1 | 1 | 1 |
| ENDISR | 1 | 1 | X | X | X | X | X | X |
| RDR | 1 | 1 | X | X | X | X | X | X |
| DIACLR2 | 1 | 1 | 1 | X | 1 | 1 | 1 | 1 |

Table 31. Diagnosis bits (DIA_REG2)

| DIA21 | DIA20 | DIA11 | DIA10 | Description | Remark |
| :---: | :---: | :---: | :---: | :--- | :---: |
| 0 | 0 | 0 | 1 | Short Circuit to Ground at OUT1 (SCG1) | Latched |
| 0 | 0 | 1 | 0 | Short Circuit to Ground at OUT2 (SCG2) | Latched |
| 0 | 1 | 0 | 1 | Short Circuit to Battery at OUT1 (SCB1) | Latched |
| 0 | 1 | 1 | 0 | Short Circuit to Battery at OUT2 (SCB2) | Latched |
| 0 | 1 | 1 | 1 | Short Circuit over Load (SCL) | Latched |
| 1 | 0 | 0 | 0 | Short Circuit to Battery at Disabled Output | Latched |
| 1 | 0 | 0 | 1 | Short Circuit to Ground at Disabled Output | Latched |
| 1 | 0 | 1 | 0 | Open Load at disabled or active Output (OL) | Latched |
| 1 | 1 | 1 | 1 | No Failure | - |

Table 32. Configuration register (CONFIG_REG)

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Name | FW | MUX | SPWM | SDIR | CL1 | CL2 | RESET | OLDA |
| Bit | Name | Content |  |  |  |  |  |  |
| 7 | FW | Free-Wheeling: 0: FW via Body Diode; 1: FW with active short of Body Diode |  |  |  |  |  |  |
| 6 | MUX | Multiplex Bit: 0: control by bits SPWM and SDIR; 1: Control by inputs PWM and DIR |  |  |  |  |  |  |
| 5 | SDIR | Direction: Same as input DIR |  |  |  |  |  |  |
| 4 | SPWM | PWM: Same as input PWM |  |  |  |  |  |  |
| 3 | CL1 | See Table 33: Current Level (CONFIG_REG). |  |  |  |  |  |  |
| 2 | CL2 |  |  |  |  |  |  |  |
| 1 | RESET | Reset: 0: Reset of device configuration to default; 1: No change |  |  |  |  |  |  |
| 0 | OLDA | Open-Load Diagnosis in active mode: 1: OLD on; 0: OLD off |  |  |  |  |  |  |
| $\text { Reset }(5.4$ 1) | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| POR | 1 | 1 | 1 | 1 | 1 | 0 | 1 | 0 |
| SPIR | 1 | 1 | 1 | 1 | 1 | 0 | 1 | 0 |
| ENDISR | X | X | X | X | X | X | 1 | X |
| DISR | X | 1 | 1 | 1 | X | X | 1 | X |
| RDR | X | X | X | X | X | X | 1 | X |

Only the bit 'RESET' in this register can be written, all other bits are 'read-only'

Table 33. Current Level (CONFIG_REG)

| CL1 | CL2 | Current Level | Typical Current |
| :---: | :---: | :---: | :---: |
| 0 | 0 | No Change | No Change |
| 0 | 1 | 2 | 5.0 A |
| 1 | 0 | 3 (default value) | 6.6 A |
| 1 | 1 | 4 | 8.6 A |

Table 34. Status and configuration register (STATCON_REG)

