

# High-Voltage, Current-Mode PWM Controller

## Features

- ▶ 10 to 450V input voltage range
- ▶ <1.3mA supply current
- ▶ >1.0MHz clock
- ▶ >20:1 dynamic range @ 500KHz
- ▶ 49% Maximum duty cycle version
- ▶ Low internal noise

## Applications

- ▶ Off-line high frequency power supplies
- ▶ Universal input power supplies
- ▶ High density power supplies
- ▶ Very high efficiency power supplies
- ▶ Extra wide load range power supplies

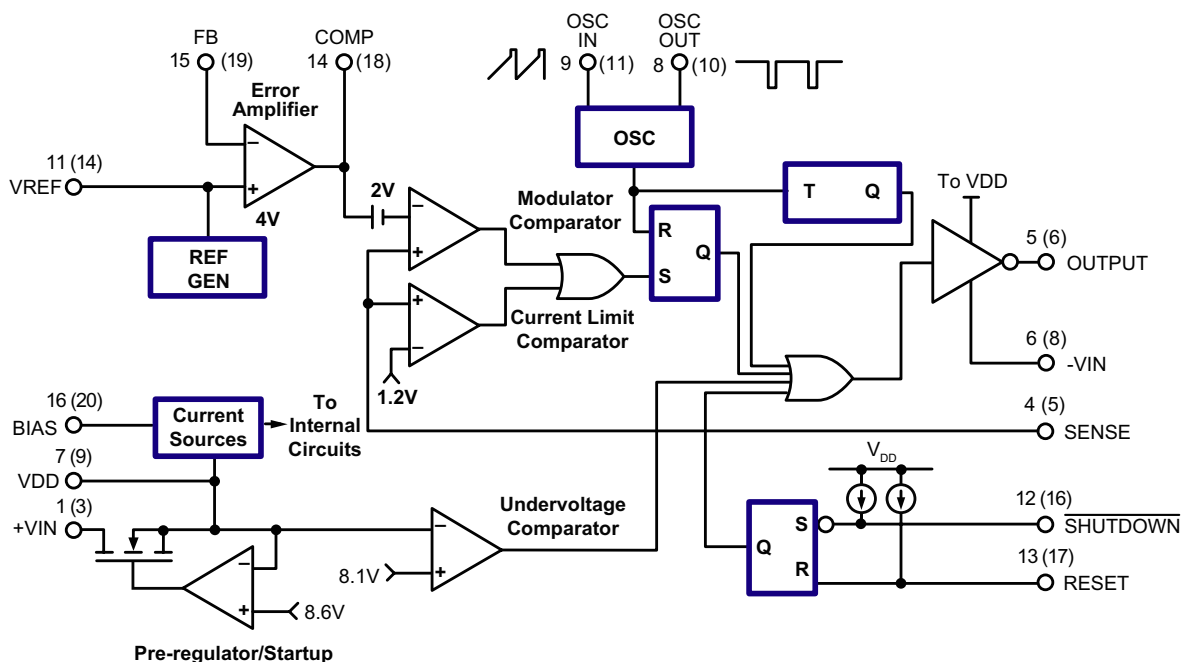
## General Description

The Supertex HV9120 is a Switch Mode Power Supply (SMPS) controller subsystem that can start and run directly from almost any DC input, from a 12V battery to a rectified and filtered 240V AC line. It contains all the elements required to build a single-switch converter except for the switch, magnetic assembly, output rectifier(s) and filter(s).

A unique input circuit allows the HV9120 to self-start directly from a high voltage input, and subsequently take the power to operate from one of the outputs of the converter it is controlling, allowing very efficient operation while maintaining input-to-output galvanic isolation limited in voltage only by the insulation system of the associated magnetic assembly. A  $\pm 2\%$  internal bandgap reference, internal operational amplifier, very high speed comparator, and output buffer allow production of rugged, high performance, high efficiency power supplies of 50W or more, which can still be over 80% efficient at outputs of 1.0W or less. The wide dynamic range of the controller system allows designs with extremely wide line and load variations with much less difficulty and much higher efficiency than usual. The exceptionally wide input voltage range also allows better usage of energy stored in input dropout capacitors than with other PWM ICs. Remote on/off controls allow either latching or nonlatching remote shutdown. During shutdown, the power required is under 6.0mW.

**For detailed circuit and application information, please refer to application notes AN-H13, AN-H21 to AN-H24.**

## Functional Block Diagram



**Note:**  
Pin numbers in parentheses are for PLCC package.

## Ordering Information

Part Number	Package Options	Packing
HV9120NG-G	16-Lead SOIC	45/Tube
HV9120NG-G M934	16-Lead SOIC	2500/Reel
HV9120P-G	16-Lead PDIP	24/Tube
HV9120PJ-G	20-Lead PLCC*	48/Tube
HV9120PJ-G M910	20-Lead PLCC*	1000/Reel

-G Indicates package is RoHS compliant ("Green")

\* Obsolescence notice issued for the product in the 20-Lead PLCC package.



