

Circuit Note

## **CN-0417**

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Devices Connected/Referenced			
ADL5606	1800 MHz to 2700 MHz, 1 W RF Driver Amplifier		
LTM8045	Inverting or SEPIC µModule, DC-to-DC Converter with Up to 700 mA Output Current		
ADT6402	Low Cost, 2.7 V to 5.5 V, Pin-Selectable Temperature Switch		

### USB powered 2.4 GHz RF Power Amplifier with Over Temperature Management

#### **EVALUATION AND DESIGN SUPPORT**

- Circuit Evaluation Boards
  - ► CN-0417 Circuit Evaluation Board (EVAL-CN0417-EBZ)
- Design and Integration Files
  - Schematics, Layout Files, Bill of Materials

#### **CIRCUIT FUNCTION AND BENEFITS**

Most of the modern radio-link systems capable of transmitting signals have limited output power. Depending on the environment and signal power, the range of transmission varies. For longer range of operation or environments with more RF interference, a higher output power is required. In this case, RF power amplifiers are used to increase the magnitude of power of transmitted signals to a level high enough to reach a given distance.

The circuit shown in Figure 1 is a small USB powered RF power amplifier optimized for 2400 MHz operation. The amplifier typically provides 20 dB of gain through its RF band of operation, boosting signals for various communication protocols such as ISM, MC-GSM, W-CDMA, TD-SCDMA and LTE.

It requires 5 V USB supply for normal operation. An automatic over temperature management circuit ensures that the board temperature remains within recommended operating range. The input and output, both populated with edge mounted SMA connectors, are dc blocked and matched to 50  $\Omega$  for ease of use.

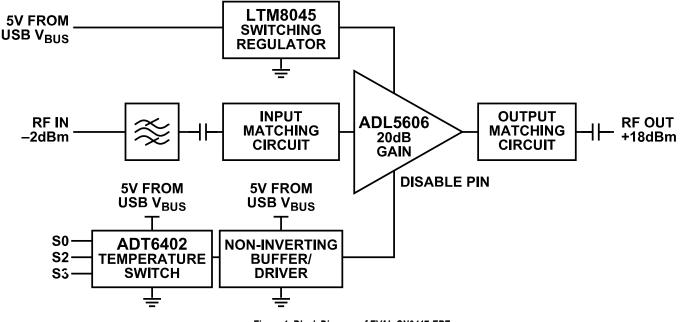


Figure 1. Block Diagram of EVAL-CN0417-EBZ

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#### **CIRCUIT DESCRIPTION**

#### AMPLIFIER

The ADL5606 is a broadband, two-stage, 1 W RF driver amplifier that operates over a frequency range of 1800 MHz to 2700 MHz. The device can be used in a wide variety of wired and wireless communication protocols, including ISM, MC-GSM, W-CDMA, TD-SCDMA, and LTE.

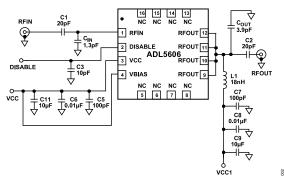


Figure 2. Basic Connections on ADL5606

The ADL5606 operates on a 5 V supply voltage and a supply current of 362 mA. The driver also incorporates a fast power up/ power-down function for time division duplex (TDD) applications such as WiMAX and Wi-Fi, applications that require a power saving mode and applications that intermittently transmit data. When disabled, the ADL5606 draws approximately 4 mA of current from the power supply and 1.4 mA from the DISABLE pin.

#### **IMPEDANCE MATCHING**

The RF input (Pin 1) and RF outputs (Pin 9 to Pin 12) of the ADL5606 can be matched to  $50 \Omega$  with at most one external component and the microstrip line used as an inductor.

The recommended component values for ADL5606 matching are provided in the product data sheet for three frequency bands: 1960 MHz, 2140 MHz, and 2630 MHz. For this application, the recommended matching capacitance values used were from the 2630 MHz operation.

