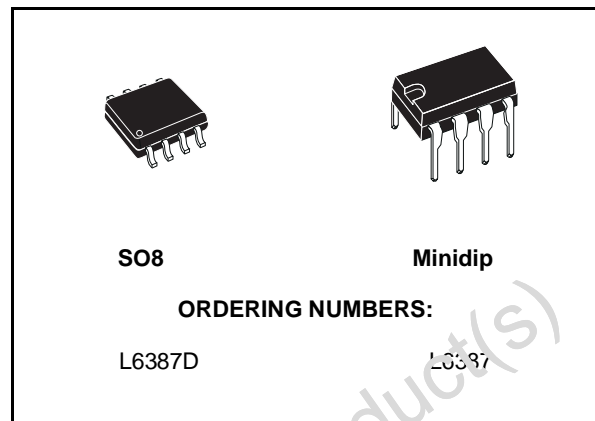


HIGH-VOLTAGE HIGH AND LOW SIDE DRIVER

- HIGH VOLTAGE RAIL UP TO 600 V
- dV/dt IMMUNITY ± 50 V/nsec IN FULL TEMPERATURE RANGE
- DRIVER CURRENT CAPABILITY:
400 mA SOURCE,
650 mA SINK
- SWITCHING TIMES 50/30 nsec RISE/FALL WITH 1nF LOAD
- CMOS/TTL SCHMITT TRIGGER INPUTS WITH HYSTERESIS AND PULL DOWN
- INTERNAL BOOTSTRAP DIODE
- OUTPUTS IN PHASE WITH INPUTS

