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Linear CCD Image Sensor

Description

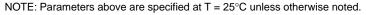
The KLI–2104 Image Sensor is a high dynamic range, multi-spectral, linear solid-state image sensor designed for demanding color scanning applications.

The KLI–2104 contains three parallel linear photodiode arrays, each with 2098 active photosites for the detection of red, green, and blue (R, G, B) signals. A fourth channel, comprised of 4,196 pixels, provides high resolution luminance information. This combination allows the KLI–2104 to provide high resolution scans with accurate color reproduction.

The device offers high sensitivity, low noise, and negligible lag.

Table 1. GENEINAL SPECIFICATIONS						
Parameter	Typical Value					
Architecture	Quadri-Linear CCD					
Total Number of Pixels Chroma Luma	3 × 2222 1 × 4244					
Number of Active Pixels Chroma Luma	3 × 2098 1 × 4196					
Pixel Size Chroma Luma	14 μm 7 μm					
Inter-Array Spacing G to R, R to B B to L	84 μm 87.5 μm					
Active Image Size	29.4 mm (Diagonal)					
Chip Size	35.64 mm (H) × 1.06 mm (V)					
Saturation Signal Chroma Luma	208,000 e ⁻ 140,000 e ⁻					
Output Sensitivity	12 μV/e ⁻					
Peak Quantum Efficiency R; G; B; L	73%; 55%; 62%; 88%					
Responsivity R; G; B; L	33; 36; 56; 16 V/μJ/cm ²					
Total Read Noise	30 e⁻					
Dark Current Chroma Luma	0.22 pA/Pixel 0.07 pA/Pixel					
Dynamic Range Chroma Luma	80 dB 75 dB					
Charge Transfer Efficiency	0.99999					
Photoresponse Non-Uniformity	15% Peak-Peak					
Operating Frequency	20 MHz per Output					
Package	CERDIP					
Cover Glass Options	MAR Coated, 2 Sides					

Table 1. GENERAL SPECIFICATIONS





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Figure 1. KLI–2104 Linear CCD Image Sensor

Features

- Quadri-Linear Color Array Design (G, R, B, L) for High Resolution with Accurate Color Reproduction
- High Sensitivity Photosites
- Low Noise Design with Negligible Image Lag
- Pixel-Summing Support for Extended Sensitivity and Dynamic Range
- 5.0 V Clock Inputs with Two-Phase Register Clocking
- Choice of Multi-Layer Anti-Reflective Coated (MAR) or Clear Coverglass

Applications

- Digitization
- Photography

ORDERING INFORMATION

See detailed ordering and shipping information on page 2 of this data sheet.

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ORDERING INFORMATION

Table 2. ORDERING INFORMATION – KLI–2104 IMAGE SENSOR

Part Number	Description	Marking Code
KLI-2104-DAA-EB-AA	Color (RGB), No Microlens, CERDIP Package (Leadframe), Clear Cover Glass (No Coatings), Standard Grade	
KLI-2104-DAA-EB-AE	Color (RGB), No Microlens, CERDIP Package (Leadframe), Clear Cover Glass (No Coatings), Engineering Grade	KLI–2104 Lot Number
KLI-2104-DAA-ED-AA	Color (RGB), No Microlens, CERDIP Package (Leadframe), Clear Cover Glass with AR Coating (Both Sides), Standard Grade	Serial Number
KLI-2104-DAA-ED-AE	Color (RGB), No Microlens, CERDIP Package (Leadframe), Clear Cover Glass with AR Coating (Both Sides), Engineering Grade	

See the ON Semiconductor *Device Nomenclature* document (TND310/D) for a full description of the naming convention used for image sensors. For reference documentation, including information on evaluation kits, please visit our web site at <u>www.onsemi.com</u>.

DEVICE DESCRIPTION

Architecture

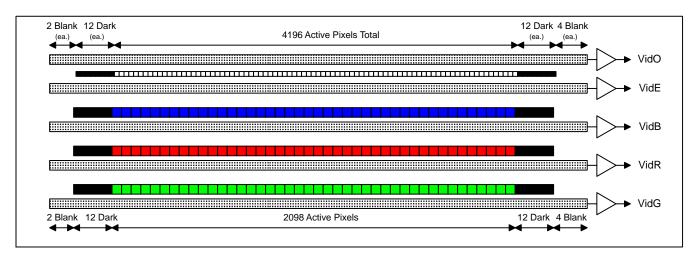
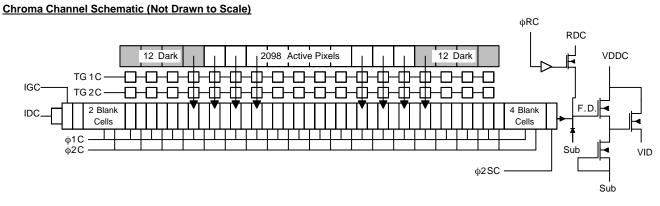
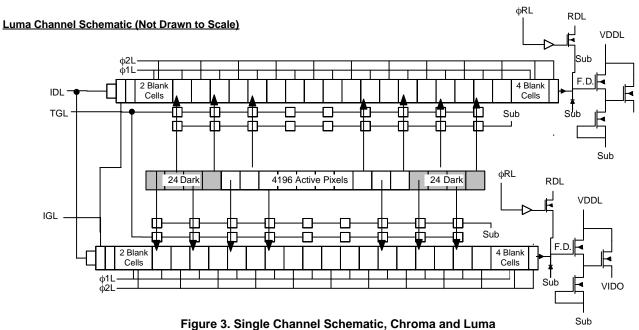