## Typical Thermal Resistance

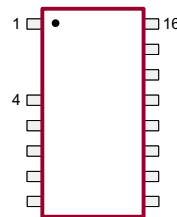
Package	$\theta_{ja}$
16-Lead SOIC	83°C/W
16-Lead PDIP	51°C/W
20-Lead PLCC	66°C/W

## Absolute Maximum Ratings

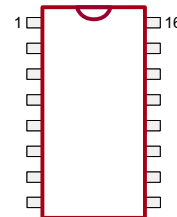
Parameter	Value
Input voltage, $+V_{IN}$	450V
Device supply voltage, $V_{DD}$	15.5V
Logic input voltage	-0.3V to $V_{DD}$ +0.3V
Linear input voltage	-0.3V to $V_{DD}$ +0.3V
Pre regulator input current (continuous), $I_{IN}$	2.5mA
Operating junction temperature, $T_J$	150°C
Storage temperature	-65 to +150°C
Power dissipation:	
16-Lead SOIC	900mW
16-Lead PDIP	1000mW
20-Lead PLCC	1400mW

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. Voltages are referenced to  $-V_{IN}$

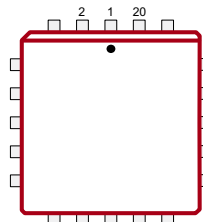
## Pin Configurations



16-Lead SOIC (NG)



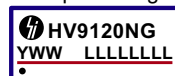
16-Lead PDIP (P)



20-Lead PLCC (PJ)

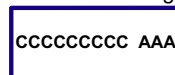
## Product Marking

Top Marking



Y = Last Digit of Year Sealed  
 WW = Week Sealed  
 L = Lot Number  
 C = Country of Origin\*  
 A = Assembler ID\*  
 — = "Green" Packaging  
 \*May be part of top marking

Bottom Marking



Package may or may not include the following marks: Si or

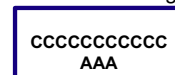
16-Lead SOIC (NG)

Top Marking



Y = Last Digit of Year Sealed  
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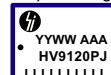
Bottom Marking



Package may or may not include the following marks: Si or

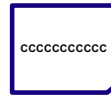
16-Lead PDIP (P)

Top Marking



YY = Year Sealed  
 WW = Week Sealed  
 L = Lot Number  
 A = Assembler ID  
 C = Country of Origin\*  
 — = "Green" Packaging  
 \*May be part of top marking

Bottom Marking



Package may or may not include the following marks: Si or

20-Lead PLCC (PJ)

## Electrical Characteristics

(Unless otherwise specified,  $V_{DD} = 10V$ ,  $+V_{IN} = 48V$ ,  $R_{BIAS} = 390K\Omega$ ,  $R_{OSC} = 330K\Omega$ ,  $T_A = 25^\circ C$ .)

Sym	Parameter	#	Min	Typ	Max	Units	Conditions
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### Reference

$V_{REF}$	Output voltage	-	3.92	4.00	4.08	V	$R_L = 10M\Omega$
			3.84	4.00	4.16		$R_L = 10M\Omega$ , $T_A = -55$ to $125^\circ C$
$Z_{OUT}$	Output impedance	#	15	30	45	K $\Omega$	---
$I_{SHORT}$	Short circuit current	-	-	125	250	$\mu A$	$V_{REF} = -V_{IN}$
$\Delta V_{REF}$	Change in $V_{REF}$ with temperature	#	-	0.25	-	mV/ $^\circ C$	$T_A = -55$ to $125^\circ C$

### Oscillator

$f_{MAX}$	Oscillator frequency	-	1.0	3.0	-	MHz	$R_{OSC} = 0\Omega$
$f_{OSC}$	Initial accuracy <sup>1</sup>	-	80	100	120	KHz	$R_{OSC} = 330K\Omega$
		-	160	200	240		$R_{OSC} = 150K\Omega$
$\Delta V_{OSC}$	Voltage stability	-	-	-	15	%	$9.5V < V_{DD} < 13.5V$
$TC_{OSC}$	Temperature coefficient	#	-	170	-	ppm/ $^\circ C$	$T_A = -55$ to $125^\circ C$

### PWM

$D_{MAX}$	Maximum duty cycle	#	49.0	49.4	49.6	%	---
$D_{MIN}$	Minimum duty cycle	-	-	-	0	%	---
	Maximum pulse width before pulse drops out	#	-	80	125	ns	---

### Current Limit

$V_{LIM}$	Maximum input signal	-	1.0	1.2	1.4	V	$V_{FB} = 0V$
$t_D$	Delay to output	#	-	80	120	ns	$V_{SENSE} = 1.5V$ , $V_{COMP} \leq 2.0V$