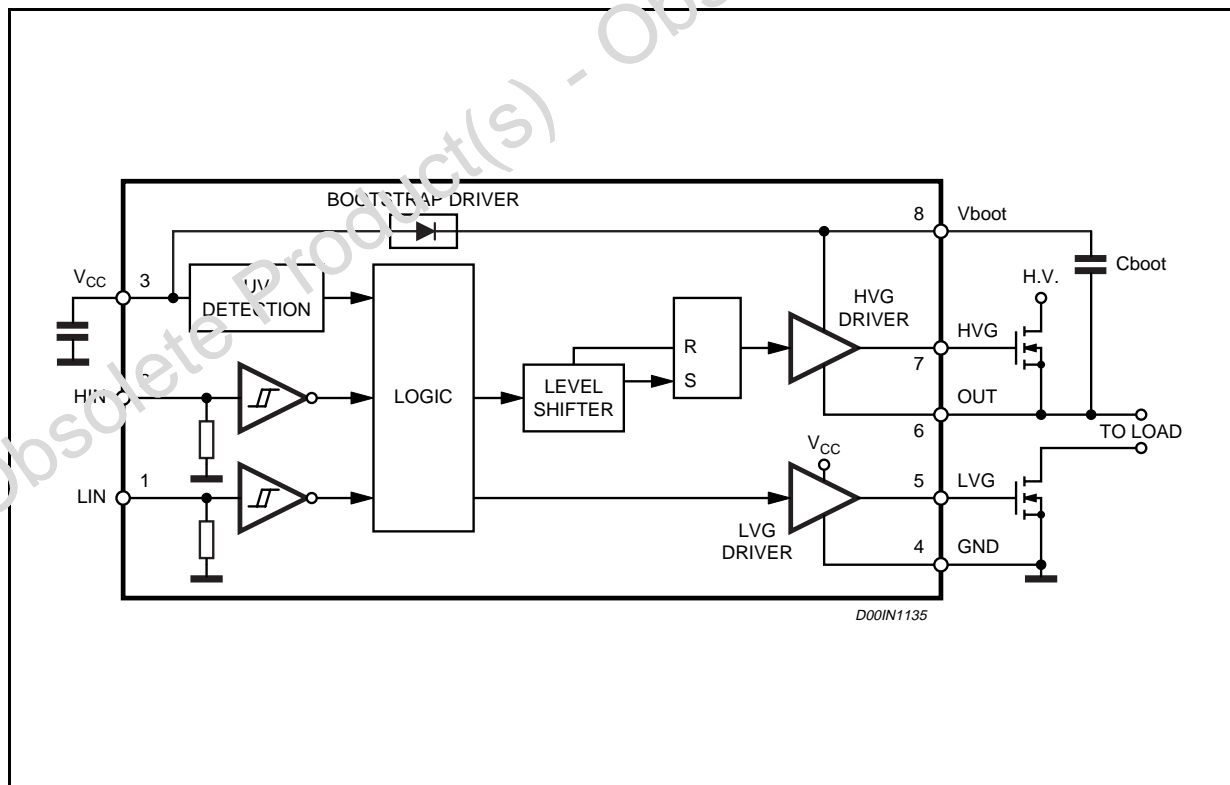


DESCRIPTION

The L6387 is an high-voltage device, manufactured with the BCD "OFF-LINE" technology. It has a Driver structure that enables to drive independent referenced N Channel Power MOS or

IGBT. The Upper (Floating) Section is enabled to work with voltage F_{all} up to 600V. The Logic Inputs are CMOS/TTL compatible for ease of interfacing with controlling devices.

BLOCK DIAGRAM

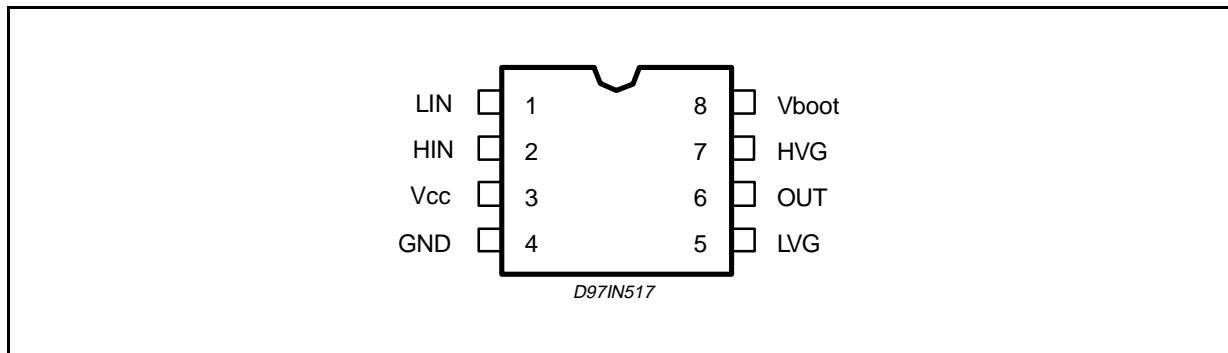


ABSOLUTE MAXIMUM RATINGS

| Symbol | Parameter | Value | Unit |
|----------|--------------------------------------|------------------|------|
| Vout | Output Voltage | -3 to Vboot - 18 | V |
| Vcc | Supply Voltage | - 0.3 to +18 | V |
| Vboot | Floating Supply Voltage | - 1 to 618 | V |
| Vhvg | Upper Gate Output Voltage | - 1 to Vboot | V |
| Vlvg | Lower Gate Output Voltage | -0.3 to Vcc +0.3 | V |
| Vi | Logic Input Voltage | -0.3 to Vcc +0.3 | V |
| dVout/dt | Allowed Output Slew Rate | 50 | V/ns |
| Ptot | Total Power Dissipation (Tj = 85 °C) | 750 | mW |
| Tj | Junction Temperature | 150 | °C |
| Ts | Storage Temperature | -50 to 150 | °C |

Note: ESD immunity for pins 6, 7 and 8 is guaranteed up to 900V (Human Body Model)

PIN CONNECTION



THERMAL DATA

| Symbol | Parameter | SO8 | Minidip | Unit |
|-----------------------|--|-----|---------|------|
| R _{th j-amb} | Thermal Resistance Junction to Ambient | 150 | 100 | °C/W |

