

DESCRIPTION

Demonstration circuit 1317A-G is isolated input to high current output **1/8th Brick footprint** converter featuring the LT[®]1952 switching controller with Active Reset circuit. The Active Reset circuit can improve the efficiency in wide input voltage applications. Also, the Active Reset allows the implementation of self-driven synchronous secondary rectifiers in some applications.

The DC1317A-G (Figure 7) converts isolated 9V to 36V input to 12V output and provides over 8A of output current depending on cooling. When determining the cooling requirements the actual input voltage range and continuous maximum output current must be taken into account. The converter operates at 200kHz with the peak efficiency greater than 94%. The DC1317 can be easily modified to generate output voltages in the range from 0.6V to 48V. The output currents are limited by total output power of up to 150W.

The available versions of DC1317A are:

DC1317A-A, 34-75Vin to 3.3V, 35A

DC1317A-B, 18-72Vin to 5V, 25A

DC1317A-C, 18-72Vin to 12V, 8A-12A

DC1317A-D, 18-72Vin to 24V, 5A

DC1317A-E, 36-72Vin to 5V, 12A

DC1317A-F, 9-36Vin to 3.3V, 22A

* DC1317A-F5, 9-36Vin to 5V, 20A

DC1317A-G, 9-36Vin to 12V, 8A

* DC1317A-G18, 9-36Vin to 18V, 5A

DC1317A-H, 9-36Vin to 48V, 1.5A

* The DC1317A-G18 is a slightly modified design featuring 18V @6A output. Please contact the LTC factory for details.

The DC1317 circuit features soft-start which prevents output voltage overshoot on startup or when recovering from overload condition.

The DC1317 has precise over-current protection circuit that allows for continuous operation under short circuit conditions. The low power dissipation under short circuit conditions insures high reliability even during short circuits.

The LT1952 can be synchronized to an external clock of up to 400kHz. Please refer to LT1952 data sheet for design details and applications information.

Design files for this circuit board are available. Call the LTC factory.

LT is a trademark of Linear Technology Corporation

Table 1. Performance Summary

PARAMETER	CONDITION	VALUE
Minimum Input Voltage	I _{OUT} = 0A to 8A	9V
Maximum Input Voltage	I _{OUT} = 0A to 8A	36V
V _{OUT}	V _{IN} = 9V to 36V, I _{OUT} = 0A to 8A (27A _{max})	12V ±3%
Typical Output Ripple V _{OUT}	V _{IN} = 9V to 36V, I _{OUT} = 0A to 8A	50mV _{p-p}
Nominal Switching Frequency		200kHz

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ACTIVE RESET ISOLATED 9-36V INPUT TO 12V @8A DC/DC POWER CONVERTER

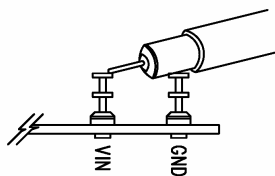


Figure 2. Scope Probe Placement for Measuring Input or Output Ripple

ACTIVE RESET CIRCUIT

The Active Reset circuit on DC1317A-F demo board consists of a small P-Channel MOSFET Q13 and reset capacitor C25. The MOSFET Q13 is used to connect the reset capacitor across the transformer T1 primary winding during the reset period when Q1 MOSFET is off. The voltage across capacitor C25 automatically adjusts with the duty cycle to provide complete transformer reset under all operating conditions.

Also the active reset circuit shapes the reset voltage into a square waveform that results in lower drain voltages for Q1 and Q2 MOSFETs. The lower MOSFET drain voltages allow lower voltage and lower $R_{ds(on)}$ MOSFETs to be used. The MOSFETs must be avalanche rated for the peak reset voltage. If non-avalanche rated MOSFETs are used a proper drain voltage derating should be used.

The main benefit of active reset circuit in the case of DC1317A-G demo board is high efficiency (shown in Figure 3), wide input range, high power density and small size. To achieve such high efficiency all of the power components were carefully selected. Please consult LT factory for assistance if any changes to the circuit are required.