Figure 2. Block Diagram





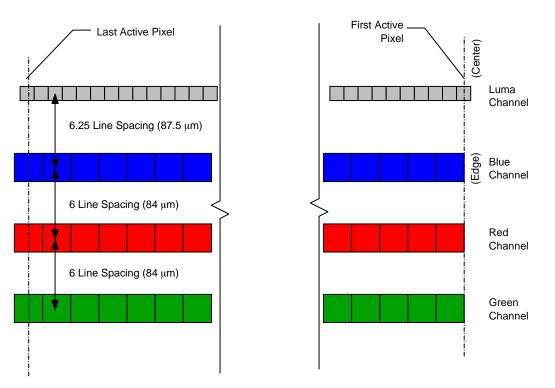


Figure 4. Active Pixel and Channel Alignment – KLI–2104

Imaging

During the integration period, an image is obtained by gathering electrons generated by photons incident upon the photodiodes. The charge collected in the photodiode array is a linear function of the local exposure. The charge is stored in the photodiode itself and is isolated from the CCD shift registers during the integration period by the transfer gates TG1 and TG2 for the chroma channels, which are held at a barrier potential. (The luminance channel has only one transfer gate, TG). At the end of the integration period, the CCD register clocking is stopped with the $\phi 1$ and $\phi 2$ gates being held in a 'high' and 'low' state respectively. Next, the TG gates are turned 'on' causing the charge to drain from the photodiode into the TG1 storage region. As TG1 is turned back 'off' charge is transferred through TG2 and into the $\varphi 1$ storage region. The TG2 gate is then turned 'off', isolating the shift registers from the accumulation region once again. For the luminance channel, only one TG transfer is required. Complementary clocking of the $\phi 1$ and φ2 phases now resumes for readout of the current line of data while the next line of data is integrated.

Charge Transport and Sensing

Readout of the signal charge is accomplished by two-phase, complementary clocking of the $\phi 1$ and $\phi 2$ gates. The register architecture has been designed for high speed clocking with minimal transport and output signal degradation, while still maintaining low (5 Vp-p min) clock swings for reduced power dissipation, lower clock noise and simpler driver design. The data in all registers is clocked simultaneously toward the output structures. The signal is then transferred to the output structures in a parallel format at the falling edge of the ϕ 2 clock. Re-settable floating diffusions are used for the charge-to-voltage conversion while source followers provide buffering to external connections. The potential change on the floating diffusion is dependent on the amount of signal charge and is given by $\Delta V_{FD} = \Delta Q/C_{FD}$, where ΔV_{FD} is the change in potential on the floating diffusion, ΔQ is the amount of charge, and C_{FD} is the capacitance of the floating diffusion node. Prior to each pixel output, the floating diffusion is returned to the RD level by the reset clock, ϕ R.

Pixel Summing (Chroma Channels Only)

Enabling the pixel – summing feature can vary the effective resolution of the color channels of this sensor. A separate pin is provided for the last shift register gate labeled φ 2SC. This gate, when clocked appropriately, stores the summation of signal from adjacent pixels. This combined charge packet is then transferred onto the sense node. As an example, the sensor can be operated in 2-pixel summing mode (1,049 pixels), by supplying a clock to φ 2SC which is a 75% duty cycle signal at 1/2 the frequency of the φ 2C signal, and modifying the φ RC clock as depicted in Figure 25. Applications that require full resolution mode (2,098 pixels), must tie the φ 2SC pin to the φ 2C pin. Refer to Figure 24 for additional details.

The luma channel outputs are in an odd and even configuration. The odd pixel value and the even pixel value are available simultaneously during the $\phi 2$ clock low phase. In this manner, pixel summing is an option off-chip.

Physical Description

Pin Description and Device Orientation

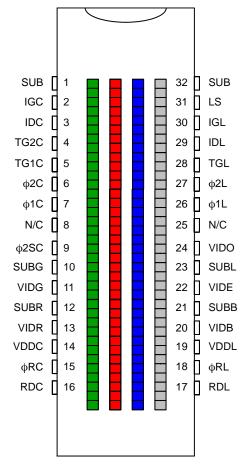


Figure 5. Pinout Diagram

Table 3. PACKAGE PIN DESCRIPTION

Pin	Name	Description
1	SUB	Substrate/Ground
2	IGC	Test Input – Input Diode, Chroma
3	IDC	Test Input – Input Diode, Chroma
4	TG2C	Transfer Gate 2 Clock, Chroma
5	TG1C	Transfer Gate 1 Clock, Chroma
6	φ2C	Phase 2 CCD Clock, Chroma
7	φ1C	Phase 1 CCD Clock, Chroma
8	N/C	No Connection (Ground)
9	H2SC	Phase 2 Summing Gate, Chroma
10	SUBx	Ground Reference (R, G, B)
11	VIDx	Output Video (R, G, B)
12	SUBx	Ground Reference (R, G, B)
13	VIDx	Output Video (R, G, B)
14	VDDC	Amplifier Supply (Chroma)
15	φRC	Reset Clock, Chroma
16	RDC	Reset Drain, Chroma