PIN DESCRIPTION

| N. | Name | Type | Function |
|----|---------|------|---------------------------------|
| 1 | LIN | I | Lower Driver Logic Input |
| 2 | HIN | I | Upper Driver Logic Input |
| 3 | Vcc | I | Low Voltage Power Supply |
| 4 | GND | | Ground |
| 5 | LVG (*) | O | Low Side Driver Output |
| 6 | VOUT | O | Upper Driver Floating Reference |
| 7 | HVG (*) | O | High Side Driver Output |
| 8 | Vboot | | Bootstrap Supply Voltage |

(*) The circuit guarantees 0.3V maximum on the pin (@ I_{sink} = 10mA). This allows to omit the "bleeder" resistor connected between the gate and the source of the external MOSFET normally used to hold the pin low.

RECOMMENDED OPERATING CONDITIONS

| Symbol | Pin | Parameter | Test Condition | Min. | Typ. | Max. | Unit |
|------------|-----|-------------------------|-----------------------|--------|------|------|------|
| Vout | 6 | Output Voltage | | Note 1 | | 580 | V |
| Vboot-Vout | 8 | Floating Supply Voltage | | Note 1 | | 17 | V |
| fsw | | Switching Frequency | HVG,LVG load CL = 1nF | | | 400 | kHz |
| Vcc | 2 | Supply Voltage | | | | 17 | V |
| Tj | | Junction Temperature | | -45 | | 125 | °C |

Note 1: If the condition Vboot - Vout < 18V is guaranteed, Vout can range from -3 to 580V.

ELECTRICAL CHARACTERISTICS
AC Operation (Vcc = 15V; Tj = 25°C)

| Symbol | Pin | Parameter | Test Condition | Min. | Typ. | Max. | Unit |
|--------|--------|---|----------------|------|------|------|------|
| ton | 1 vs 7 | High/Low Side Driver Turn-On Propagation Delay | Vout = 0V | | 110 | | ns |
| toff | 2 vs 5 | High/Low Side Driver Turn-Off Propagation Delay | Vout = 600V | | 105 | | ns |
| tr | 7,5 | Rise Time | CL = 1000pF | | 50 | | ns |
| tf | 7,5 | Fall Time | CL = 1000pF | | 30 | | ns |

DC OPERATION (Vcc = 15V; Tj = 25°C)

| Symbol | Pin | Parameter | Test Condition | Min. | Typ. | Max. | Unit |
|--|-----|---------------------------------------|-----------------------|------|------|------|------|
| Low Supply Voltage Section | | | | | | | |
| Vcc | 3 | Supply Voltage | | | | 17 | V |
| Vccth1 | | Vcc UV Turn On Threshold | | 5.5 | 6 | 6.5 | V |
| Vccth2 | | Vcc UV Turn Off Threshold | | 5 | 5.5 | 6 | V |
| Vcchys | | Vcc UV Hysteresis | | | 0.5 | | V |
| Iqccu | | Undervoltage Quiescent Supply Current | Vcc ≤ 9V | | 150 | 220 | µA |
| Iqcc | | Quiescent Current | Vcc = 15V | | 250 | 320 | µA |
| Rdson | | Bootstrap Driver on Resistance (*) | Vcc ≥ 12.5V | | 125 | | Ω |
| Bootstrapped supply Voltage Section | | | | | | | |
| VBS | 8 | Bootstrap Supply Voltage | | | | 17 | V |
| IQBS | | VBS Quiescent Current | HVG ON | | | 200 | µA |
| ILK | | High Voltage Leakage Current | VS = VB = 600V | | | 10 | µA |
| High/Low Side Driver | | | | | | | |
| Iso | 5,7 | Source Short Circuit Current | VIN = Vih (tp < 10µs) | 300 | 400 | | mA |
| Isi | | Sink Short Circuit Current | VIN = Vil (tp < 10µs) | 450 | 650 | | mA |
| Logic Inputs | | | | | | | |
| Vil | 2,3 | Low Level Logic Threshold Voltage | | | | 1.5 | V |
| Vih | | High Level Logic Threshold Voltage | | 3.6 | | | V |
| Iih | | High Level Logic Input Current | VIN = 15V | | 50 | 70 | µA |
| Iil | | Low Level Logic Input Current | VIN = 0V | | | 1 | µA |