If PCB space is limited, a smaller 1825R2-3366.006 transformer from Champs engineering can be used. However, the efficiency will be about 1% lower. For more information please consult LT factory.

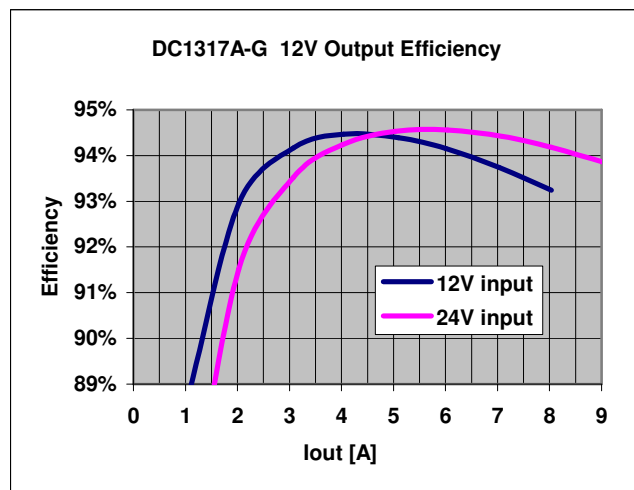


Figure 3. High efficiency of DC1317A-G allows the board to be used in thermally critical applications

18V OUTPUT CIRCUIT

The DC1317A-G demo board can easily be modified for 18V output. Only few minor component changes are required and are shown in the schematic in Figure 8.

The efficiency of 18V Output version of DC1317A-G is shown in the Figure 4. The efficiency is slightly lower than the efficiency of 12V version due to higher DC resistance of transformer and output MOSFETs.

Also, the 18V output can be further increased to **19.5V** just by changing the voltage divider on the FB pin of LT4430 circuit. The 19.5V version is suitable for supplying laptop computer power from 12V automotive power source.

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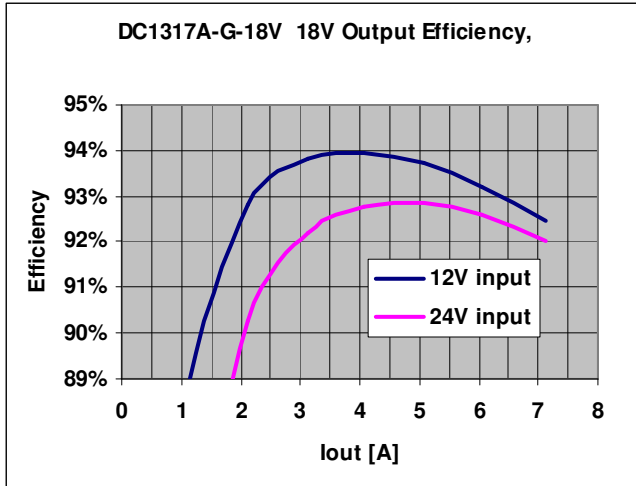


Figure 4. The efficiency of modified DC1317A-G demo board for 18V output.

OUTPUT LOAD STEP RESPONSE

The load step response of DC1317A-G is very fast even though relatively small amount of output capacitance is present (200uF ceramic and 220uF electrolytic). This is thanks to fast error amplifier of LT4430, optimal amount of current slope compensation of LT1952, fast opto coupler and fast error amplifier of LT1952. If higher load steps need to be handled more output capacitance can be added in order to keep the voltage transients at the desired level. The load step transients are shown in Figure 5.

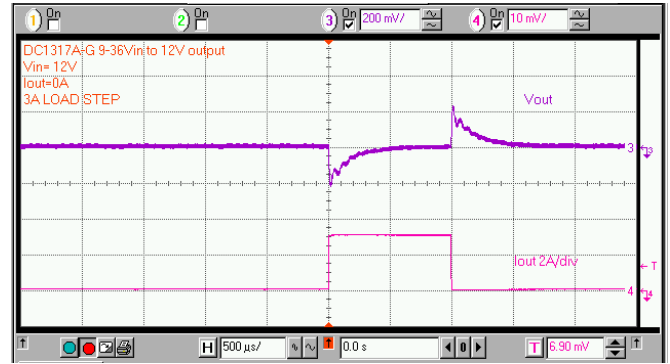


Figure 5. Fast transient response of DC1317A-F is superior to many competing power modules without the additional output capacitors.