Pin	Name	Description
17	RDL	Reset Drain, Luma
18	φRL	Reset Clock, Luma
19	VDDL	Amplifier Supply (Luma)
20	VIDx	Output Video (R, G, B)
21	SUBx	Ground Reference (R, G, B)
22	VIDE	Output Video (Luma Even Channel)
23	SUBL	Ground Reference (Luma)
24	VIDO	Output Video (Luma Odd Channel)
25	N/C	No Connection (Ground)
26	φ1L	Phase 1 CCD Clock, Luma
27	φ2L	Phase 2 CCD Clock, Luma
28	TGL	Transfer Gate Clock, Luma
29	IDL	Test Input – Input Diode, Luma
30	IGL	Test Input – Input Gate, Luma
31	LS	Light Shield/Exposure Drain
32	SUB	Substrate/Ground

IMAGING PERFORMANCE

Specifications given under nominally specified operating conditions for the given mode of operation at 25° C, $f_{CLK} = 1$ MHz, 2.1 ms integration time, MAR cover glass, color filters, and an active load as in the schematic shown in Figure 23 of a typical output bias/buffer circuit, unless otherwise specified. See notes on next page for further descriptions.

Each limit identified as a maximum and/or a minimum is tested and guaranteed for every device. Nominal values are to be considered typical performance values that are design and manufacturing targets. These values are not guaranteed.

Table 4. SPECIFICATIONS

Description	Symbol	Min.	Nom.	Max.	Units	Notes	Verification Plan ¹⁵
Saturation Output Voltage, Chroma	V _{SAT} , Chroma	2.0	2.5	-	Vp-р	1, 9	Die
Saturation Output Voltage, Luminance	V _{SAT} , Luma	1.2	1.75	_	Vр-р	1, 9	Die
Output Sensitivity	$\Delta V_{OUT} / \Delta N_e$	-	12	-	μV/e⁻		Design
Saturation Signal Charge, Chroma	Ne,sat chroma	-	208,000	-	e−		Design
Saturation Signal Charge, Luminance	N _{e,SAT} Luma	-	146,000	-	e-		Design
Responsivity	R, Chroma					2, 9, 10	
Quantum Efficiency Blue Channel @ 460 nm Green Channel @ 540 nm Red Channel @ 650 nm Luma Channel @ 550 nm	QE, Chroma QE, Luma	- - - -	73 55 62 88		%	2, 9, 10 ±10% ±10% ±10% ±10%	Design
Dynamic Range Chroma Luma	DR, Chroma DR, Luma		80 75		dB	3	Design
Dark Noise, Chroma and Luma	Noise, Dark	_	30	-	e-		Design
Dark Signal Non-Uniformity, Chroma Luma	DSNU, Chroma DSNU, Luma		2 2	16 16	mV p-p	14	Die
Dark Current Chroma Luma	I _{DARK} , Chroma I _{DARK} , Luma		0.22 0.07	0.5 0.2	pA/Pixel	4	Die
Charge Transfer Efficiency Chroma Luma	CTE, Chroma CTE, Luma	0.999995 0.999995	0.999998 0.999998	1 1	-	5	Die
Lag Chroma Luma	L, Chroma L, Luma		0.05 0.1	1 1	%	1 st Field	Die
DC Output Offset	V _{ODC}	5	6.6	8	V	9	Die
Photoresponse Non-Uniformity, Low Frequency, Chroma	PRNUC, Low	-	6	20	% р-р	6	Die
Photoresponse Non-Uniformity, Medium Frequency, Chroma	PRNUC, Med	-	6	20	% р-р	7	Die
Photoresponse Non-Uniformity, High Frequency, Chroma	PRNUC, High	-	3	15	%	8	Die
Photoresponse Non-Uniformity, Low Frequency, Luma	PRNUL, Low	-	6	20	% р-р	6	Die
Photoresponse Non-Uniformity, Medium Frequency, Luma	PRNUL, Med	-	6	20	% р-р	7	Die
Photoresponse Non-Uniformity, High Frequency, Luma	PRNUL, High	-	3	15	%	8	Die
Darkfield Defect, Brightpoint	Dark Def	_	-	0	Allowed	12	Die

Table 4. SPECIFICATIONS (continued)

Description	Symbol	Min.	Nom.	Max.	Units	Notes	Verification Plan ¹⁵
Brightfield Defect, Dark or Bright	Bfld Def	-	-	0	Allowed	13	Die
Smear, Photodiode to CCD Crosstalk Blue Channel @ 450 nm Green Channel @ 550 nm Red Channel @ 650 nm	Smear, Chroma	- - -	0.2 0.05 0.4	- - -	%		Design
Smear Luma channel @ 550 nm	Smear, Luma	_	1.2	_	%		Design
Linearity, Maximum from Best Fit Straight Line Blue Channel Green Channel Red Channel Luma Channel	Linearity, Chroma Linearity, Luma	- - - -	0.6 1.2 1 1	- - - -	%		Design
DC Amplifier Gain	Gain, DC	-	0.75	-			Design
Amplifier Output Resistance	R _{OUT}	-	220	-	Ω		Design
Output Buffer Bandwidth	f _{-3dB}	-	72	-	MHz		Design

1. Calculated under a flat field illumination. Defined as the maximum output level achievable before linearity or PRNU performance is degraded beyond specification.