(*) Rdson is tested in the following way: $R_{DSON} = \frac{(V_{CC} - V_{CBOOT1}) - (V_{CC} - V_{CBOOT2})}{I_1(V_{CC}, V_{CBOOT1}) - I_2(V_{CC}, V_{CBOOT2})}$

where I1 is pin 8 current when Vcboot = Vcboot1, I2 when Vcboot = Vcboot2.

Figure 1. Typical Rise and Fall Times vs. Load Capacitance

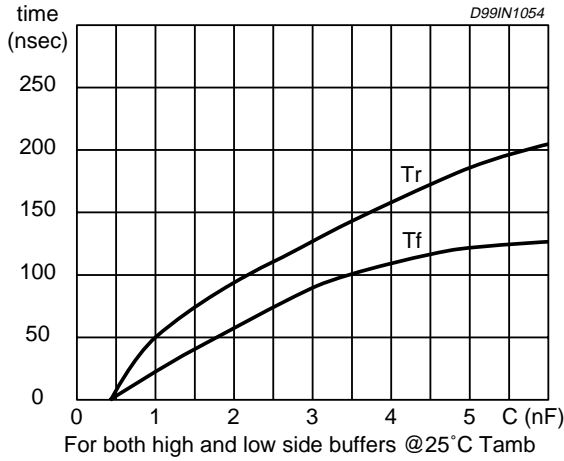
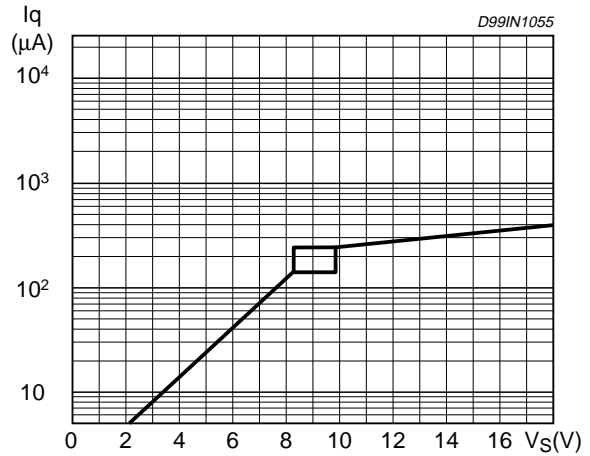


Figure 2. Quiescent Current vs. Supply Voltage



Input Logic

L6387 Input Logic is V_{CC} (17V) compatible. An interlocking features is offered (see truth table below) to avoid undesired simultaneous turn ON of both Power Switches driven.

Table 1.

| | | | | | |
|--------|-----|---|---|---|---|
| Input | HIN | 0 | 0 | 1 | 1 |
| | LIN | 0 | 1 | 0 | 1 |
| Output | HVG | 0 | 0 | 1 | 0 |
| | LVG | 0 | 1 | 0 | 0 |

BOOTSTRAP DRIVER

A bootstrap circuitry is needed to supply the high voltage section. This function is normally accomplished by a high voltage fast recovery diode (fig. 3a). In the L6387 a patented integrated structure replaces the external diode. It is realized by a high voltage DMOS, driven synchronously with the low side driver (LVG), with in series a diode, as shown in fig. 3b

An internal charge pump (fig. 3b) provides the DMOS driving voltage .

The diode connected in series to the DMOS has been added to avoid undesirable turn on of it.

CBOOT selection and charging:

To choose the proper C_{BOOT} value the external MOS can be seen as an equivalent capacitor. This capacitor C_{EXT} is related to the MOS total gate charge :

$$C_{EXT} = \frac{Q_{gate}}{V_{gate}}$$

The ratio between the capacitors C_{EXT} and C_{BOOT} is proportional to the cyclical voltage loss .