### Error Amplifier

$V_{FB}$	Feedback voltage	-	3.92	4.00	4.08	V	$V_{FB}$ shorted to COMP
$I_{IN}$	Input bias current	-	-	25	500	nA	$V_{FB} = 4.0V$
$V_{OS}$	Input offset voltage	-	nulled during trim			-	---
$A_{VOL}$	Open loop voltage gain	#	60	80	-	dB	---
GB	Unity gain bandwidth	#	1.0	1.3	-	MHz	---
$Z_{OUT}$	Out impedance	#	see Fig. 1			$\Omega$	---
$I_{SOURCE}$	Output source current	-	-1.4	-2.0	-	mA	$V_{FB} = 3.4V$
$I_{SINK}$	Output sink current	-	0.12	0.15	-	mA	$V_{FB} = 4.5V$
PSRR	Power supply rejection	#	see Fig. 2			dB	---

#### Notes:

# Guaranteed by design.

1. Stray capacitance on OSC In pin must be  $\leq 5pF$ .

## Electrical Characteristics (cont.)

(Unless otherwise specified,  $V_{DD} = 10V$ ,  $+V_{IN} = 48V$ ,  $R_{BIAS} = 390K\Omega$ ,  $R_{OSC} = 330K\Omega$ ,  $T_A = 25^\circ C$ .)

Sym	Parameter	#	Min	Typ	Max	Units	Conditions
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### Pre-Regulator/Startup

$+V_{IN}$	Input voltage	-	10	-	450	V	$I_{IN} < 10\mu A$ ; $V_{CC} > 9.4V$
$+I_{IN}$	Input leakage current	-	-	-	10	$\mu A$	$V_{DD} > 9.4V$
$V_{TH}$	$V_{DD}$ pre-regulator turn-off threshold voltage	-	8.0	8.7	9.4	V	$I_{PREREG} = 10\mu A$
$V_{LOCK}$	Undervoltage lockout	-	7.0	8.1	8.9	V	---

### Supply

$I_{DD}$	Supply current	-	-	0.75	1.3	mA	$C_L < 75pF$
$I_Q$	Quiescent supply current	-	-	0.55	-	mA	$\overline{SHUTDOWN} = -V_{IN}$
$I_{BIAS}$	Nominal bias current	-	-	20	-	$\mu A$	---
$V_{DD}$	Operating range	-	9.0	-	13.5	V	---

### Shutdown Logic

$t_{SD}$	$\overline{SHUTDOWN}$ delay	#	-	50	100	ns	$C_L = 500pF$ , $V_{SENSE} = -V_{IN}$
$t_{SW}$	$\overline{SHUTDOWN}$ pulse width	#	50	-	-	ns	
$t_{RW}$	RESET pulse width	#	50	-	-	ns	---
$t_{LW}$	Latching pulse width	#	25	-	-	ns	$\overline{SHUTDOWN}$ and RESET low
$V_{IL}$	Input low voltage	-	-	-	2.0	V	---
$V_{IH}$	Input high voltage	-	7.0	-	-	V	---
$I_{IH}$	Input current, input high voltage	-	-	1.0	5.0	$\mu A$	$V_{IN} = V_{DD}$
$I_{IL}$	Input current, input low voltage	-	-	-25	-35	$\mu A$	$V_{IN} = 0V$

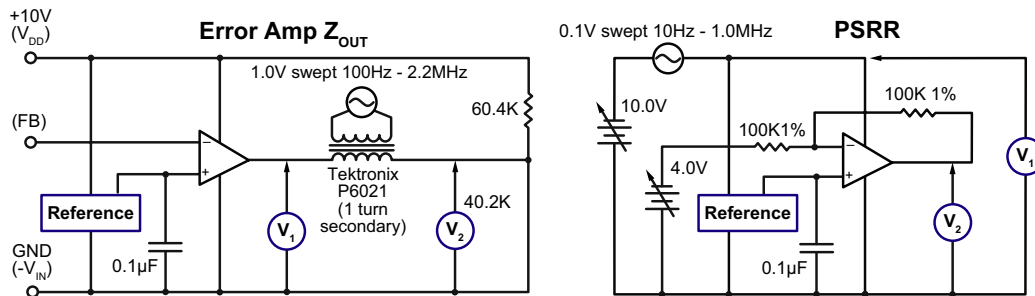
### Output

$V_{OH}$	Output high voltage	-	$V_{DD} - 0.25$	-	-	V	$I_{OUT} = 10mA$	
		-	$V_{DD} - 0.3$	-	-	V	$I_{OUT} = 10mA$ , $T_A = -55$ to $125^\circ C$	
$V_{OL}$	Output low voltage	-	-	-	0.2	V	$I_{OUT} = -10mA$	
		-	-	-	0.3	V	$I_{OUT} = -10mA$ , $T_A = -55$ to $125^\circ C$	
$R_{OUT}$	Output resistance	Pull up	-	-	15	25	$\Omega$	$I_{OUT} = \pm 10mA$
		Pull down	-	-	8.0	20	$\Omega$	
		Pull up	-	-	20	30	$\Omega$	$I_{OUT} = \pm 10mA$ , $T_A = -55$ to $125^\circ C$
		Pull down	-	-	10	30	$\Omega$	
$t_R$	Rise time	#	-	30	75	ns	$C_L = 500pF$	
$t_F$	Fall time	#	-	20	75	ns	$C_L = 500pF$	

#### Note:

# Guaranteed by design.

## Test Circuits



### Note:

Set feedback voltage so that  $V_{COMP} = V_{DIVIDE} \pm 1.0mV$  before connecting transformer.