With color filter. Values specified at filter peaks. 50% bandwidth = ±30 nm. Color filter arrays become transparent after 710 nm. It is
recommended that a suitable IR cut filter be used to maintain spectral balance and optimal MTF. See quantum efficiency plots in Figure 7.

3. This device utilizes 2-phase clocking for cancellation of driver displacement currents. Symmetry between φ1 and φ2 phases must be maintained to minimize clock noise.

4. Dark current doubles approximately every +7°C.

Measured per transfer, 2 phases per pixel. For the typical total line (Chroma): (0.99999)⁴²⁵⁶ = 0.9583. For the typical total line (Luma): (0.99999)⁴²⁵⁶ = 0.9583. It should be noted that this parameter degrades with increasing horizontal clock frequency.

6. Low frequency response is measured across the entire array with a 1,000 pixel-moving window and a 5 pixel median filter evaluated under a flat field illumination.

7. Medium frequency response is measured across the entire array with a 50 pixel-moving window and a 5 pixel median filter evaluated under a flat field illumination.

8. High frequency response non-uniformity represents individual pixel defects evaluated under a flat field illumination. An individual pixel value may deviate above or below the average response for the entire array by a certain threshold.

9. Increasing the current load (nominally 6 mA) to improve signal bandwidth will decrease these parameters.

10. If resistive loads are used to set current, the amplifier gain will be reduced, thereby reducing the output sensitivity and net responsivity.

11. Where defective pixels are allowed, they will be separated by at least one non-defective pixel within and across channels.

12. Pixels whose response is greater than the average response by the specified threshold, (16 mV). See Figure 6.

13. Pixels, whose response is greater or less than the average response by the specified threshold, contained in the high frequency PRNU specification for that channel. See Figure 6.

14. Absolute difference between the maximum and minimum average signal level for an entire video channel.

15. A "die" parameter is measured on every sensor during production testing. A "design" parameter is quantified during design verification and not guaranteed by specification.

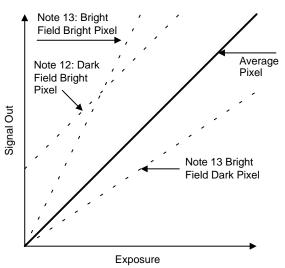
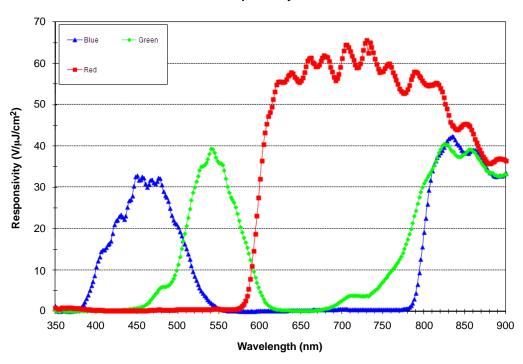


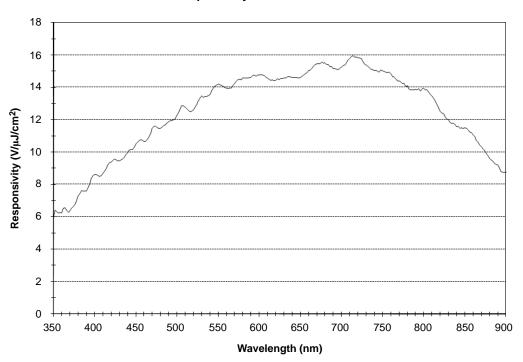
Figure 6. Defective Pixel Classification

TYPICAL PERFORMANCE MEASURES

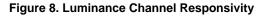


KLI-2104 Responsivity - Color Channels

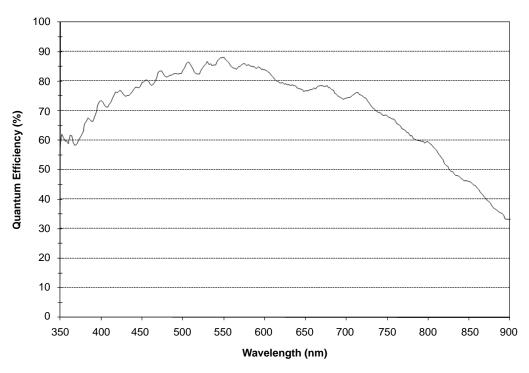




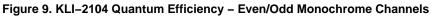
KLI-2104 Responsivity - Even/Odd Monochrome Channels