It has to be:

$$C_{BOOT} \gg \gg C_{EXT}$$

e.g.: if Q_{gate} is 30nC and V_{gate} is 10V, C_{EXT} is 3nF. With C_{BOOT} = 100nF the drop would be 300mV.

If HVG has to be supplied for a long time, the C_{BOOT} selection has to take into account also the leakage losses.

e.g.: HVG steady state consumption is lower than 200µA, so if HVG T_{ON} is 5ms, C_{BOOT} has to supply 1µC to C_{EXT}. This charge on a 1µF capacitor means a voltage drop of 1V.

The internal bootstrap driver gives great advantages: the external fast recovery diode can be avoided (it usually has great leakage current). This structure can work only if V_{OUT} is close to GND (or lower) and in the meanwhile the LVG is on. The charging time (T_{charge}) of the C_{BOOT} is the time in which both conditions are fulfilled and it has to be long enough to charge the capacitor.

The bootstrap driver introduces a voltage drop due to the DMOS R_{DS(on)} (typical value: 125 Ohm). At low frequency this drop can be neglected. Anyway increasing the frequency it must be taken in to account.

The following equation is useful to compute the



drop on the bootstrap DMOS:

$$V_{\text{drop}} = I_{\text{charge}} R_{\text{dson}} \rightarrow V_{\text{drop}} = \frac{Q_{\text{gate}}}{T_{\text{charge}}} R_{\text{dson}}$$

where Q_{gate} is the gate charge of the external power MOS, R_{dson} is the on resistance of the bootstrap DMOS, and T_{charge} is the charging time of the bootstrap capacitor.

For example: using a power MOS with a total gate charge of 30nC the drop on the bootstrap

DMOS is about 1V, if the T_{charge} is 5μs. In fact:

$$V_{\text{drop}} = \frac{30\text{nC}}{5\mu\text{s}} \cdot 125\Omega \sim 0.8\text{V}$$

V_{drop} has to be taken into account when the voltage drop on C_{BOOT} is calculated: if this drop is too high, or the circuit topology doesn't allow a sufficient charging time, an external diode can be used.

Figure 3. Bootstrap Driver.