## Detailed Description

### Pre regulator

The pre regulator/startup circuit for the HV9120 consists of a high-voltage n-channel depletion-mode DMOS transistor driven by an error amplifier to form a variable current path between the VIN terminal and the VDD terminal. Maximum current (about 20 mA) occurs when  $V_{DD} = 0$ , with current reducing as  $V_{DD}$  rises. This path shuts off altogether when  $V_{DD}$  rises to somewhere between 7.8 and 9.4V, so that if  $V_{DD}$  is held at 10 or 12V by an external source (generally the supply the chip is controlling), no current other than leakage is drawn through the high voltage transistor. This minimizes dissipation.

An external capacitor between VDD and VSS is generally required to store energy used by the chip in the time between shutoff of the high voltage path and the VDD supply's output rising enough to take over powering the chip. This capacitor should have a value of 100X or more the effective gate capacitance of the MOSFET being driven, i.e.,

$$C_{STORAGE} \geq 100 \times (\text{gate charge of FET at } 10V)$$

as well as very good high frequency characteristics. Stacked polyester or ceramic caps work well. Electrolytic capacitors are generally not suitable. A common resistor divider string is used to monitor  $V_{DD}$  for both the undervoltage lockout circuit and the shutoff circuit of the high voltage FET. Setting the undervoltage sense point about 0.6V lower on the string than the FET shutoff point guarantees that the undervoltage lockout always releases before the FET shuts off.

### Bias Circuit

An external bias resistor, connected between the bias pin and VSS is required by the HV9120 to set currents in a series of current mirrors used by the analog sections of the chip. Nominal external bias current requirement is 15 to 20µA, which can be set by a 390 to 510KΩ resistor if a 10V

$V_{DD}$  is used, or a 510 to 680KΩ resistor if  $V_{DD}$  will be 12V. A precision resistor is not required;  $\pm 5\%$  is fine.

### Clock Oscillator

The clock oscillator of the HV9120 consists of a ring of CMOS inverters, timing capacitors, a capacitor discharge FET, and a frequency dividing flip-flop. A single external resistor between the OSC IN and OSC OUT pins is required to set oscillator frequency (see graph).

One difference exists between the Supertex HV9120 and competitive 9120s: The oscillator is shut off when a shutoff command is received. This saves about 150µA of quiescent current, which aids in the construction of power supplies to meet CCITT specification I-430, and in other situations where an absolute minimum of quiescent power dissipation is required.

### Reference

The Reference of the HV9120 consists of a stable bandgap reference followed by a buffer amplifier which scales the voltage up to approximately 4.0V. The scaling resistors of the reference buffer amplifier are trimmed during manufacture so that the output of the error amplifier, when connected in a gain of -1 configuration, is as close to 4.0V as possible. This nulls out any input offset of the error amplifier. As a consequence, even though the observed reference voltage of a specific part may not be exactly 4.0V, the feedback voltage required for proper regulation will be.

A  $\approx 50K\Omega$  resistor is placed internally between the output of the reference buffer amplifier and the circuitry it feeds (reference output pin and non-inverting input to the error amplifier). This allows overriding the internal reference with a low-impedance voltage source  $\leq 6.0V$ . Using an external reference reinstates the input offset voltage of the error amplifier, and its effect of the exact value of feedback voltage

required. In general, because the reference voltage of the Supertex HV9120 is not noisy, as some previous examples have been, overriding the reference should seldom be necessary.

Because the reference of the HV9120 is a high impedance node, and usually there will be significant electrical noise near it, a bypass capacitor between the reference pin and VSS is strongly recommended. The reference buffer amplifier is intentionally compensated to be stable with a capacitive load of 0.01 to 0.1  $\mu\text{F}$ .

## Error Amplifier

The error amplifier in the HV9120 is a true low-power differential input operational amplifier intended for around-the-amplifier compensation. It is of mixed CMOS-bipolar construction: A PMOS input stage is used so the common-mode range includes ground and the input impedance is very high. This is followed by bipolar gain stages which provide high gain without the electrical noise of all-MOS amplifiers. The amplifier is unity-gain stable.

## Current Sense Comparators

The HV9120 uses a true dual-comparator system with independent comparators for modulation and current limiting. This allows the designer greater latitude in compensation design, as there are no clamps (except ESD protection) on the compensation pin. Like the error amplifier, the comparators are of low-noise BiCMOS construction.

## Remote Shutdown

The  $\overline{\text{SHUTDOWN}}$  and RESET pins of the HV9120 can be used to perform either latching or non-latching shutdown of a converter as required. These pins have internal current source pull-ups so they can be driven from open-drain logic. When not used, they should be left open or connected to VDD.