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Name | CONFIG2 | CONFIG1 | CONFIGO | DIACLR2 | DIACLR1 | STATUS2 | STATUS1 | STATUS0 |
| Bit | Name | Content |  |  |  |  |  |  |
| 7 | CONFIG2 | VDD Test Threshold: 0 : VDD Threshold Test is on 1: VDD Threshold test is off |  |  |  |  |  |  |
| 6 | CONFIG1 | VDD Test Threshold 1:Lower VDD Test Threshold is lifted 0:Upper VDD Test Threshold is lowered |  |  |  |  |  |  |
| 5 | CONFIGO | VDD Over-Voltage Latch: 0 : Latch is disabled 1: Latch is enabled |  |  |  |  |  |  |
| 4 | DIACLR2 | Reset DIA_REG2: 0: Reset errors in DIA_REG2 1: No action (Reading this bit always returns „1.) |  |  |  |  |  |  |
| 3 | DIACLR1 | Transfer Errors: <br> 0: All latched errors of DIA_REG1 are transferred to DIA_REG2. DIA21,DIA20, DIA11, DIA10are set to "1111". During VS-Undervoltage DIACLR1 is disabled 1: No action (Reading this bit always returns „1.) |  |  |  |  |  |  |
| 2 | STATUS2 | Logic Level at Pin /ABE |  |  |  |  |  |  |
| 1 | STATUS1 | VDD Under-Voltage: 0: Under-Voltage 1: No Under-Voltage |  |  |  |  |  |  |
| 0 | STATUS0 | VDD Over-Voltage: 0: Over-Voltage <br> 1: No Over-Voltage <br> This information is not reset during VS-Undervoltage. It will be reset by CONFIGO, SPI reset or internal VDD reset |  |  |  |  |  |  |
| Reset ${ }^{\text {5.4.1) }}$ | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| POR | 1 | 1 | 0 | 1 | 1 | X | 1 | 1 |
| SPIR | 1 | 1 | 0 | 1 | 1 | 1 | 1 | 1 |
| ENDISR | X | X | X | 1 | 1 | 1 | 1 | 1 |
| DISR | X | X | X | 1 | 1 | 1 | 1 | 1 |
| RDR | X | X | X | 1 | 1 | 1 | 1 | 1 |

Only the bits 'CONFIG' and 'DIACLR' in this register can be written, all other bits are 'readonly'

Table 35. Special register (SPECIAL_REG)

| Bit | $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Name | Not specified |  |  |  | OLDAFILT <br> ER | Not <br> specified | SR | Not <br> specified |
| SPRCSPE <br> C |  |  |  |  |  |  |  |  |

Controller access:
write-access: WR_SPECIAL
read-access: RD_SPECIAL

| Bit | Name | Content |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7 | Not specified | - |  |  |  |  |  |  |
| 6 | Not specified | - |  |  |  |  |  |  |
| 5 | Not specified | - |  |  |  |  |  |  |
| 4 | OLDAFILT ER | 1: openload in on detected after one retry 0 : openload in on detected after 40 ms |  |  |  |  |  |  |
| 3 | Not specified | - |  |  |  |  |  |  |
| 2 | SR | 1: fast slew rate 0: slow slew rate |  |  |  |  |  |  |
| 1 | Not specified | - |  |  |  |  |  |  |
| 0 | SPRCSPE <br> C | 1: spread spectrum = active <br> 0: spread spectrum = disabled <br> Spread spectrum = active provides the internal state machine with slightly jittering clock. <br> Spread spectrum = disabled. The internal state machine runs with constant clock frequency. |  |  |  |  |  |  |
| Reset (5.4.1) | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| POR | 0 | 1 | 0 | 1 | 1 | 0 | 0 | 1 |
| SPIR | 0 | 1 | 0 | 1 | 1 | 0 | 0 | 1 |
| ENDISR | X | X | X | X | X | X | X | X |
| DISR | X | X | X | X | X | X | X | X |
| RDR | X | X | X | X | X | X | X | X |

### 5.4.1 Reset sources

- POR: Reset due to a VDD power up on VDD (Power-Up Reset)
- ENDISR: Reset caused by an enable or disable of the power stages (DIS or /ABE edge triggered) (Enable-/Disable Reset)
- DISR: Reset caused by disabling the power stages (DIS or /ABE level triggered) (Disable power stage reset)


### 5.4.2 Configuration registers reset sources

- POR: Reset due to a VDD power up on VDD (Power-Up Reset)
- SPIR: Reset by setting bit RESET in the configuration register (CONFIG_REG) (SPIReset)
- RDIR: Reset caused by a read access to the corresponding register (Read Register)
- DIACLR1: Reset by setting bit DIACLR1 in the Status and Configuration Register STATCON (Diagnosis Reset 1)
- DIACLR2: Reset by setting bit DIACLR2 in the Status and Configuration Register STATCON (Diagnosis Reset 2)


## 6 Application circuit

Figure 20. Application circuit


## 7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK ${ }^{\circledR}$ packages, depending on their level of environmental compliance. ECOPACK ${ }^{\circledR}$ specifications, grade definitions and product status are available at: www.st.com.
ECOPACK ${ }^{\circledR}$ is an ST trademark.