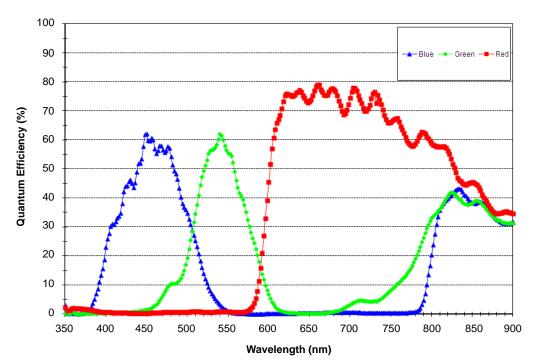






KLI-2104 Quantum Efficiency – Even/Odd Monochrome Channels

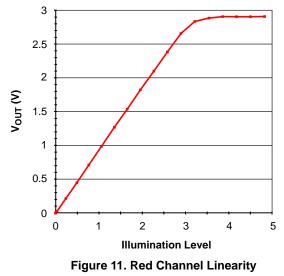




KLI-2104 Quantum Efficiency - Color Channels

Figure 10. KLI–2104 Quantum Efficiency – Color Channels





GREEN Channel Linearity, Typical Response – Green LED Illumination

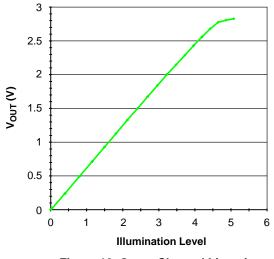
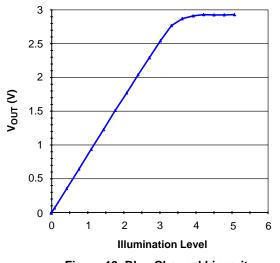


Figure 12. Green Channel Linearity





Dark Noise vs. Temperature Typical Performance – 1 MHz Data Rate

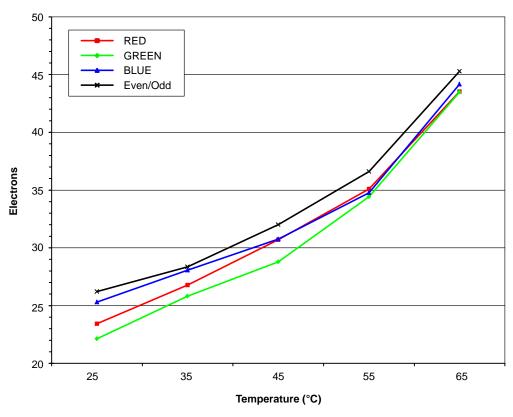
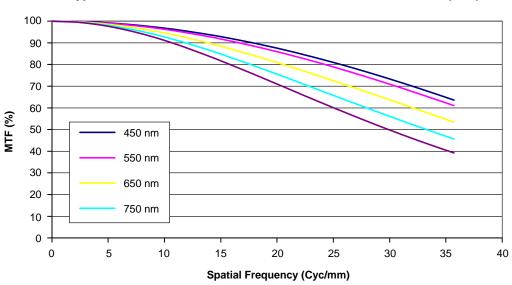
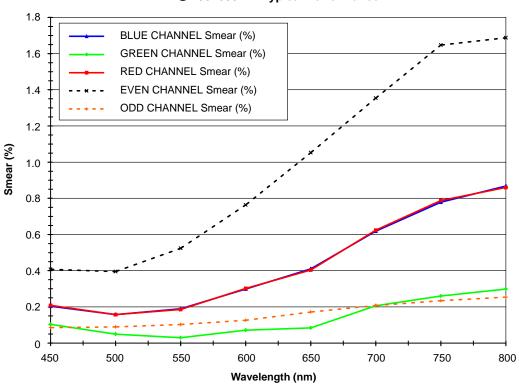


Figure 14. Dark Noise vs. Temperature

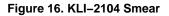


Typical Modulation Transfer Function KLI–2104 Chroma Channels (MTF)

Figure 15. Typical Modulation Transfer



KLI-2104 Smear (Photodiode-to-CCD Crosstalk) @ 450-800 nm Typical Performance





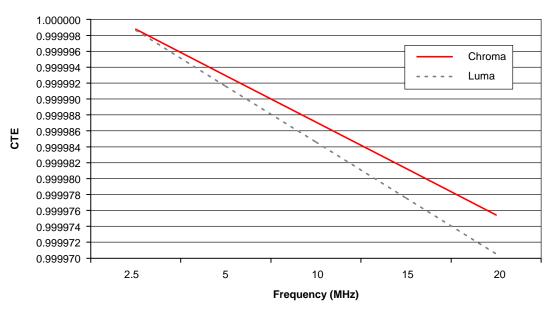
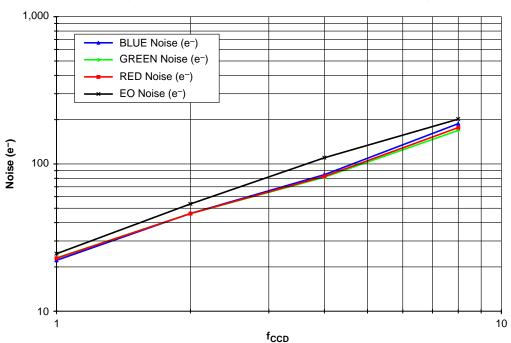
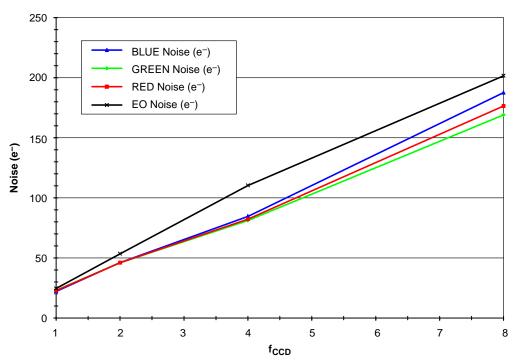