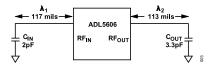


Figure 3. ADL5606 Matching Network Parameters

The placement of these matching capacitors are critical and is shown in Figure 3. The recommended spacing for ADL5606 matching is provided in the product data sheet for the same three frequency bands: 1960 MHz, 2140 MHz, and 2630 MHz. Again, for this application, the recommended spacing used was for the 2630 MHz operation. This was further optimized using simulations in Advanced Design System (ADS) for the desired 2400 MHz operation. Results from simulation gave 117 mils for the input matching length and 113 mils for the output matching length. The component

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spacing is referenced from the center of the matching component to the edge of the ADL5606.

#### **BANDPASS FILTER**

The input signal is filtered by a bandpass filter, centered at 2450 MHz with a 100 MHz bandwidth of operation. Figure 4 shows the electrical performance of the bandpass filter.

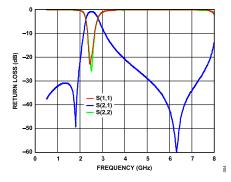


Figure 4. Typical Electrical Performance of Bandpass Filter

It has a frequency of operation from 2400 MHz to 2500 MHz, return loss of 1.5 dB maximum and return loss of 9.5 dB maximum. The maximum input power to the bandpass filter is 2 W.

#### **USB POWER MANAGEMENT**

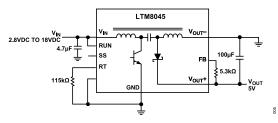


Figure 5. Basic SEPIC Connections for LTM8045 with 5 volts Output

The circuit is powered by the 5 V  $V_{BUS}$  voltage available on the USB port, which is then regulated by the LTM8045.

The LTM8045 is an integrated switching dc-to-dc converter that contains a current mode controller, power switching element, power coupled inductor, power Schottky diode, and a modest amount of input and output capacitance.

The device is configured as a SEPIC converter capable of accepting input voltage up to 18 V dc. The output is adjustable between 2.5 V and 15 V. It can provide approximately 430 mA at V<sub>IN</sub> = 5 V when V<sub>OUT</sub> = 5 V or -5 V.

The LTM8045 output voltage is set by connecting the feedback resistor ( $R_{FB}$ ) from  $V_{OUT}$ + to the FB pin. This voltage serves as the supply voltage for the amplifier ADL5606. Its value is determined from the following equation:

$$R_{FB} = \frac{V_{OUT} - 1.215}{0.0833} \quad (k\Omega)$$
 (1)

## **CIRCUIT DESCRIPTION**

The LTM8045 has an operational switching frequency range between 200 kHz and 2 MHz. The switching frequency of the LTM8045 is configured using an external resistor from the RT pin to ground. Value of the eternal resistor can be determined using the following equation:

$$R_T = \frac{91.9}{f_{OSC}} - 1 \quad (k\Omega) \tag{2}$$

where  $f_{OSC}$  is the typical switching frequency in MHz.

Though the LTM8045 is flexible enough to accommodate a wide range of operating frequencies, a haphazardly chosen one may result in undesirable operation under certain operating or fault conditions. A frequency that is too high can reduce efficiency, generate excessive heat, or even damage the LTM8045 in some fault conditions. A frequency that is too low can result in a final design that has too much output ripple or too large of an output capacitor.

The recommended switching frequency and resistor value for optimal efficiency over the given input and output conditions are provided in the LTM8045 data sheet. In the circuit, the  $R_T$  resistor value is achieved by using two resistors in parallel to provide two places for feedback that sum together with a double filter, achieving a better filter performance.

Because the LTM8045 is a coupled inductor SEPIC, it is susceptible to large switching spikes. That is why in addition to ferrite bead at the output, an LC filter stage was for the switching spikes at 80 MHz to 150 MHz.

## **RF PERFORMANCE**

When operating the CN0417 in the 2.4GHz frequency range, the input signal is amplified from the RF input to output at a typical gain of 20 dB. Figure 6 shows the typical S-parameters of the circuit.