Figure 21. PowerSSO24 mechanical data and package dimensions

| Dim. | mm |  |  | inch |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. | Min. | Typ. | Max. |
| A |  |  | 2.45 |  |  | 0.0965 |
| A2 | 2.15 |  | 2.35 | 0.084 |  | 0.0925 |
| a1 | 0 |  | 0.10 | 0 |  | 0.003 |
| b | 0.33 |  | 0.51 | 0.013 |  | 0.020 |
| c | 0.23 |  | 0.32 | 0.009 |  | 0.012 |
| $\mathrm{D}^{(1)}$ | 10.10 |  | 10.50 | 0.398 |  | 0.413 |
| $E^{(1)}$ | 7.40 |  | 7.60 | 0.291 |  | 0.299 |
| e0 |  | . 80 |  |  | 0.031 |  |
| e3 |  | 8.80 |  |  | 0.346 |  |
| F2 |  | . 30 |  |  | 0.090 |  |
| G |  |  | 0.10 |  |  | 0.004 |
| G1 |  |  | 0.06 |  |  | 0.002 |
| H | 10.10 |  | 10.50 | 0.398 |  | 0.413 |
| h |  |  | 0.40 |  |  | 0.016 |
| k | $0^{\circ}$ (min.), $8^{\circ}$ (max.) |  |  |  |  |  |
| L | 0.55 |  | 0.85 | 0.0217 |  | 0.0335 |
| 0 |  | 1.20 |  |  |  | 0.047 |
| Q |  | 0.80 |  |  |  | 0.031 |
| S |  | 2.90 |  |  |  | 0.114 |
| T |  | 3.65 |  |  |  | 0.143 |
| U |  | 1.0 |  |  |  | 0.039 |
| N | $10^{\circ}$ (max) |  |  |  |  |  |
| X | 4.10 |  | 4.70 | 0.161 |  | 0.185 |
| Y | $\begin{gathered} \hline 6.50 \\ 4.90^{(4)} \end{gathered}$ |  | $\begin{gathered} 7.10 \\ 5.50^{(4)} \end{gathered}$ | $\begin{array}{\|c\|} \hline 0.256 \\ 0.192^{(4)} \end{array}$ |  | $\begin{array}{\|c\|} \hline 0.279 \\ 0.216^{(4)} \end{array}$ |


| OUTLINE AND |
| :---: |
| MECHANICAL DATA |

"D and E1" do not include mold flash or protusions.
Mold flash or protusions shall not exceed 0.15 mm ( 0.006 ")
(2) No intrusion allowed inwards the leads.
(3) Flash or bleeds on exposed die pad shall not exceed 0.4 mm per side



Figure 22. PowerSSO36 mechanical data and package dimensions

| DIM. | mm |  |  | inch |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| A | 2.15 |  | 2.45 | 0.084 |  | 0.0965 |
| A2 | 2.15 |  | 2.35 | 0.084 |  | 0.0925 |
| a1 | 0 |  | 0.10 | 0 |  | 0.004 |
| b | 0.18 |  | 0.36 | 0.007 |  | 0.014 |
| c | 0.23 |  | 0.32 | 0.009 |  | 0.012 |
| D $^{(1)}$ | 10.10 |  | 10.50 | 0.398 |  | 0.413 |
| E $^{(1)}$ | 7.4 |  | 7.6 | 0.291 |  | 0.299 |
| e |  | 0.5 |  |  | 0.019 |  |
| e3 |  | 8.5 |  |  | 0.335 |  |
| F |  | 2.3 |  |  | 0.090 |  |
| G |  |  | 0.10 |  |  | 0.004 |
| H | 10.10 |  | 10.50 | 0.398 |  | 0.413 |
| h |  |  | 0.40 |  |  | 0.016 |
| k | $0^{\circ}$ |  | $8^{\circ}$ | $0^{\circ}$ |  | $8^{\circ}$ |
| L | 0.55 |  | 0.85 | 0.0217 |  | 0.0335 |
| M |  | 4.3 |  |  | 0.169 |  |
| N |  |  | $10^{\circ}$ |  |  | $10^{\circ}$ |
| O |  | 1.2 |  |  | 0.047 |  |
| Q |  | 0.8 |  |  | 0.031 |  |
| S |  | 2.9 |  |  | 0.114 |  |
| T |  | 3.65 |  |  | 0.144 |  |
| U |  | 1.0 |  |  | 0.039 |  |
| X |  | See variations table |  |  |  |  |
| Y |  |  |  |  |  |  |
| (1) "D" and "E" do not include mold flash or protrusions Mold flash |  |  |  |  |  |  |
| or protrusions shall not exceed 0.15 mm per side(0.006") |  |  |  |  |  |  |