Figure 17. CTE vs. Frequency



Typical KLI-2104 Dark Noise vs. CCD Clock Frequency

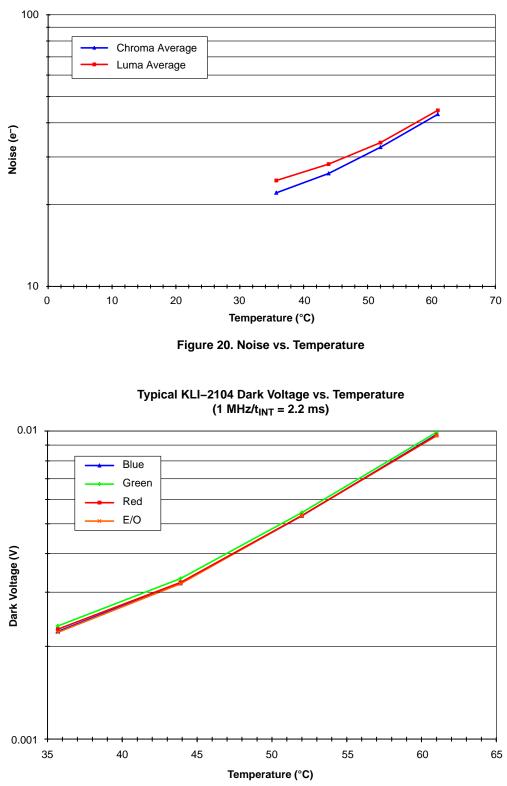




Typical KLI-2104 Dark Noise vs. CCD Clock Frequency

Figure 19. Typical KLI–2104 Dark Noise vs. CCD Clock Frequency

Typical KLI–2104 Noise vs. Temperature (1 MHz)





OPERATION

Table 5. ABSOLUTE MAXIMUM RATINGS

Description	Symbol	Minimum	Maximum	Units	Notes
Gate Pin Voltage	V _{GATE}	0	16	V	1, 2
Pin-to-Pin Voltage	V _{PIN-PIN}	-	16	V	1, 3
Diode Pin Voltages	V _{DIODE}	-0.5	16	V	1, 4
Output Bias Current	I _{DD}	-10	-1	mA	5
Output Load Capacitance	C _{VID,LOAD}	-	10	pF	9
CCD Clocking Frequency	f _{CLK}	-	20	MHz	6
Operating Temperature	T _{OP}	0	70	°C	7
Storage Temperature	T _{ST}	-25	80	°C	8

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Referenced to substrate voltage.

2. Includes pins: H1n, H2n, TGx, oRx, OGx, IGx.

3. Voltage difference (either polarity) between any two pins.

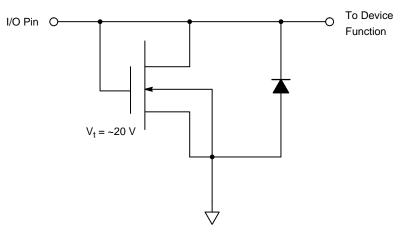
4. Includes pins: VIDn, VSSn, RDx, VDDx, LS and IDx.

- 5. Care must be taken not to short output pins to ground during operation as this may cause permanent damage to the output structures.
- 6. Charge transfer efficiency will degrade at frequencies higher than the maximum clocking frequency. VIDn load resistor values may need to be decreased as well.

7. Noise performance will degrade with increasing temperatures.

- 8. Long term storage at the maximum temperature will accelerate color filter degradation.
- 9. Exceeding the upper limit on output load capacitance will greatly reduce the output frequency response. Thus, direct probing of the output pins with conventional oscilloscope probes is not recommended.
- 10. The absolute maximum ratings for the entire table indicate the limits of this device beyond which damage may occur. The Operating ratings indicate the conditions that the device is functional. Operating at or near these ratings do not guarantee specific performance limits. Guaranteed specifications and test conditions are contained in the Imaging Performance section.

Device Input ESD Protection Circuit (Schematic)



CAUTION: To allow for maximum performance, this device was designed with limited input protection; thus, it is sensitive to electrostatic induced damage. These devices should be installed in accordance with strict ESD handling procedures!