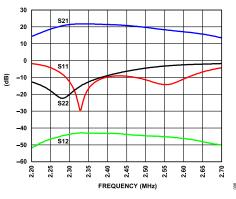


Figure 6. S-Parameters of EVAL-CN0417-EBZ

## OVER TEMPERATURE MANAGEMENT

Due to the small board size and high-power dissipation of the gain amplifier, the board could get hot beyond 85 °C. Operation of the ADL5606 beyond this temperature affects its performance. To combat this issue, the circuit provides an over temperature

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management circuit which will automatically disable the ADL5606 amplifier upon over temperature detection, and enable it again when below a temperature threshold. It consists of a temperature switch, a non-inverting buffer, and a red LED as an indicator as shown in Figure 7. The output of this is connected to the disable pin of the ADL5606.

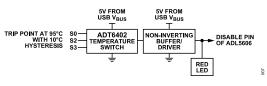


Figure 7. Over Temperature Management Block Diagram

The ADT6402 is a push-pull output temperature switch powered by the 5V from the USB bus. Its output is active high if the temperature measured is greater than the pin-selectable trip level.

The temperature trip level and hysteresis value can be set through S0, S1 and S2 pins of the ADT6402. These three pins can be connected to VCC, tied to GND, or left floating. The ADT6402 decodes the inputs on S0, S1, and S2 to determine the temperature trip point and hysteresis value, as outlined in Figure 8. Overall accuracy for the ADT6402 is  $\pm$ 6°C (maximum) from -45°C to +115°C.

			Temperature	
S2	S1	S0	Trip Point	Hysteresis
0	0	0	+45℃	2°C
0	0	1	+55°C	2°C
0	0	Float	+65°C	2°C
0	1	0	+75°C	2°C
0	1	1	+85°C	2°C
0	1	Float	+95°C	2°C
0	Float	0	+105°C	2°C
0	Float	1	+115°C	2°C
0	Float	Float	+55°C	10°C
1	0	0	+65°C	10°C
1	0	1	+75°C	10°C
1	0	Float	+85°C	10°C
1	1	0	+95℃	10°C
1	1	1	+105°C	10°C
1	1	Float	+115°C	10°C
1	Float	0	+5°C	2°C
1	Float	1	-5°C	2°C
1	Float	Float	-15℃	2°C
Float	0	0	−25°C	2°C
Float	0	1	-35°C	2°C
Float	0	Float	-45°C	2°C
Float	1	0	+5°C	10°C
Float	1	1	-5°C	10°C
Float	1	Float	–15℃	10°C
Float	Float	0	–25℃	10°C
Float	Float	1	−35°C	10°C
Float	Float	Float	–45°C	10°C

 $^1$  0 = pin tied to GND, 1 = pin tied to  $V_{\mbox{\tiny CC}},$  Float = pin left floating.

#### Figure 8. ADT6402 Trip points and hysteresis selection table

In this circuit, a 95 °C trip point with 10 °C hysteresis is selected. Hysteresis prevents oscillation on the output pin when the temperature is approaching the trip point and after the output pin is activated. Given this, the temperature reading of the ADT6402 must

#### **CIRCUIT DESCRIPTION**

go as low as 85 °C before the over temperature circuit output goes low, which turns on the amplifier as shown in Figure 9.

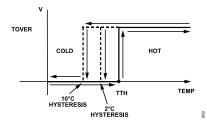


Figure 9. ADT6402 Temperature switch output with hysteresis

In the disabled state, the ADL5606 draws approximately 1.4 mA of current from the DISABLE pin. Since the ADT6402 is only able to source  $800\mu$ A, a non-inverting buffer is used to satisfy the 1.4mA current requirement for the DISABLE pin. The buffer is also powered by the 5V from the USB bus.

#### LAYOUT CONSIDERATIONS

Because the size of the board is quite small, the board temperature reaches approximately 85°C around the vicinity of the ADL5606 in normal operation. To spread this heat evenly and keep the board temperature below 85°C, this design has 6 layer stack-up, the

ground copper plane was made thicker at 3 oz, and more vias were added around the ADL5606 chip. Sigrity simulation, shown in Figure 10, can also be used to estimate the board temperature.