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PowerSSO36 (slug-down)


## 8 Revision history

Table 36. Document revision history

| Date | Revision | Changes |
| :---: | :---: | :---: |
| 06-Jul-2012 | 1 | Initial release. |
| 13-Sep-2012 | 2 | Updated Features and Description on page 1. |
| 17-Dec 2012 | 3 | Updated Table 12: Power output switching times ( $8 \mathrm{~V}<\mathrm{VS}<18 \mathrm{~V}$ ): changed limits for DovoutHS/dt (2.8-8) V/ $\mu \mathrm{s}$. <br> Updated Table 13: Current feedback (CF): <br> changed LSL for CF voltage range to 0.03 V for $\mathrm{VS}>6.5 \mathrm{~V}$, OUTx $=0$ <br> $\mathrm{A}, \mathrm{TJ}=-40^{\circ} \mathrm{C}$; Current level $2,3,4$. <br> Updated Table 9, 12, 14, 16, 17 and 18: <br> added "Guaranteed through scan " for the following filters: $\mathrm{t}_{\mathrm{b}}$, $\mathrm{t}_{\text {trans }}$, $\mathrm{t}_{\mathrm{d} \text { dis }}, \mathrm{t}_{\mathrm{d} \text { pwon }}, \mathrm{t}_{\mathrm{d} \text { en }}, \mathrm{t}_{\text {FIL_OFF, }}, \mathrm{t}_{\text {FIL_ON }}, \mathrm{t}_{\text {delay retest }}, \mathrm{t}_{\text {DIAGOL }}, \mathrm{t}_{\text {DIAGOL1 }}$, tDIAGOL2. <br> Updated Table 9 added limits for //ABE input pulldown current in range 0 V < VABE < 1.5V . <br> Updated Table 20: Input pin disturbances, added " Not tested in production". |
| 31-May-2013 | 4 | Modified limits for Tdelay_retest [LSL $=290 \mu \mathrm{~s}, 410 \mu \mathrm{~s}]$ at page 18. <br> Modified bit \#4 and \#5 in DIA_REG at page 28. <br> Changed note for BIT6 in STATCON_REG register at page 35. <br> Changed in table 25 at pag. 31 addresses for RD_SPECIAL / WR_SPECIAL. <br> Added Section 6: Application circuit on page 38. <br> Updated Table 15 with Tracking limits between Overcurrent and Current limitation. <br> Modified pin names on Figure 3: PSSO36 pin connection (top view) and adde notes in the Table 3: L9959T (Two H-Bridge drivers in one package) pinout.. <br> Modified limit for parameter " $\mathrm{V}_{\mathrm{S}}$ current consumption in passive mode" to : $(0,2.5 \mathrm{~mA})$ in the Table 7: Supply. <br> New limits for Tdis parameter: ( $0,100 \mathrm{~ns}$ ) in the Table 21: Dynamic characteristics. <br> Modified Section 4.2: Protection and monitoring. <br> Modified Section 5.4.1: Reset sources and added new Section 5.4.2: Configuration registers reset sources. |

Table 36. Document revision history (continued)

| Date | Revision | Changes |
| :---: | :---: | :--- |
| 26-Jun-2013 | 5 | Corrected detailed pin names in function of die for psso36: Figure 3 <br> on page 7 and Table 3 on page 9. <br> Added LSL and USL in Table 12 for dVfOUTLS/dt Output fall slew- <br> rate low-side. Limts are set at $(2.5,8) \mathrm{V} / \mu \mathrm{s}$. <br> Table 13 modifed LSL at 10 mV for Current Feedback range for Vs > <br> $6.5 \mathrm{~V}, \mathrm{OUTx}=0 \mathrm{~A}, \mathrm{~T}_{\mathrm{J}}=-40^{\circ} \mathrm{C}$; Current level 2,3,4. <br> Modified RL_OL_MIN at 5Kohm in Table 16 on page 18'. <br> Renamed I_DISPU to R_DISPU and set new limits (10-45 Kohm) in <br> Table 20 on page 19. <br> Modifed hysteresis range in Figure 15 on page 25. <br> Corrected Figure 20: Application circuit on page 38. |
| 16-Sept-2013 | 6 | Updated Disclaimer |

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[^0]:    1. Not subject to production test; specified by design.
[^1]:    1. Not measured in production test. Parameter guaranteed by design.