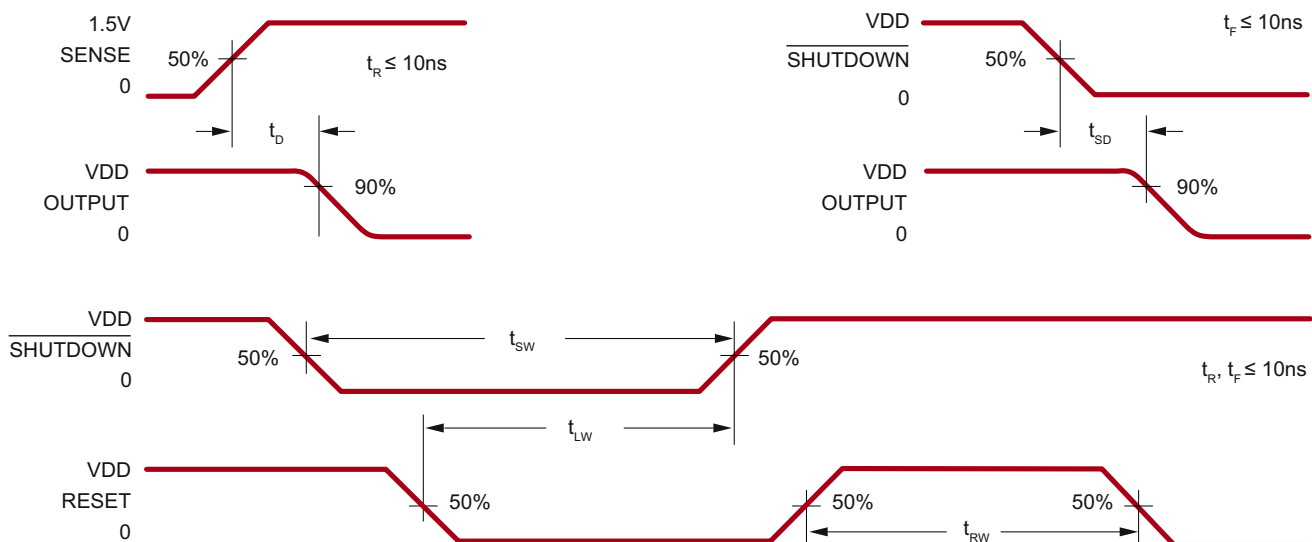
## Output Buffer

The output buffer of the HV9120 is of standard CMOS construction (P-channel pull-up, N-channel pull-down). Thus the body-drain diodes of the output stage can be used for spike clipping if necessary, and external Schottky diode clamping of the output is not required.

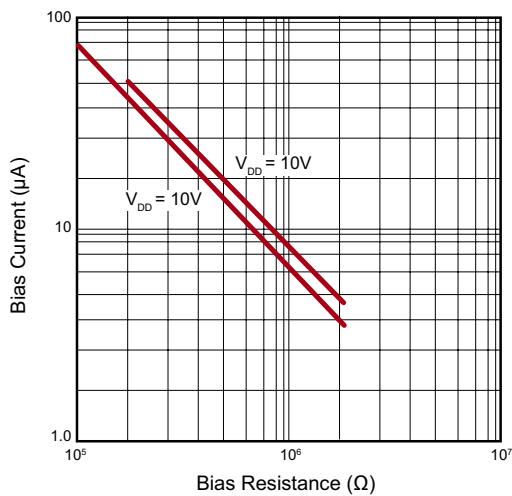
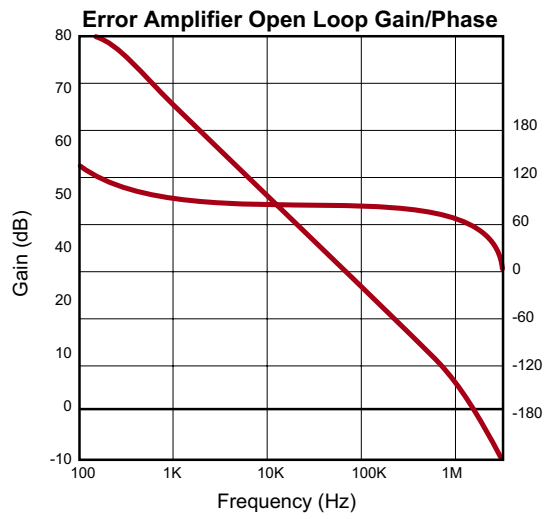
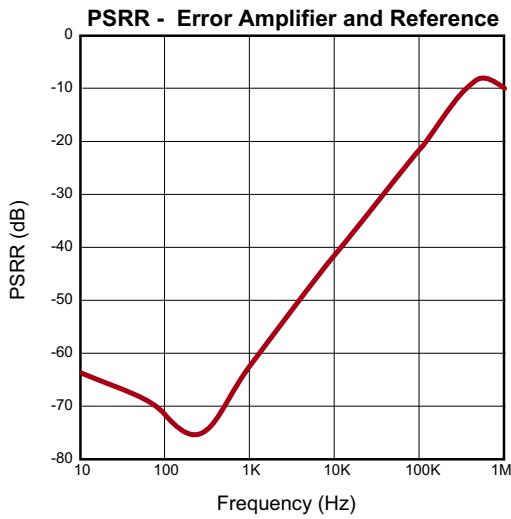
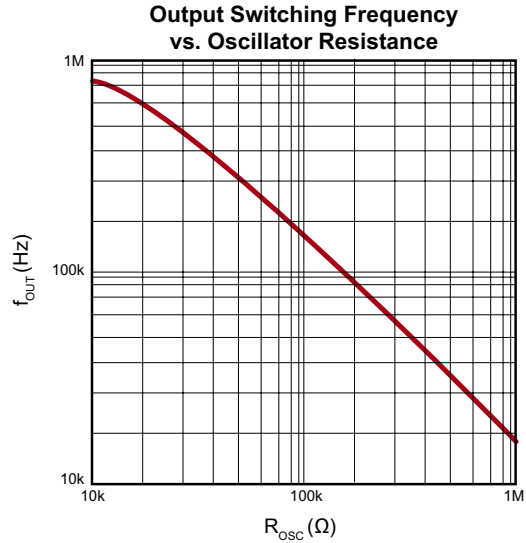
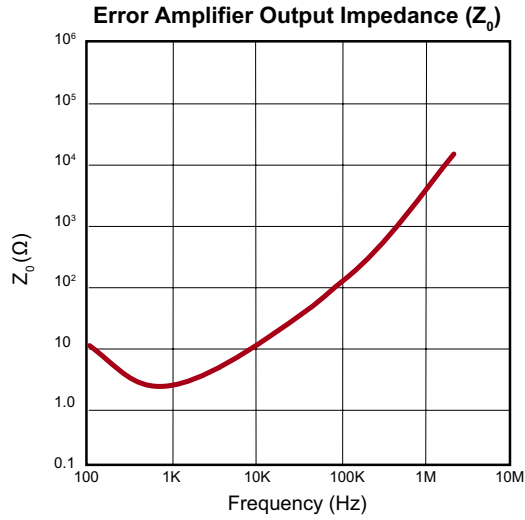
## Truth Table