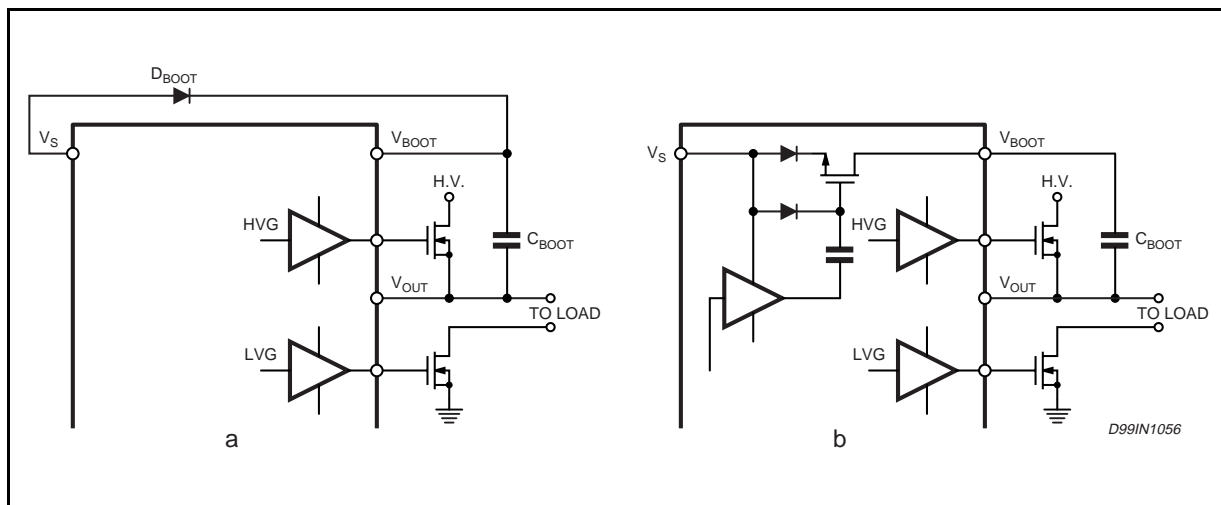


Figure 4. Turn On Time vs. Temperature

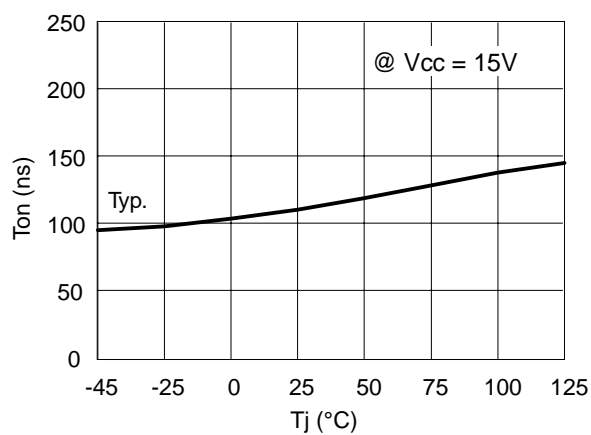


Figure 5. Turn Off Time vs. Temperature

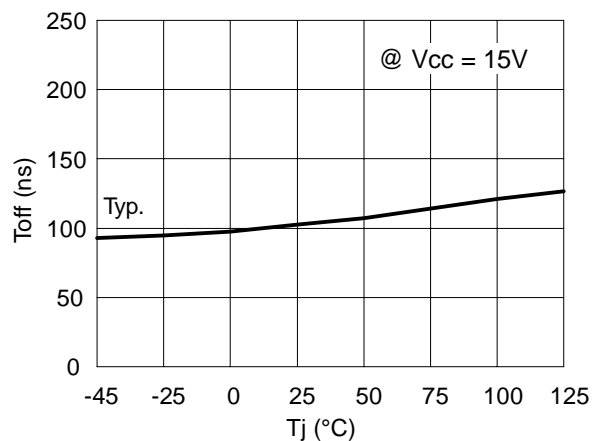
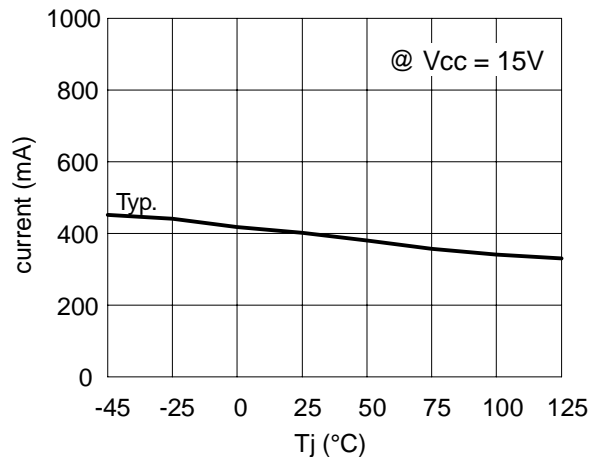
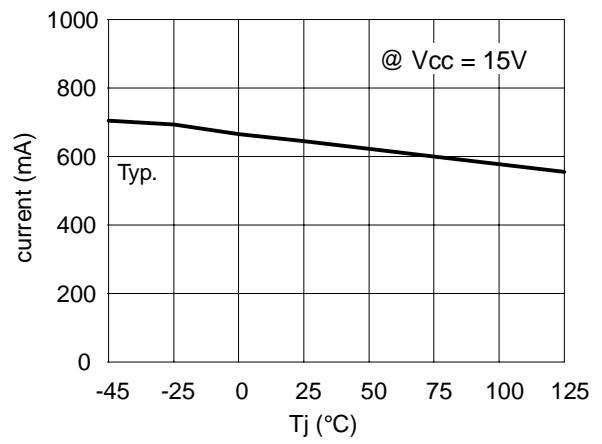
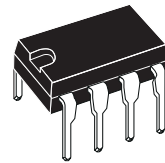