SOFT START FUNCTION

The DC1317 features LT4430 opto coupler driver that has soft start function which produces monotonic startup ramp shown in Figure 6. The rise time of output voltage is controlled by capacitor C19 that is connected to OC (Overshoot Control) pin of LT4430. Also, the soft-start function will prevent input current surges even with full load at the output.

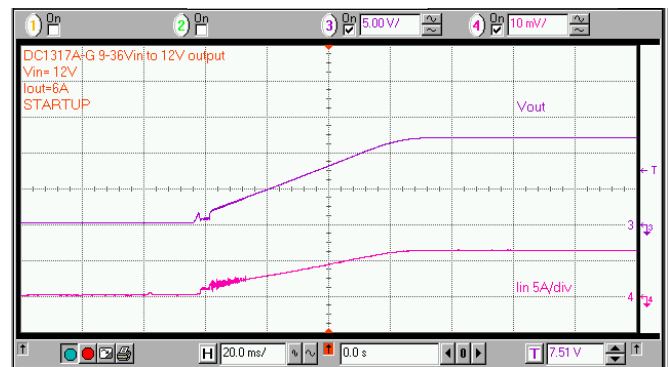


Figure 6. The LT4430 opto coupler driver produces monotonic output voltage rise at startup without output voltage overshoot.

DEBUGGING AND TESTING

The DC1317 can easily be tested and debugged by powering the bias circuit separately from the main power circuit. To place DC1317 into debug mode

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remove the resistor R1 and connect 12V, 100mA power source to +Vb node (right side of R1). By doing this, the primary PWM controller LT1952 can be activated without the main primary power being applied to +Vin.

To activate the secondary side control circuit LT4430 diode OR a 5V, 100mA power source into pin 1 of LT4430 controller.

Once the primary and secondary controllers are running the main power (+Vin) can be applied slowly while observing the switching waveforms and output voltage.

The input current supplying the power transformer T1 should not exceed 200mA without the output load. If one of the MOSFETs is damaged, the input current will exceed 200mA.

PCB LAYOUT

The PCB layout should be carefully planned to avoid potential noise problems. The PCB layout for DC1317A can be used as a guide. Since demo board DC1317A has 8 versions the PCB layout has optional components that can be removed. The schematic in figure 7 is showing the circuit without any optional components. Also, the circuit in figure 7 does not show any zero-ohm resistors. Please consult the schematic on page 7 to determine if any of the options should be included in your PCB layout.

Also, the PCB layout has a common schematic that is used just for the layout. The PCB layout schematic is not included in this quick start guide but it is included with PCB design files.

Please note that the actual circuit schematic on page 7 shows the component values. The PCB layout schematic included with design files does not show the component values.

In some cases, a different component like a diode is used in a place holder for a capacitor such as in the case of C6. Please modify the reference designators in your schematic to reflect the actual component used. The following simple PCB layout rules should be helpful.

If possible use solid ground planes on layers 2 and n-1. The ground planes will prevent the switching noise from coupling into sensitive lines.

Place sensitive lines on the inner layers that will be shielded by grounds on layers 2 and n-1.

Keep the loop formed by Q1, RCS1, Cin and T1 tight.

Keep the loop formed by Q2, Q3 and T1 tight.

Keep noise sensitive nodes like SD/VSEC, ROSC, FB, COMP, ISENSE, BLANK and DELAY as small as possible by placing the associated components close to the LT1952 and LT4430 chips.

Use local vias for all components that connect to ground planes.

Do not place any traces on the layers 2 and n-1 to avoid ground planes from being compromised.

If the PCB layout has to be done on 2 or 4-layer PCB try to stay close to the guidelines outlined above. Also, maximize the ground connections between components by placing the components tight together.

Please contact LT factory for additional assistance.