SHUTDOWN	RESET	Output
H	H	Normal operation
H	H $\rightarrow$ L	Normal operation, no change
L	H	Off, not latched
L	L	Off, latched
L $\rightarrow$ H	L	Off, latched, no change

## Shutdown Timing Waveforms



# Typical Performance Curves



**Pin Descriptions**

**16-Lead SOIC (NG)**

Pin #	Description
1	+VIN
2	-
3	-
4	SENSE
5	OUTPUT
6	-VIN
7	VDD
8	OSC OUT

Pin #	Description
9	OSC IN
10	NC
11	VREF
12	<u>SHUTDOWN</u>
13	RESET
14	COMP
15	FB
16	BIAS

**16-Lead PDIP (P)**

Pin #	Description
1	+VIN
2	NC
3	NC
4	SENSE
5	OUTPUT
6	-VIN
7	VDD
8	OSC OUT

Pin #	Description
9	OSC IN
10	NC
11	VREF
12	<u>SHUTDOWN</u>
13	RESET
14	COMP
15	FB
16	BIAS

**20-Lead PLCC (PJ)**

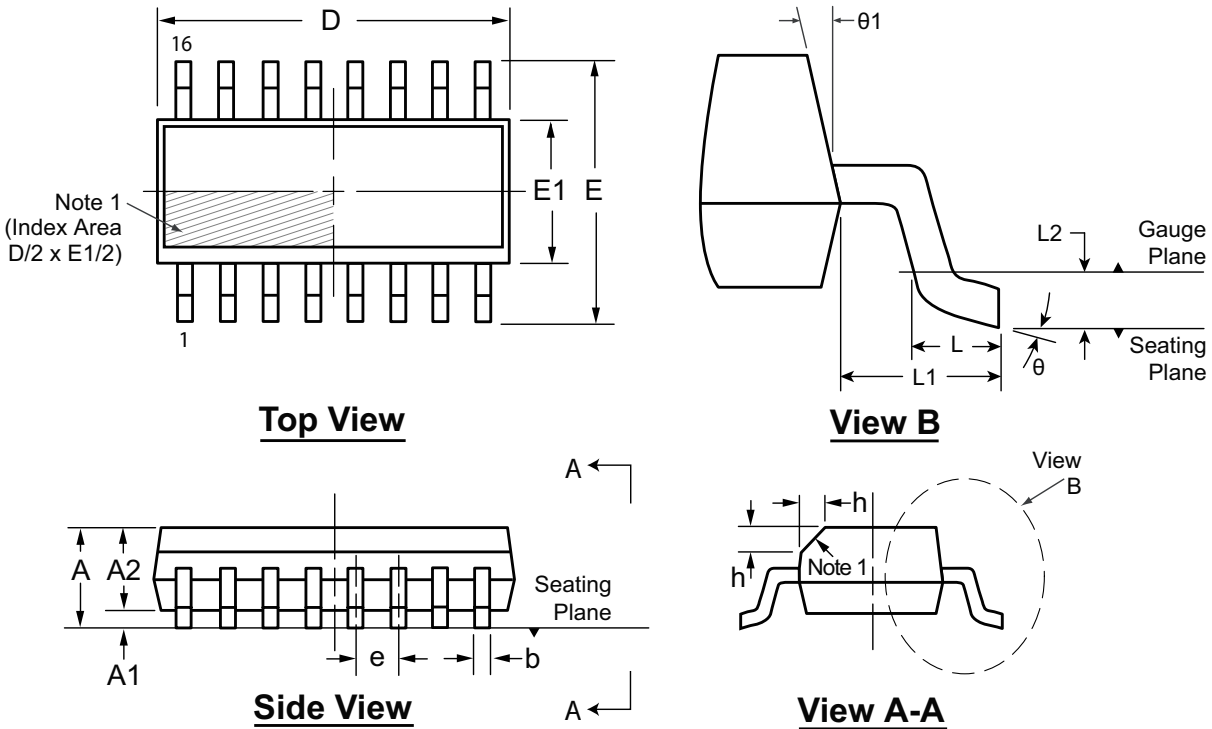
Pin #	Description
1	NC
2	NC
3	+VIN
4	NC
5	SENSE
6	OUTPUT
7	NC
8	-VIN
9	VDD
10	OSC OUT