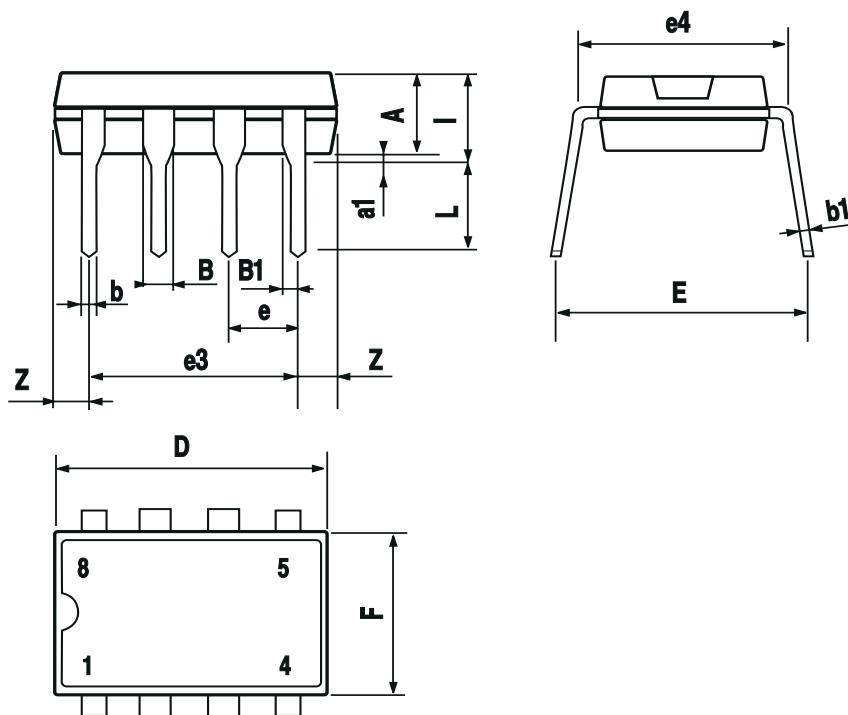
Figure 6. Output Source Current vs. Temperature**Figure 7. Output Sink Current vs. Temperature**

| DIM. | mm | | | inch | | |
|------|-------|------|-------|-------|-------|-------|
| | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| A | | 3.32 | | | 0.131 | |
| a1 | 0.51 | | | 0.020 | | |
| B | 1.15 | | 1.65 | 0.045 | | 0.065 |
| b | 0.356 | | 0.55 | 0.014 | | 0.022 |
| b1 | 0.204 | | 0.304 | 0.008 | | 0.012 |
| D | | | 10.92 | | | 0.430 |
| E | 7.95 | | 9.75 | 0.313 | | 0.384 |
| e | | 2.54 | | | 0.100 | |
| e3 | | 7.62 | | | 0.300 | |
| e4 | | 7.62 | | | 0.300 | |
| F | | | 6.6 | | | 0.260 |
| I | | | 5.08 | | | 0.200 |
| L | 3.18 | | 3.81 | 0.125 | | 0.150 |
| Z | | | 1.52 | | | 0.060 |