Figure 22. ESD Protection Circuit

DC Bias Operating Conditions

Table 6. DC BIAS OPERATING CONDITIONS

Description	Symbol	Minimum	Nominal	Maximum	Units	Notes
Substrate	V _{SUB C,L}	-	0	-	V	
Reset Drain Bias (Color)	V _{RD C,L}	11.5	12.0	12.5	V	
Output Buffer Supply	V _{DD C,L}	11.5	12.0	12.5	V	
Output Bias Current/Channel	I _{VIDPIN}	-4.0	-6.0	-8.0	mA	1
Light Shield/Drain Bias	V _{LS}	11.5	12.0	12.5	V	
Test Pin – Input Gate	V _{IG C,L}	-	0	-	V	
Test Pin – Input Diode	V _{ID C,L}	-	12.0	-	V	

 A current sink must be supplied for each output. Load capacitance should be minimized so as not to limit bandwidth. The values of R_X and R_L should be chosen to optimize for a given operating frequency, but. Rx should not be less than 75 Ω. The values shown in Figure 23 below represent one possible solution.

Typical Output Bias/Buffer Circuit

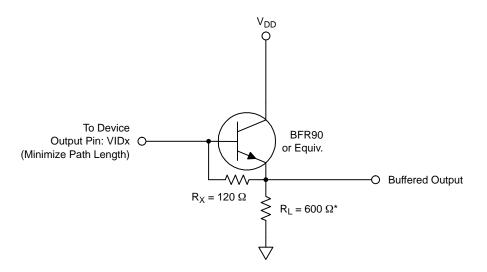


Figure 23. Typical Output Bias/Buffer Circuit

AC Operating Conditions

Table 7. AC ELECTRICAL CHARACTERISTICS

Description	Symbol	Minimum	Nominal	Maximum	Units	Notes	
CCD Element Duration	$1e = 1/f_{CLK}$	50	50	-	ns	1 e Count	
ϕ 1L, ϕ 1C, ϕ 2L, ϕ C2, Rise Time	t _R	-	30	-	ns	Typical	
Line/Integration Period	$1L = t_{INT}$	0.1064	2.128	-	ms	2128 e Counts	
PD-CCD Transfer Period	t _{PD}	1,000	-	-	ns	8 e Counts	
Transfer Gate 1 Clear	t _{TG1}	500	1,000	-	ns	1 e Count	
Transfer Gate 2 Clear	t _{TG2}	500	1,000	-	ns	1 e Count	
Reset Pulse Duration	t _{RST}	9	-	-	ns	1	
Clamp to ϕ 2 Delay	t _{CD}	5	-	-	ns	2	
Sample to Reset Edge Delay	t _{SD}	5	-	-	ns	2	
LOG Gate Duration	t _{LOG1}	1,000	-	-	ns		
LOG Gate Clear	t _{LOG2}	1,000	-	-	ns		

Minimum values given are for 20 MHz CCD operation.
 Recommended delays for Correlated Double Sampling (CDS) of output.

Table 8. CLOCK LEVELS

Description	Symbol	Minimum	Nominal	Maximum	Units
CCD Readout Clocks High	$V_{\phi1CH}, V_{\phi2CH}, V_{\phi1LH}, V_{\phi2LH}$	4.6	5.0	-	V
CCD Readout Clocks Low	$V_{\varphi 1 CL}, V_{\varphi 2 CL}, V_{\varphi 1 LL}, V_{\varphi 2 LL}$	-0.1	0.0	0.1	V
Transfer Clocks High	V _{TGLH} , V _{TG1H} , V _{TG2H}	4.6	5.0	-	V
Transfer Clocks Low	V _{TGLL} , V _{TG1L} , V _{TG2L}	-0.1	0.0	0.1	V
Reset Clock High	V _{¢RCH} , V _{¢RLH}	4.6	5.0	-	V
Reset Clock Low	$V_{\phi RCL}, V_{\phi RLL}$	-0.1	0.0	0.1	V

Care should be taken to insure that low rail overshoot does not exceed -0.5 VDC. Exceeding this value may result in non-photogenerated charged being injected into the video signal.
 Connect pin to ground potential for applications where exposure control is not required.

TIMING

Requirements and Characteristics

Table 9. CLOCK LINE CAPACITANCE

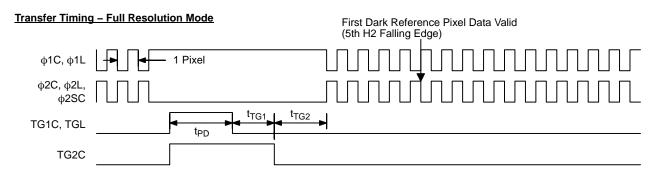
Description	Symbol	Minimum	Nominal	Maximum	Units	Notes
CHROMA	·					
Phase 1 Clock Capacitance	C _{φ1C}	-	758	-	pF	1
Phase 2 Clock Capacitance	C _{φ2C}	-	558	-	pF	1
Transfer Gate 1 Capacitance	C _{TG1C}	-	440	-	pF	
Transfer Gate 2 Capacitance	C _{TG2C}	-	222	-	pF	
Reset Gate Capacitance	C _{¢RC}	-	6	-	pF	
LUMA						
Phase 1 Clock Capacitance	C _{φ1L}	-	397	-	pF	1
Phase 2 Clock Capacitance	C _{φ2L}	-	302	-	pF	1
Transfer Gate Capacitance CTC		-	92	-	pF	
Reset Gate Capacitance	C _{¢RL}	-	6	-	pF	

1. This is the total load capacitance per CCD phase. Since the CCDs are driven from both ends of the sensor, the effective load capacitance per drive pin is approximately half the value listed.