Pin #	Description
11	OSC IN
12	NC
13	NC
14	VREF
15	NC
16	<u>SHUTDOWN</u>
17	RESET
18	COMP
19	FB
20	BIAS



# 16-Lead SOIC (Narrow Body) Package Outline (NG)

9.90x3.90mm body, 1.75mm height (max), 1.27mm pitch



**Note:**  
 1. This chamfer feature is optional. If it is not present, then a Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.

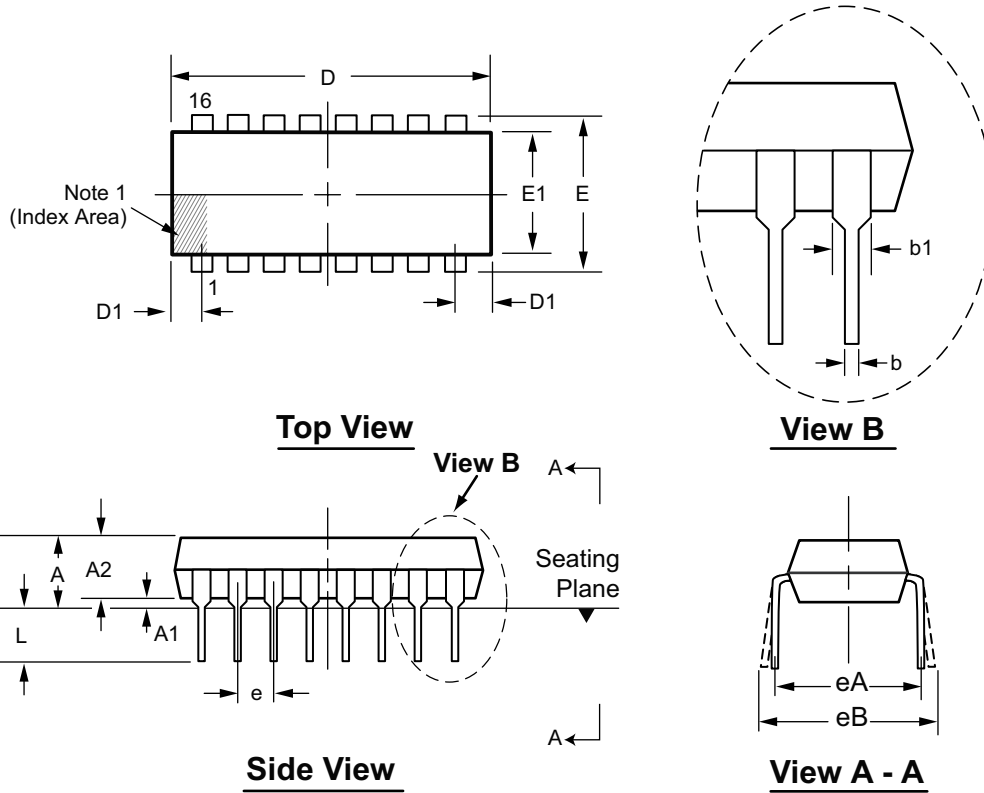
Symbol	A	A1	A2	b	D	E	E1	e	h	L	L1	L2	θ	θ1	
Dimension (mm)	MIN	1.35*	0.10	1.25	0.31	9.80*	5.80*	3.80*	1.27 BSC	0.25	0.40	1.04 REF	0.25 BSC	0°	5°
	NOM	-	-	-	-	9.90	6.00	3.90		-	-			-	-
	MAX	1.75	0.25	1.65*	0.51	10.00*	6.20*	4.00*		0.50	1.27			8°	15°

JEDEC Registration MS-012, Variation AC, Issue E, Sept. 2005.  
 \* This dimension is not specified in the JEDEC drawing.