OUTLINE AND MECHANICAL DATA

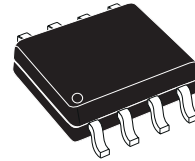


Minidip



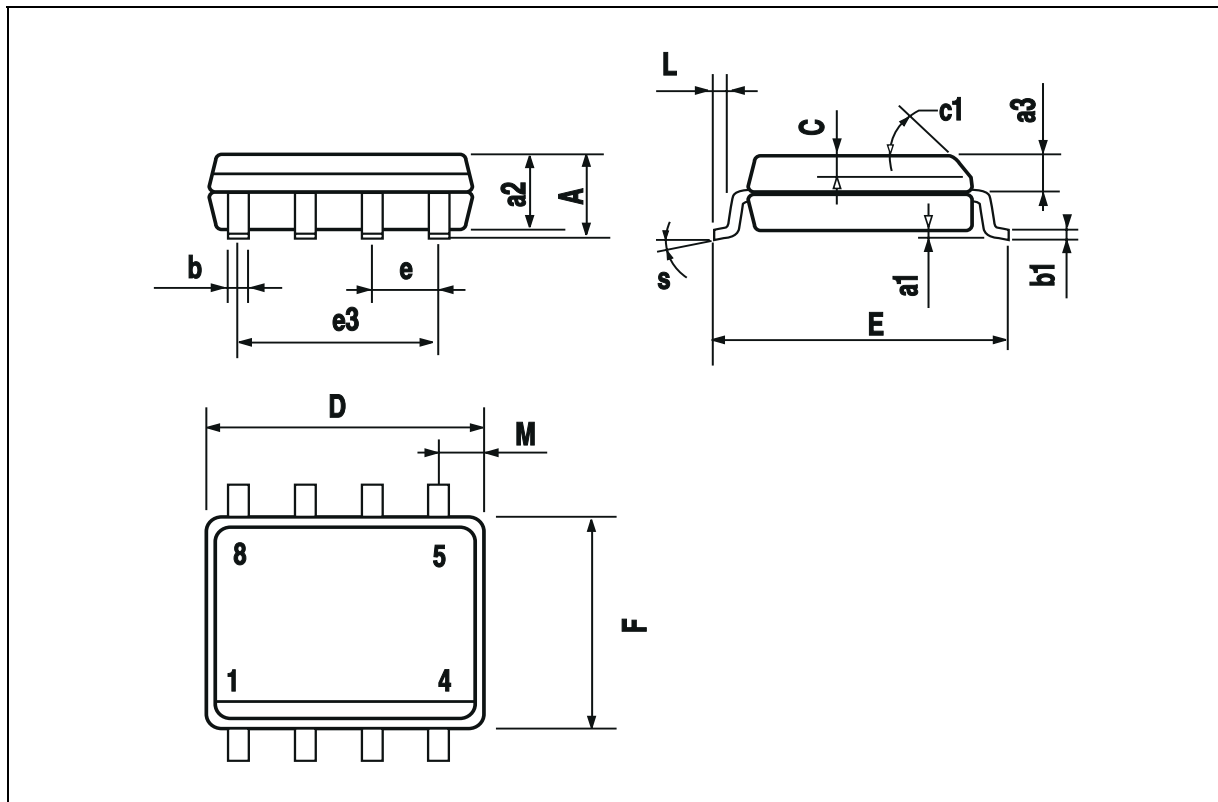
| DIM. | mm | | | inch | | |
|-------|------------|------|------|-------|-------|-------|
| | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| A | | | 1.75 | | | 0.069 |
| a1 | 0.1 | | 0.25 | 0.004 | | 0.010 |
| a2 | | | 1.65 | | | 0.065 |
| a3 | 0.65 | | 0.85 | 0.026 | | 0.033 |
| b | 0.35 | | 0.48 | 0.014 | | 0.019 |
| b1 | 0.19 | | 0.25 | 0.007 | | 0.010 |
| C | 0.25 | | 0.5 | 0.010 | | 0.020 |
| c1 | 45° (typ.) | | | | | |
| D (1) | 4.8 | | 5.0 | 0.189 | | 0.197 |
| E | 5.8 | | 6.2 | 0.228 | | 0.244 |
| e | | 1.27 | | | 0.050 | |
| e3 | | 3.81 | | | 0.150 | |
| F (1) | 3.8 | | 4.0 | 0.15 | | 0.157 |
| L | 0.4 | | 1.27 | 0.016 | | 0.050 |
| M | | | 0.6 | | | 0.024 |
| S | 8° (max.) | | | | | |

OUTLINE AND MECHANICAL DATA



SO8

(1) D and F do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm (.006inch).



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