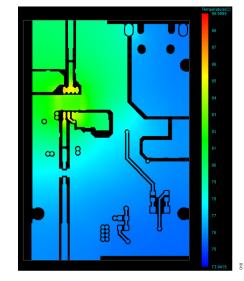


Figure 10. Board Temperature Simulation at Normal Operation

#### **COMMON VARIATIONS**

The CN0522 is designed and optimized for 915MHz band, providing a typical gain of 23dB.

For a different frequency band of operation, the ADL5604 and ADL5605 can be used. The ADL5604 is capable of a wide frequency of operation from 700 MHz to 2700 MHz, with a typical gain of 12.2 dB. The ADL5605 operates from 700 MHz to 1000 GHz at 23 dB typical gain.

Analog Devices, Inc., driver amplifiers are available in a wide range of medium power general-purpose amplifiers covering the frequency range from 400 MHz (IF) to RF microwave and W-band (86 GHz). These driver amplifiers include output power levels from 15 dBm up to approximately 1 W and covers various frequencies, bandwidths, and gain levels.

#### **CIRCUIT EVALUATION AND TEST**

#### **EQUIPMENT NEEDED**

The following equipment is needed:

- ▶ EVAL-CN0417-EBZ reference design board
- ► RF signal source
- Micro-USB power adaptor
- ▶ SMA to SMA cable

#### FUNCTIONAL BLOCK DIAGRAM

Figure 11 shows the functional block diagram of the test setup.

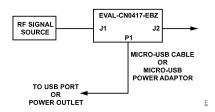


Figure 11. Test Setup Functional Block Diagram

#### SETUP AND TEST

- 1. Directly connect the RF output port of the RF signal source to the EVAL-CN0417-EBZ RF input (J1).
- Connect the RF output of the EVAL-CN0417-EBZ (J2) to the RF network analyzer equipment through an SMA male to SMA male cable.
- 3. Plug in the micro-USB cable to P1 of the EVAL-CN0417-EBZ.
- 4. Plug the other end to a power outlet or a PC USB port.
- 5. The EVAL-CN0417-EBZ then automatically turns on.

**6.** Using the network analyzer, sweep the frequency from 2.2GHz to 2.6GHz and measure the S-parameters.



Figure 12. Top View of EVAL-CN0417-EBZ Board



Figure 13. Bottom View of EVAL-CN0417-EBZ Board

Figure 12 shows a photograph of the top side of the EVAL-CN0417-EBZ containing the ADL5606 and other RF components. Figure 13 shows the bottom side of the EVAL-CN0417-EBZ with micro-USB connector and power supply circuitry.

For complete information and details regarding test setup, see the CN0417 User Guide.

# **Circuit Note**

## LEARN MORE

Musceac, Adrian. "RF power amplifiers for SDR." qradiolink.org. 2017.

ADIsimRF<sup>™</sup> RF Signal Chain Calculator: https://www.analog.com/en/design-center/adisimrf.html

LTspice<sup>®</sup> SPICE Simulation Software: https://www.analog.com/en/ design-center/design-tools

#### -and-calculators/ltspice-simulator.html

LTpowerCAD<sup>®</sup> Design Tool: https://www.analog.com/en/design-center/

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Power Dissipation vs. Die Temperature Calculator: https://www.analog.com/en/design-center/interactive

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## DATA SHEETS AND EVALUATION BOARDS

ADL5606 Data Sheet ADL5606 Evaluation Board LTM8045 Data Sheet DC1533A Evaluation Board

ADT6402 Data Sheet

#### **REVISION HISTORY**

#### 10/2021-Rev. 0 to Rev. A

Updated Figure 11				
Updated Circuit Functions and Benefits				
Paragraph1				
Added Over Temperature Management Section 3				
Added CN0522 in Common Variations5				

11/2018—Revision 0: Initial Version



ESD Caution ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

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