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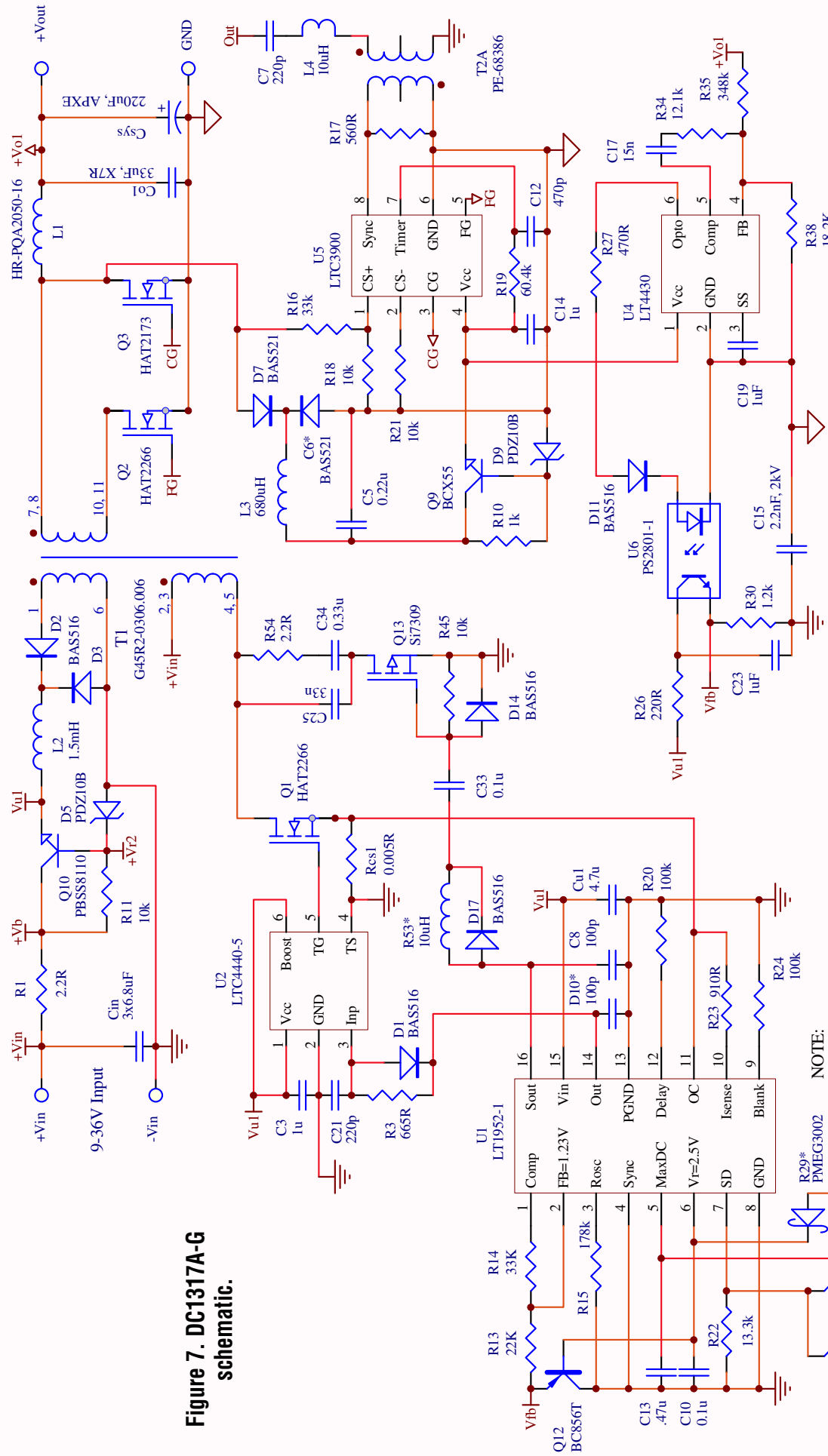


Figure 7. DC1317A-G schematic.

NOTE:

This schematic shows only the components required for operation of -G version of DC1317A demo board.
 All optional components of DC1317A demo board have been removed.
 Also, all zero-ohm resistors have been replaced with wires.
 Please consult the full DC1317A-G schematic to decide if any of the optional components should be included in your design.