Line Timing – Full Resolution Mode

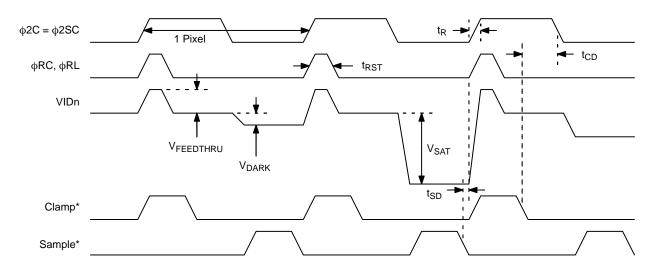
φ1C, φ1L							
φ2C, φ2L, φ2SC		4 Blank Pixels*	12 Dark Pixels*	2098 Active Pixels*	12 Dark Pixels*	2 Blank Pixels*	4 Blank Pixels*
TG1C, TGL	□			t _{INT}			
TG2C	\square						

* Pixel counts are per output.

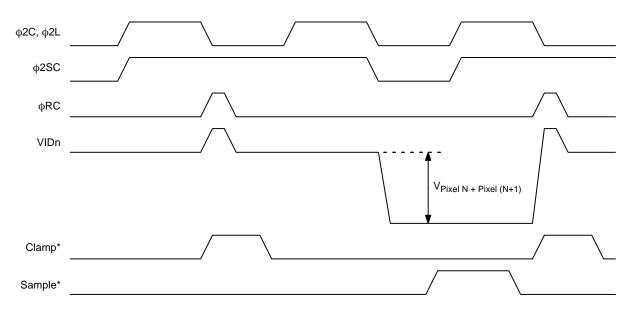




Output Timing – Full Resolution Mode



Output Timing – 2-Pixel Summing Mode



* Required for Optional Off-Chip, Analog, Correlated Double Sampling (CDS) Signal Processing.

Figure 25. Output Timing

MECHANICAL DRAWINGS

Completed Assembly

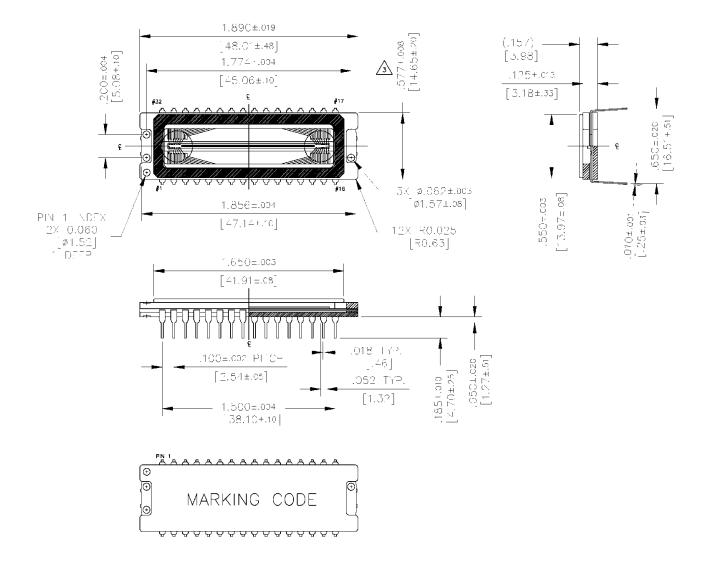
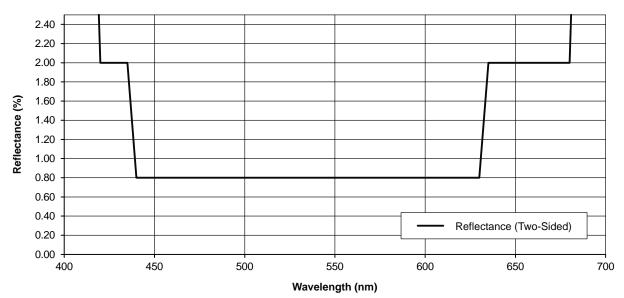


Figure 26. Completed Assembly

COVER GLASS SPECIFICATION

Two-Sided Multi-Layer Anti-Reflective Cover Glass Specification (MAR)

This device is configured with a coverglass designed to reduce reflections and maximize transmission of the visible light. The typical spectral characteristics of this glass is found below:



Maximum Reflectance Allowed (Two-Sided)



REFERENCES

For information on ESD and cover glass care and cleanliness, please download the *Image Sensor Handling and Best Practices* Application Note (AN52561/D) from www.onsemi.com.

For information on soldering recommendations, please download the Soldering and Mounting Techniques Reference Manual (SOLDERRM/D) from www.onsemi.com.

For quality and reliability information, please download the *Quality & Reliability* Handbook (HBD851/D) from <u>www.onsemi.com</u>.

For information on device numbering and ordering codes, please download the *Device Nomenclature* technical note (TND310/D) from <u>www.onsemi.com</u>.

For information on Standard terms and Conditions of Sale, please download <u>Terms and Conditions</u> from <u>www.onsemi.com</u>.

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