**Drawings are not to scale.**  
**Supertex Doc. #: DSPD-16SONG, Version G041309.**

# 16-Lead PDIP (.300in Row Spacing) Package Outline (P)

.790x.250in body, .210in height (max), .100in pitch



**Note:**  
 1. A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.

Symbol	A	A1	A2	b	b1	D	D1	E	E1	e	eA	eB	L	
Dimension (inches)	MIN	.130*	.015	.115	.014	.045	.745 <sup>†</sup>	.005	.290 <sup>†</sup>	.240	.100 BSC	.300 BSC	.300*	.115
	NOM	-	-	.130	.018	.060	.790	-	.310	.250			-	.130
	MAX	.210	.035*	.195	.023 <sup>†</sup>	.070	.810 <sup>†</sup>	.050*	.325	.280			.430	.150

JEDEC Registration MS-001, Variation AB, Issue D, June, 1993.

\* This dimension is not specified in the JEDEC drawing.

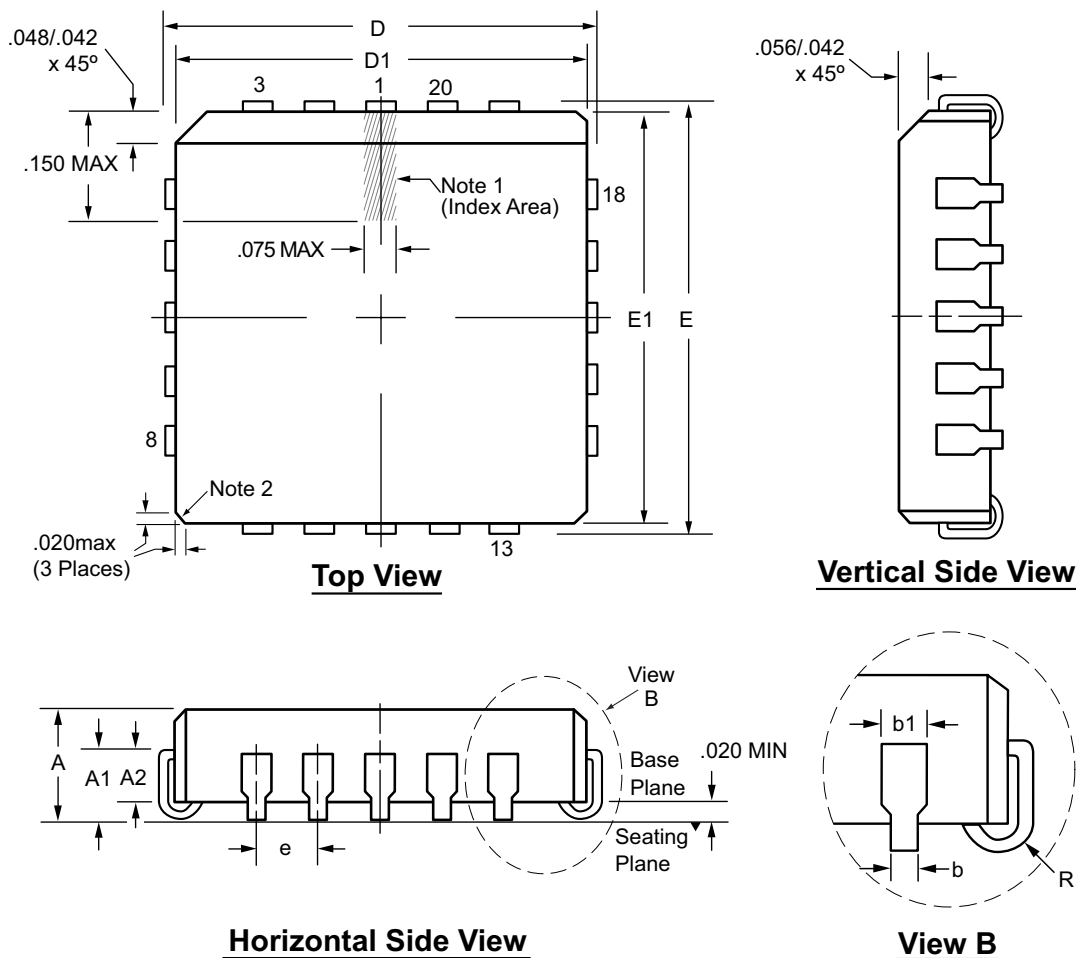
† This dimension differs from the JEDEC drawing.

**Drawings not to scale.**

**Supertex Doc. #: DSPD-16DIPP, Version C021312.**

# 20-Lead PLCC Package Outline (PJ)

.353x.353in body, .180in height (max), .050in pitch



**Notes:**

1. A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.
2. Actual shape of this feature may vary.

Symbol	A	A1	A2	b	b1	D	D1	E	E1	e	R	
Dimension (inches)	MIN	.165	.090	.062	.013	.026	.385	.350	.385	.350	.050 BSC	.025
	NOM	.172	.105	-	-	-	.390	.353	.390	.353		.035
	MAX	.180	.120	.083	.021	.032	.395	.356	.395	.356		.045

JEDEC Registration MS-018, Variation AA, Issue A, June, 1993.

Drawings not to scale.

Supertex Doc. #: DSPD-20PLCCPJ, Version C031111

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <http://www.supertex.com/packaging.html>.)

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