Demo Board DC1317A-G schematic

* Please change the ref designator to reflect the symbol.

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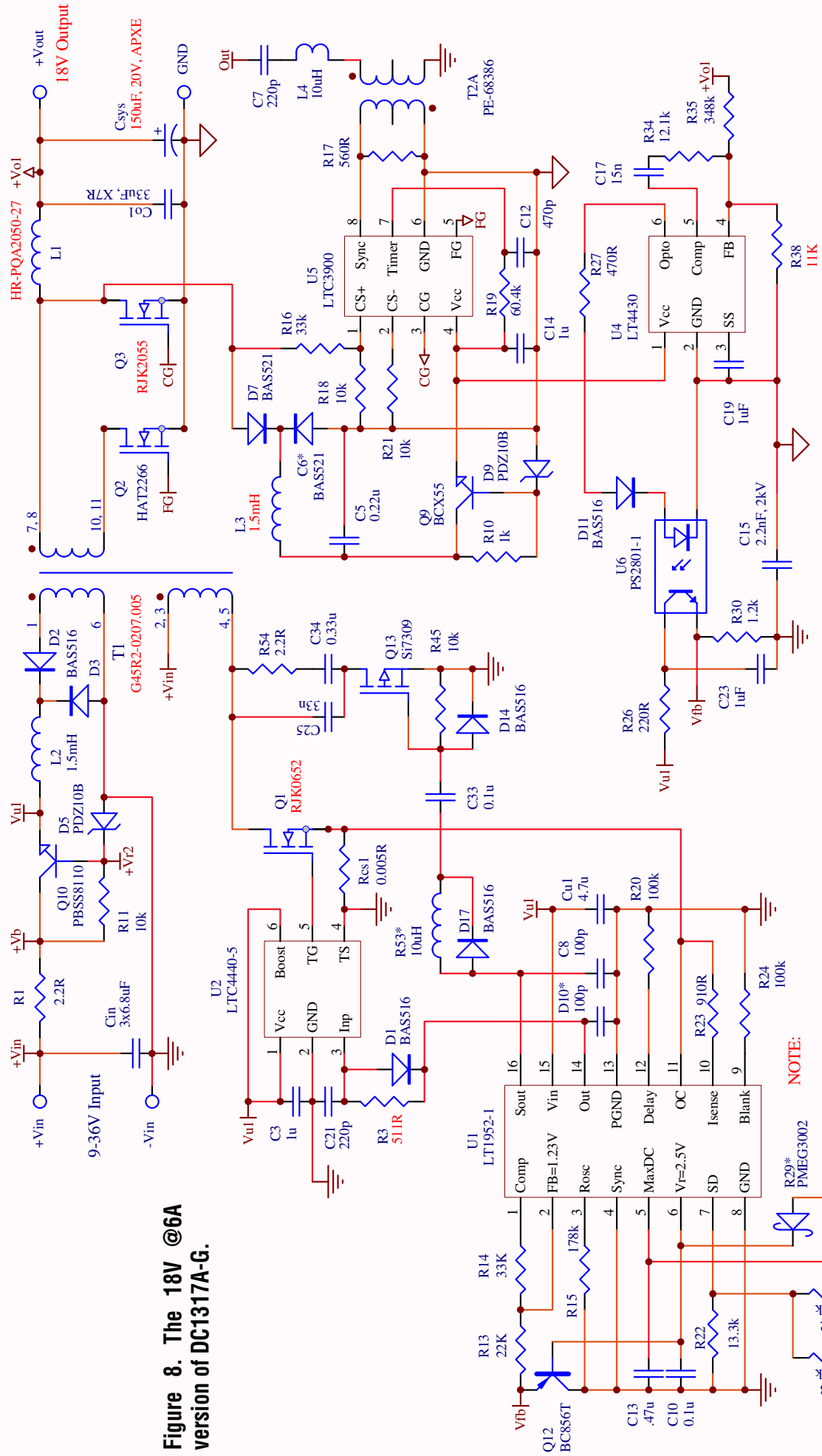
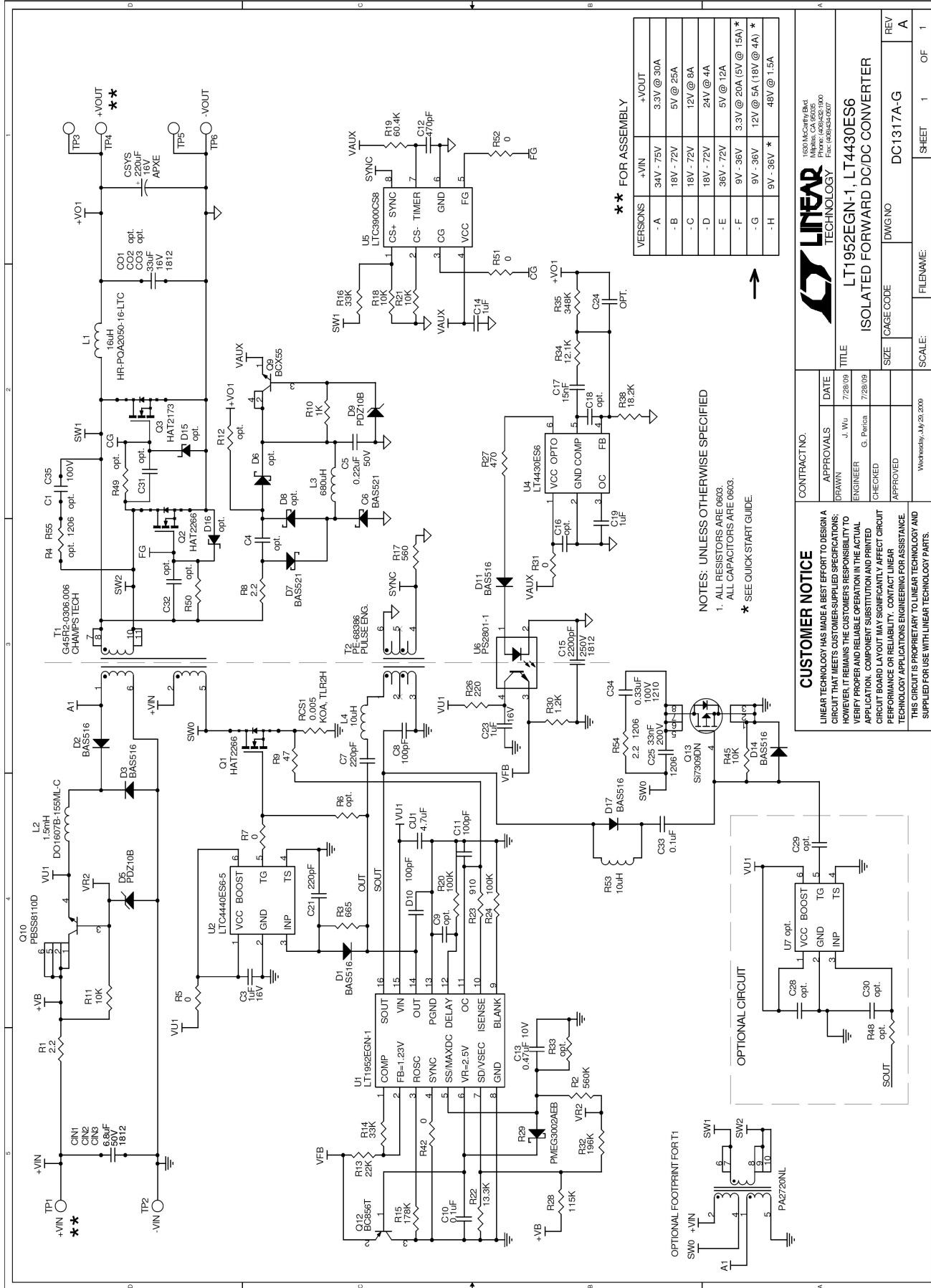


Figure 8. The 18V @6A version of DC1317A-G.

Demo Board DC1317A-G18V schematic

* Please change the ref designator to reflect the symbol.

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★ ★ FOR ASSEMBLY

VERSIONS	+VIN	+VOUT
-A	34V - 75V	3.3V @ 30A
-B	18V - 72V	5V @ 25A
-C	18V - 72V	12V @ 8A
-D	18V - 72V	24V @ 4A
-E	36V - 72V	5V @ 12A
-F	9V - 36V	3.3V @ 20A (5V @ 15A) *
-G	9V - 36V	12V @ 5A (18V @ 4A) *
-H	9V - 36V *	48V @ 1.5A

NOTES: UNLESS OTHERWISE SPECIFIED
 1. ALL RESISTORS ARE 0603.
 ★ SEE QUICK START GUIDE

LINEAR TECHNOLOGY

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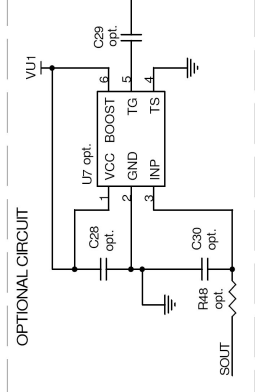
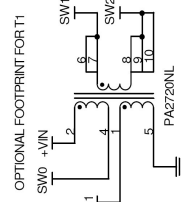
CONTRACT NO.		DATE	
APPROVALS	DRAWN	J. Wu	7/28/09
ENGINEER	G. Petica	CHECKED	
APPROVED			

LT1952EGN-1, LT4430ES6
ISOLATED FORWARD DC/DC CONVERTER

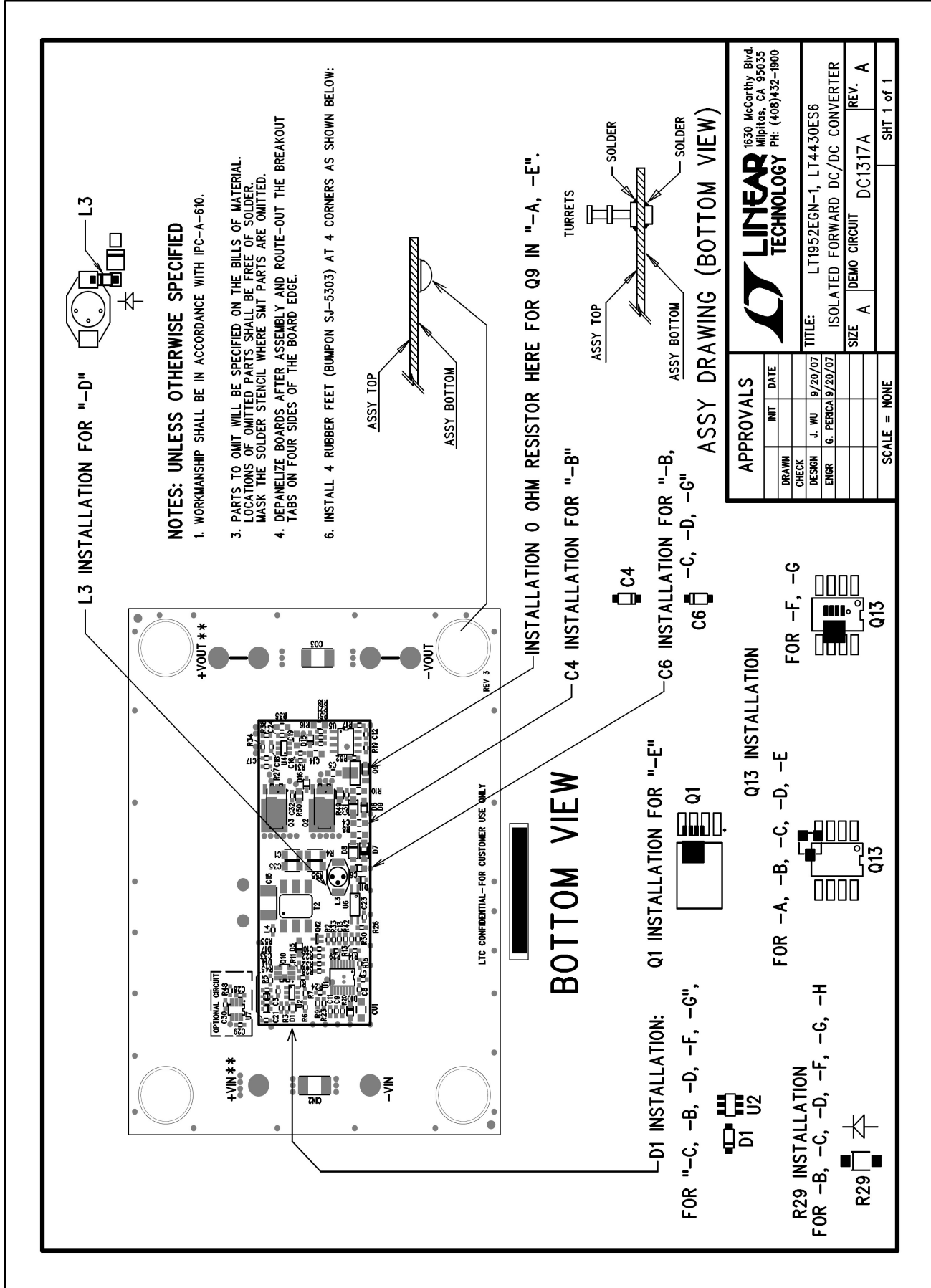
SIZE	CAGE CODE	DWG NO	REV
		DC1317A-G	A

SCALE: _____ FILENAME: _____ SHEET 1 OF 1

CUSTOMER NOTICE
 LINEAR TECHNOLOGY HAS MADE A BEST EFFORT TO DESIGN A CIRCUIT THAT MEETS CUSTOMER-SUPPLIED SPECIFICATIONS; HOWEVER, IT REMAINS THE CUSTOMER'S RESPONSIBILITY TO VERIFY PROPER AND RELIABLE OPERATION IN THE ACTUAL APPLICATION. COMPONENT SUBSTITUTION AND PRINTED CIRCUIT BOARD LAYOUT MAY SIGNIFICANTLY AFFECT CIRCUIT PERFORMANCE OR RELIABILITY. CONTACT LINEAR TECHNOLOGY APPLICATIONS ENGINEERING FOR ASSISTANCE. THIS CIRCUIT IS PROPRIETARY TO LINEAR TECHNOLOGY AND SUPPLIED FOR USE WITH LINEAR TECHNOLOGY PARTS.



QUICK START GUIDE FOR DEMONSTRATION CIRCUIT 1317A-G ACTIVE RESET ISOLATED 9-36V INPUT TO 12V @8A DC/DC POWER CONVERTER



NOTES: UNLESS OTHERWISE SPECIFIED

1. WORKMANSHIP SHALL BE IN ACCORDANCE WITH IPC-A-610.
3. PARTS TO OMIT WILL BE SPECIFIED ON THE BILLS OF MATERIAL. LOCATIONS OF OMITTED PARTS SHALL BE FREE OF SOLDER. MASK THE SOLDER STENCIL WHERE SMT PARTS ARE OMITTED.
4. DEPANELIZE BOARDS AFTER ASSEMBLY AND ROUTE-OUT THE BREAKOUT TABS ON FOUR SIDES OF THE BOARD EDGE.
6. INSTALL 4 RUBBER FEET (BUMPON SJ-5303) AT 4 CORNERS AS SHOWN BELOW:

APPROVALS		LINEAR TECHNOLOGY	
INIT	DATE	1630 McCarthy Blvd. Milpitas, CA 95035 PH: (408)432-1900	
DRAWN		TITLE: LT1952EGN-1, LT4430ESS	
CHECK	J. WU 9/20/07	ISOLATED FORWARD DC/DC CONVERTER	
DESIGN	G. PERCA 9/20/07	SIZE	A
ENGR		DEMO CIRCUIT	DC1317A
		REV.	A
SCALE = NONE		SHT 1 of 1	